

100µmPET

a ultra-high-resolution silicon-pixel-based
small animal PET scanner

R.Cardella, on behalf of the collaboration



Swiss National
Science Foundation

EPFL



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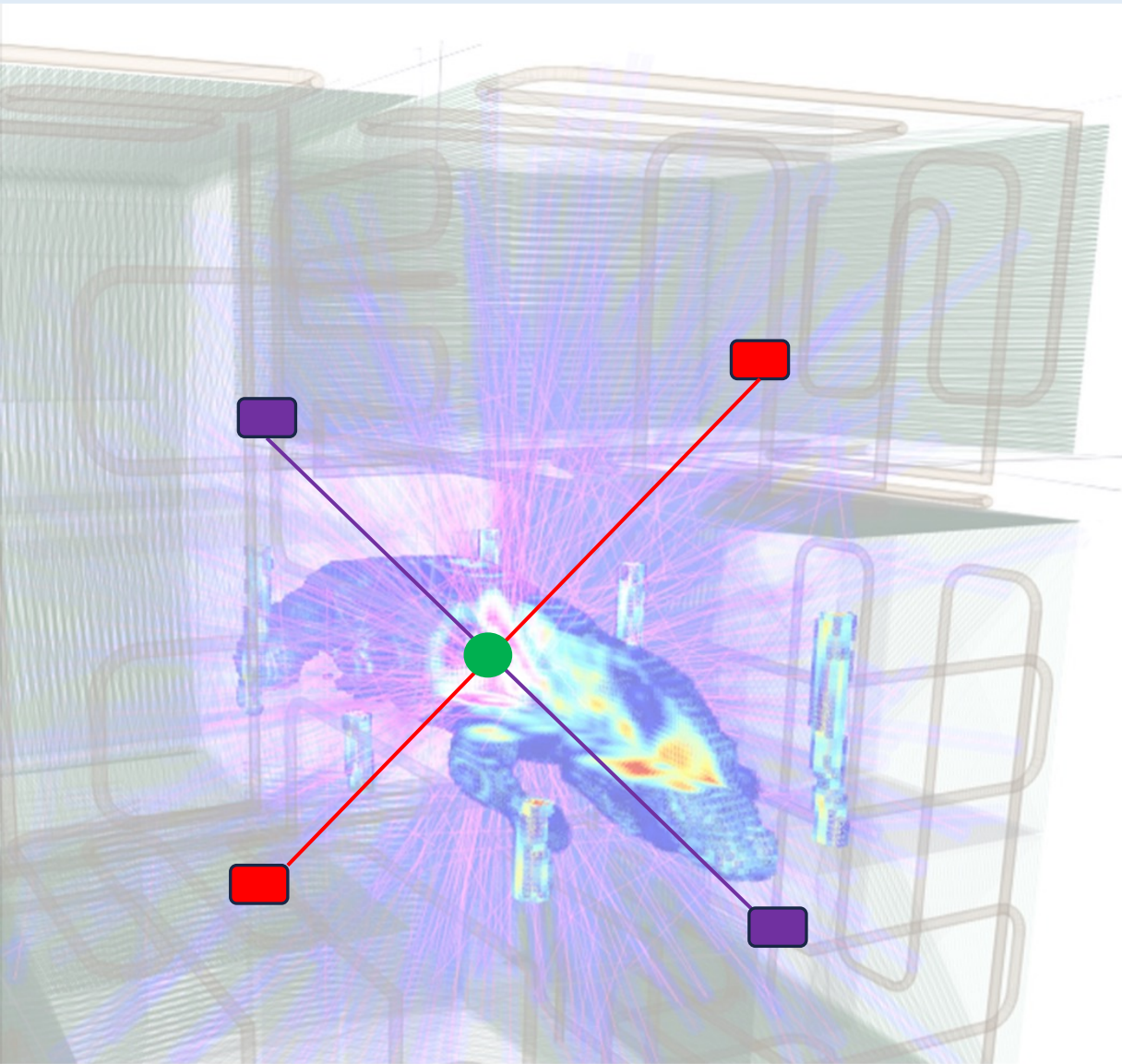


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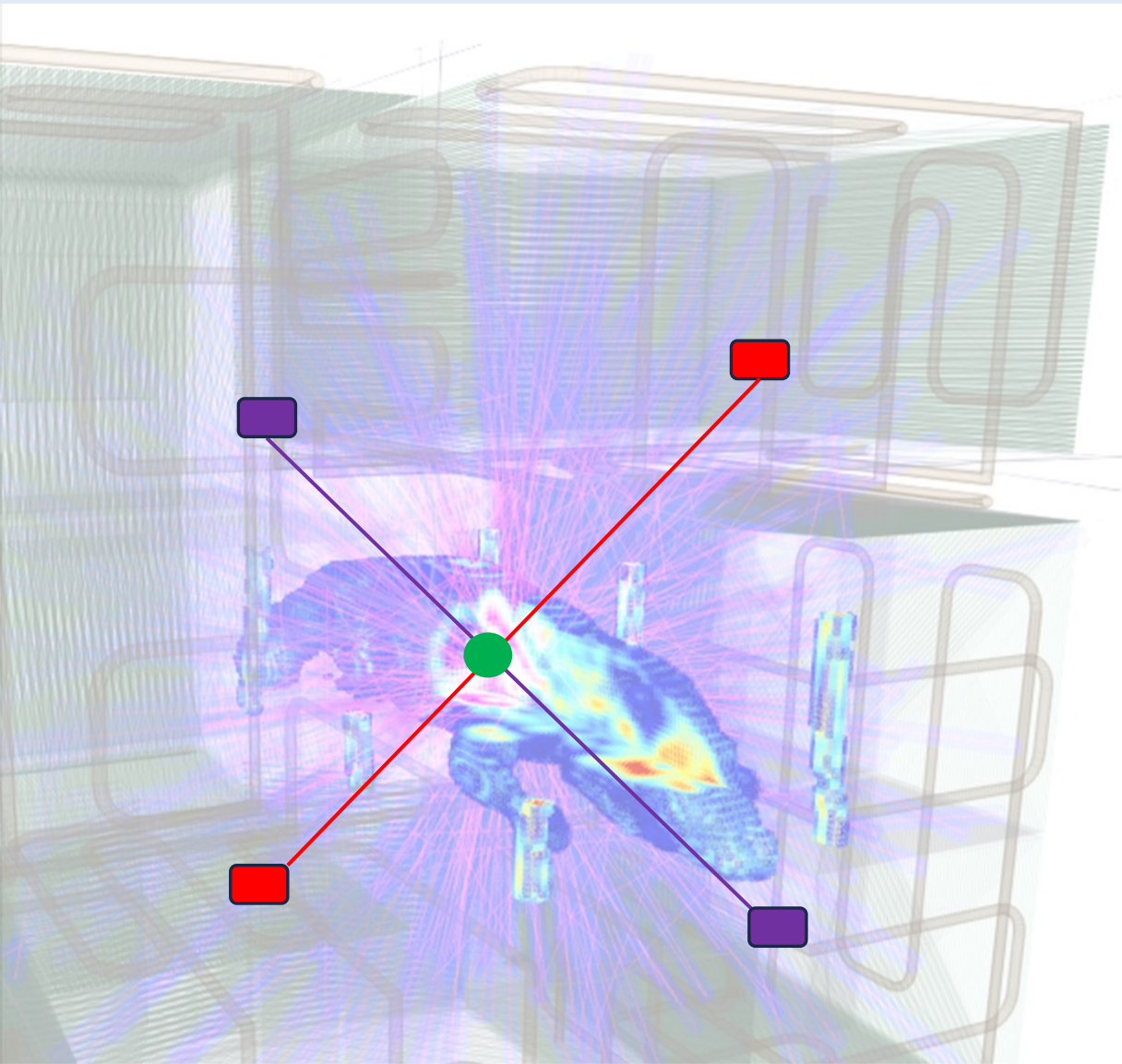
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Particle Physics



- Radiotracer injected into the body
- Cells with abnormal metabolic behavior will show an excess of radiotracers
- Positron emission -> $e^- e^+$ annihilation
- Detect 2 back-to-back Gamma at 511 keV
- Reconstruct image from Line of Response (LoR)

Atherosclerosis plaques in mouse
Pre-clinical scanner





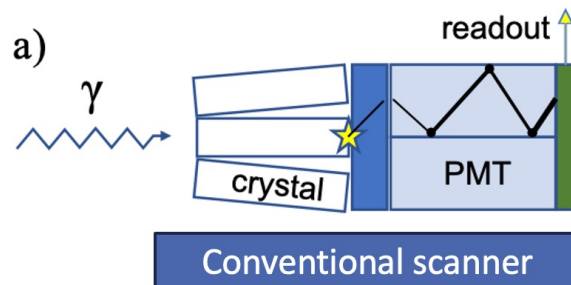
- Radiotracer injected into the body
- Cells with abnormal metabolic behavior will show an excess of radiotracers
- Positron emission $\rightarrow e^- e^+$ annihilation
- Detect 2 back-to-back Gamma at 511 keV
- Reconstruct image from Line of Response (LoR)

Image quality and Acquisition time:

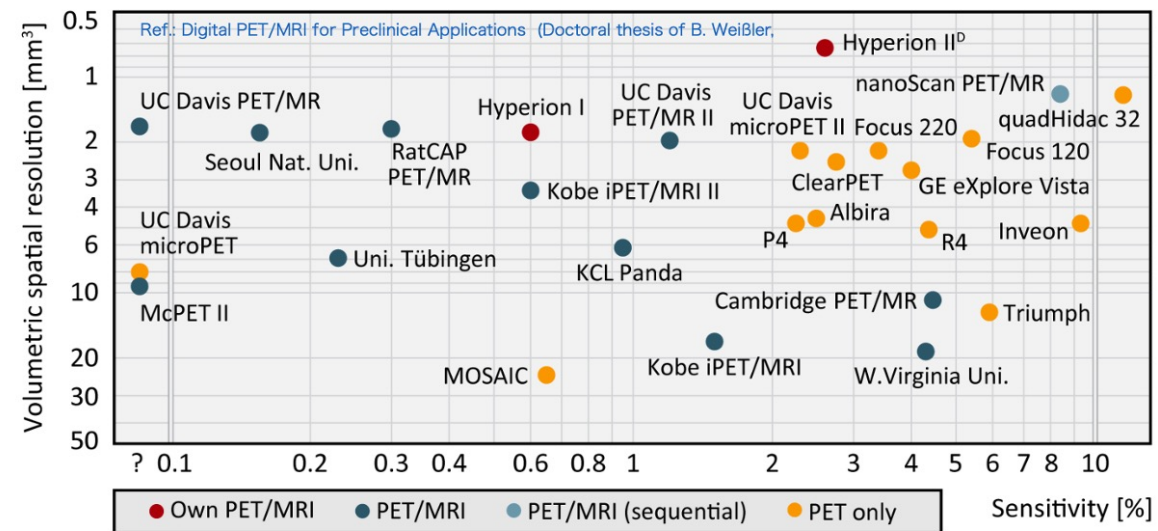
- Spatial Resolution
- Noise event rate vs. True coincidence rate
- \rightarrow Small time window and low jitter

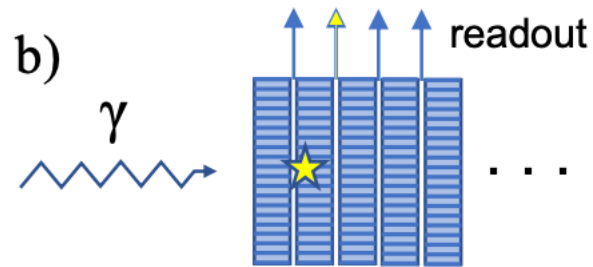
Atherosclerosis plaques in mouse
Pre-clinical scanner

Best **pre-clinical small animal** scanner combine MRI – CT with PET to achieve resolution of **0.5mm³**



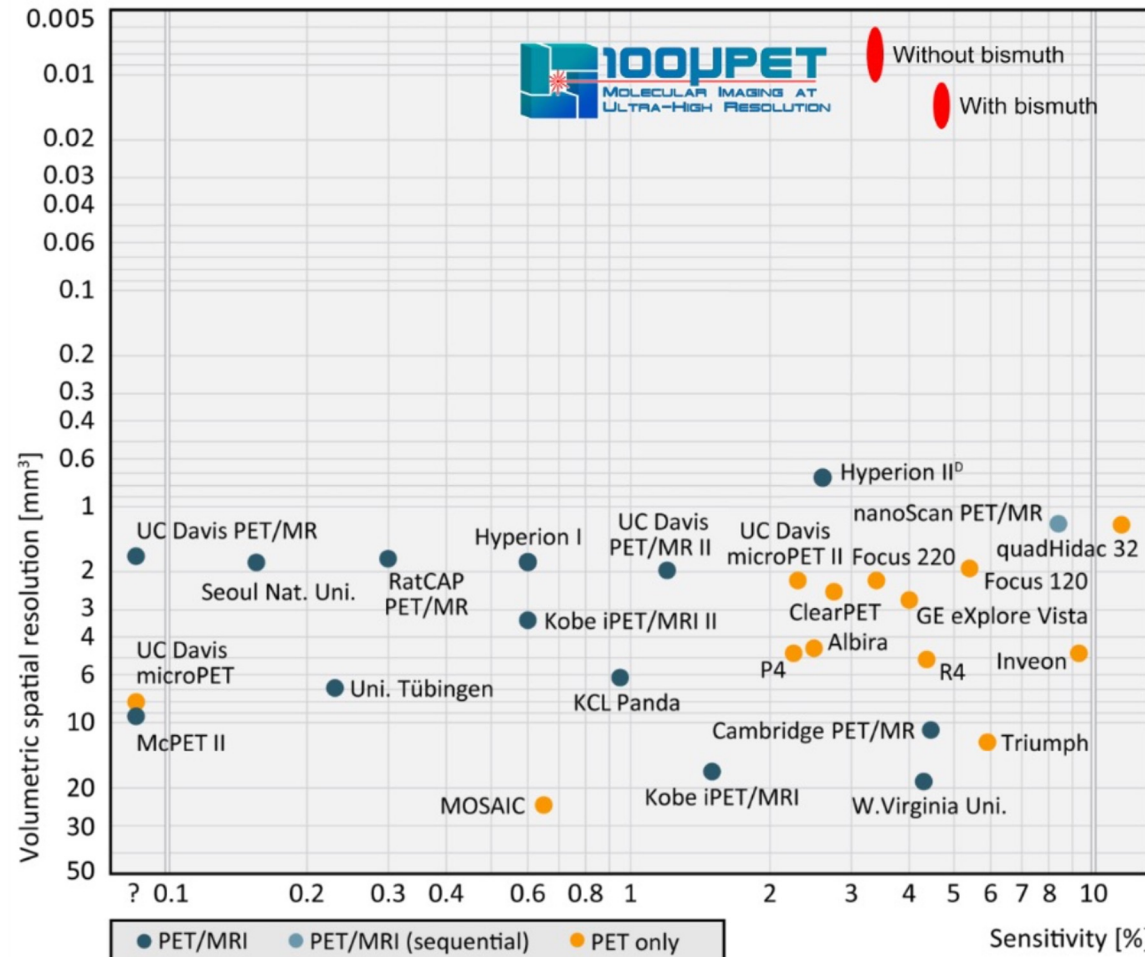
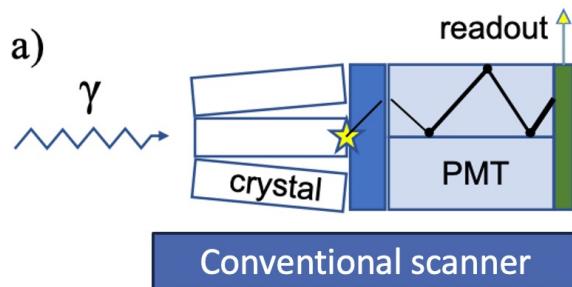
Overview of current small animal PET scanners





Our proposal (MAPS)
1 order of magnitude better!

Best **pre-clinical small animal** scanner combine MRI – CT with PET to achieve resolution of **0.5mm³**



DOI [0.1109/TRPMS.2024.3456241](https://doi.org/10.1109/TRPMS.2024.3456241)

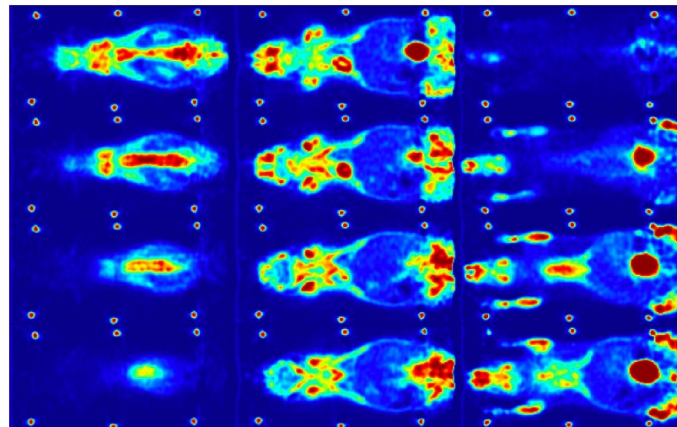


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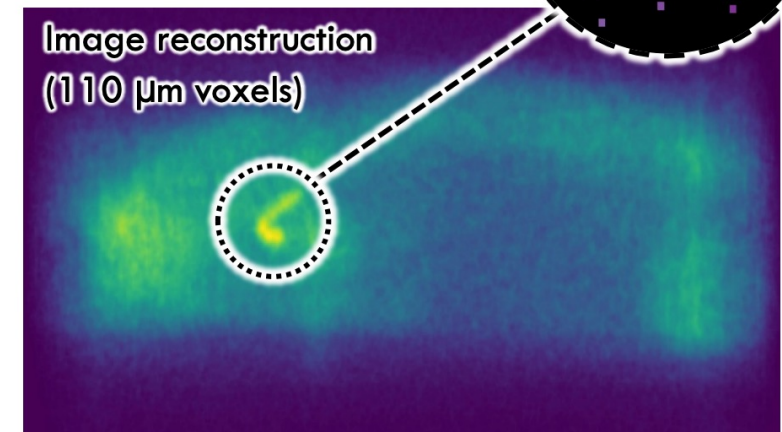
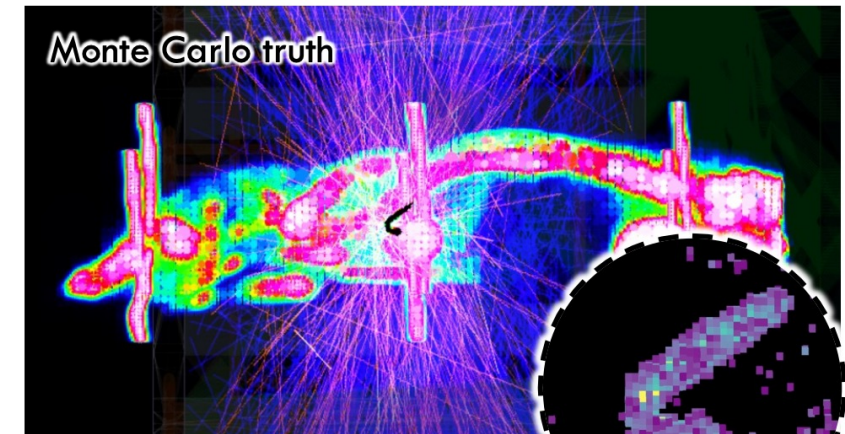
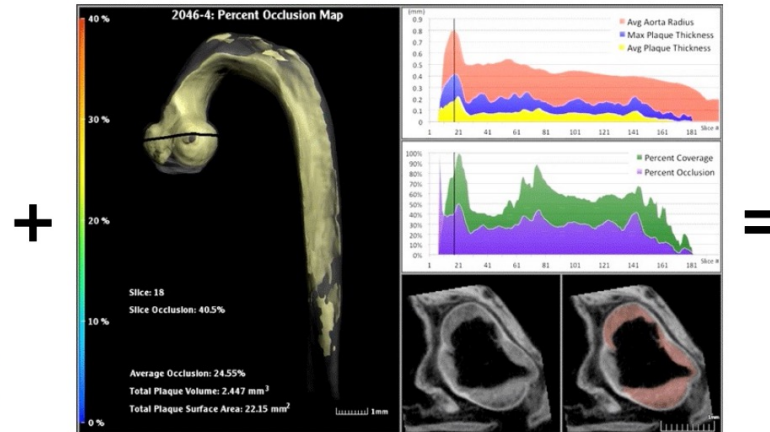
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- Mouse phantom from Digimouse PET (background) + atherosclerotic plaque microCT (signal of interest)
 - ▣ Injected activity: 30 MBq
 - ▣ Measurement time: 20 min
 - Mouse volume = 45 mL → 1.4×10^9 background annihilations
 - Plaque volume = 0.006 mL → 8.4×10^6 plaque annihilation

Digimouse PET section



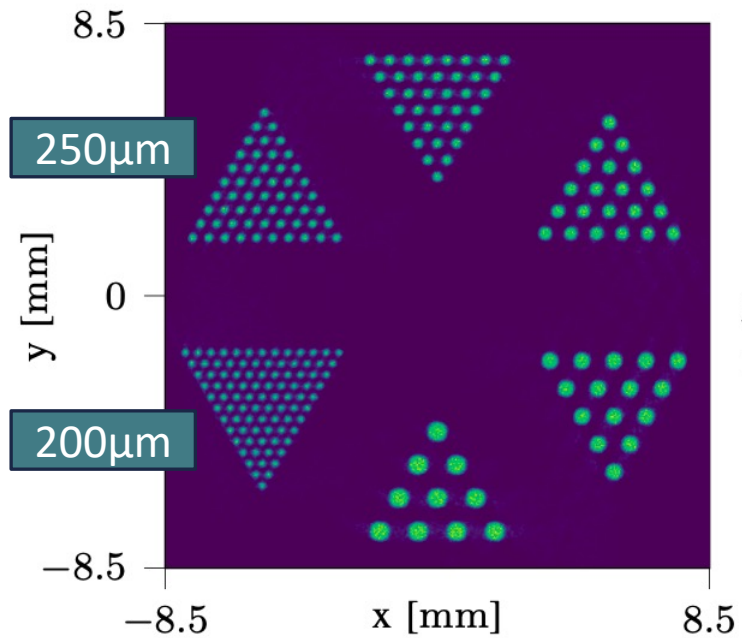
microCT of atherosclerotic plaque



PIXEL 2024

How far can we
bring down the spatial resolution?

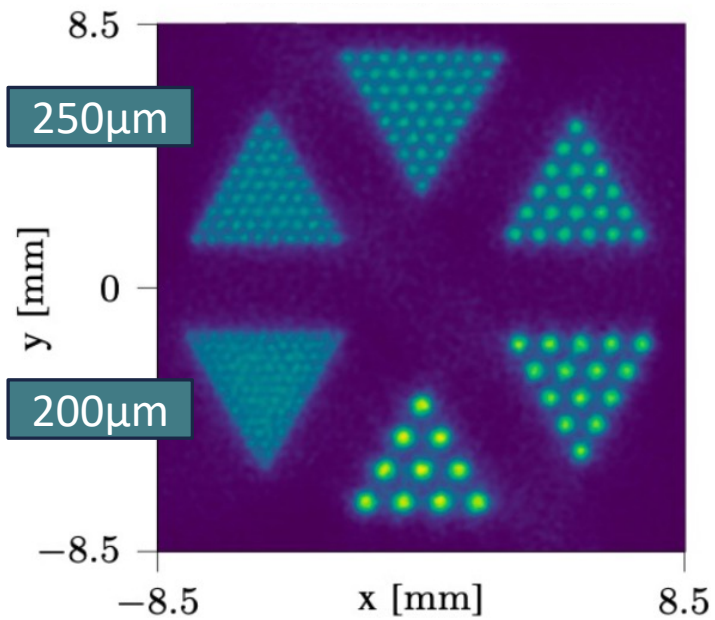
100μm pixel pitch



No acollinearity,
No positron range,
 $p = 0.1$ mm, no bismuth,
reconstructed with OSEM

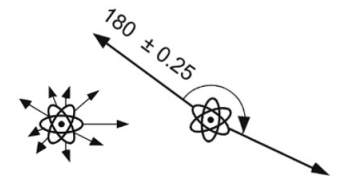
No uncertainty in Gamma emission process

100μm pixel pitch



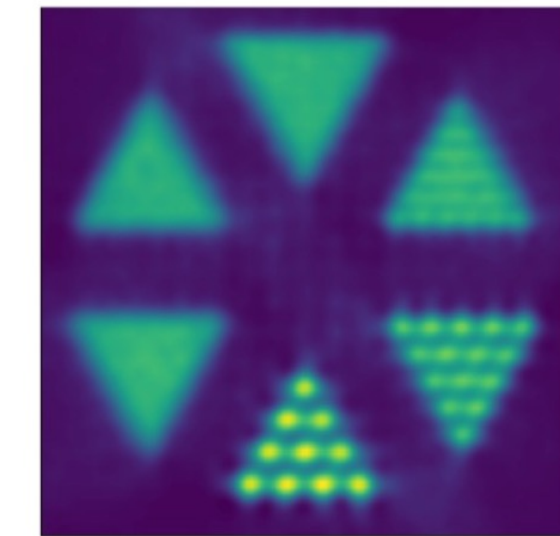
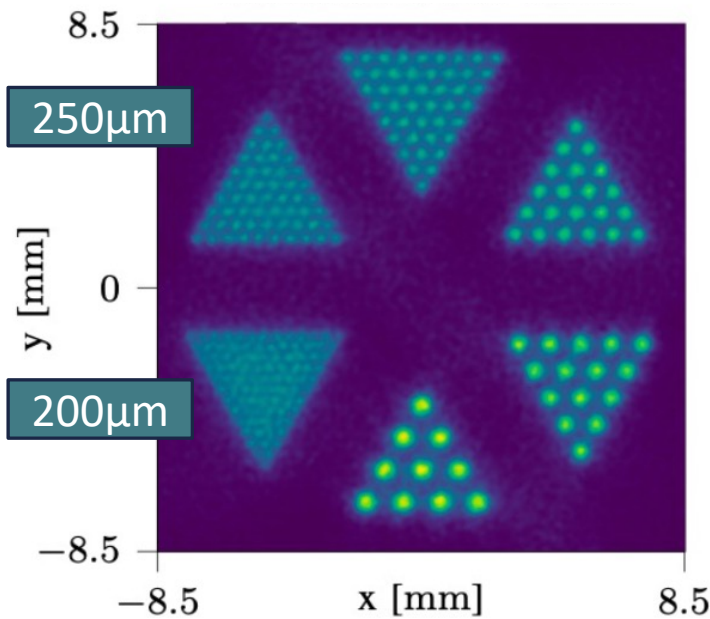
Positron range: 102μm FWHM for F18

Acolinearity: $90\mu m \approx 4 \cdot 10^{-3} R$



100μm pixel pitch

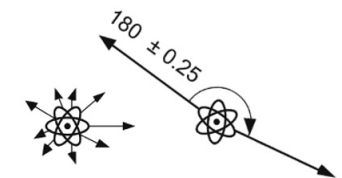
Reference scanner



Reference scanner ~ state of the art

Positron range: 102μm FWHM for F18

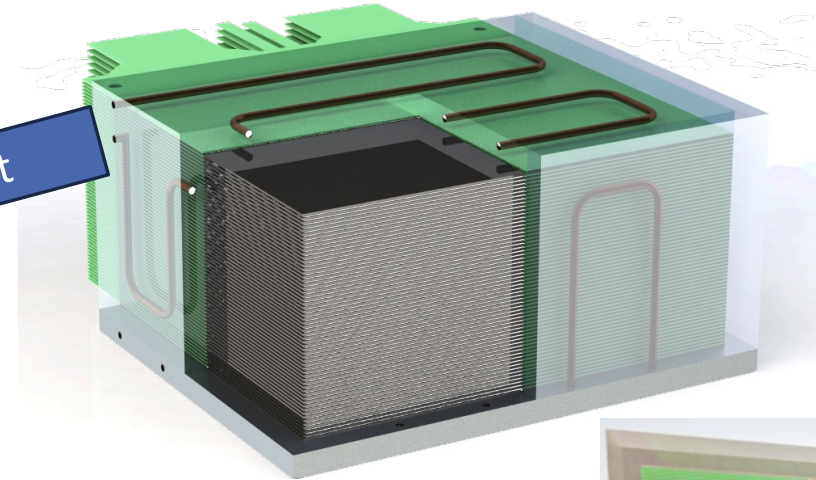
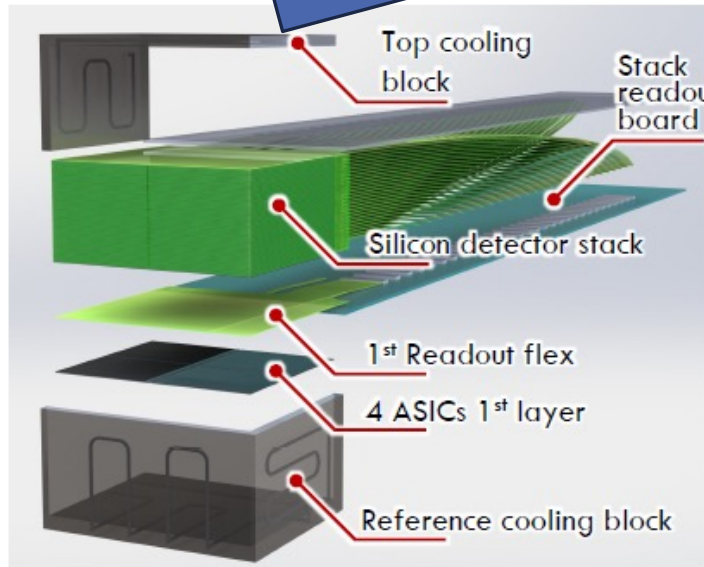
Acolinearity: $90\mu m \approx 4 \cdot 10^{-3} R$



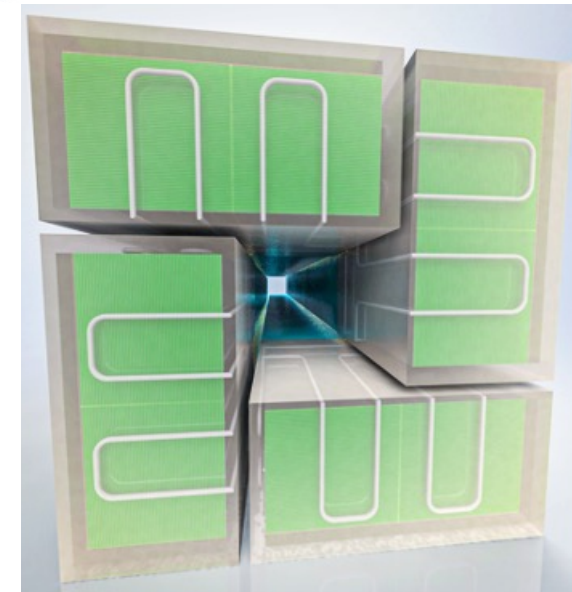
Large improvement with respect to reference scanner

Tower: 60 quad-module layers (240 chips)

Trade-off Sensitivity / Cost



100μPET Scanner:
4 towers (960 chips)



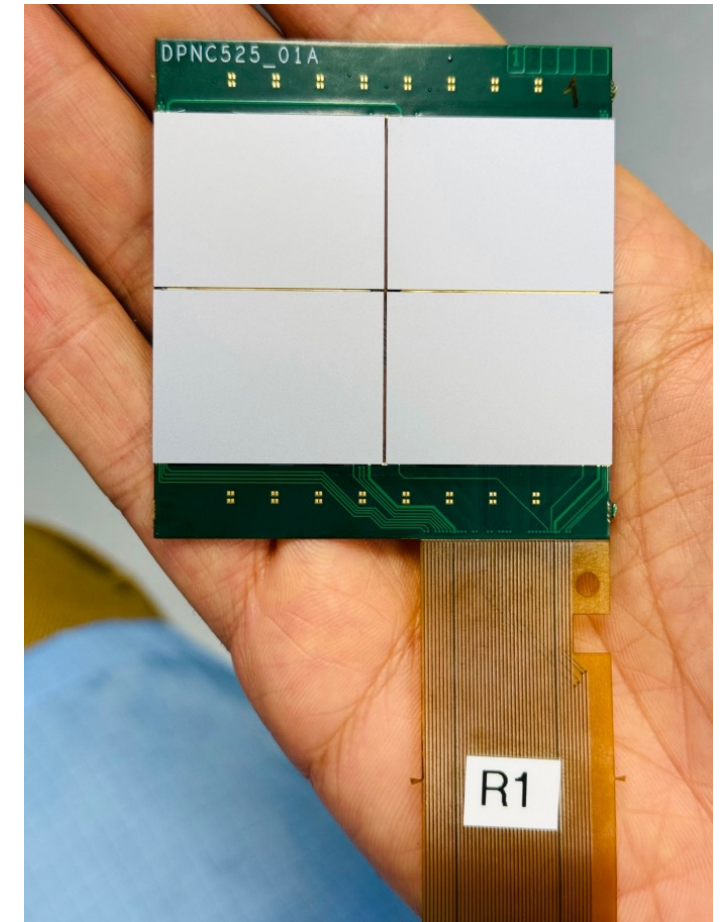
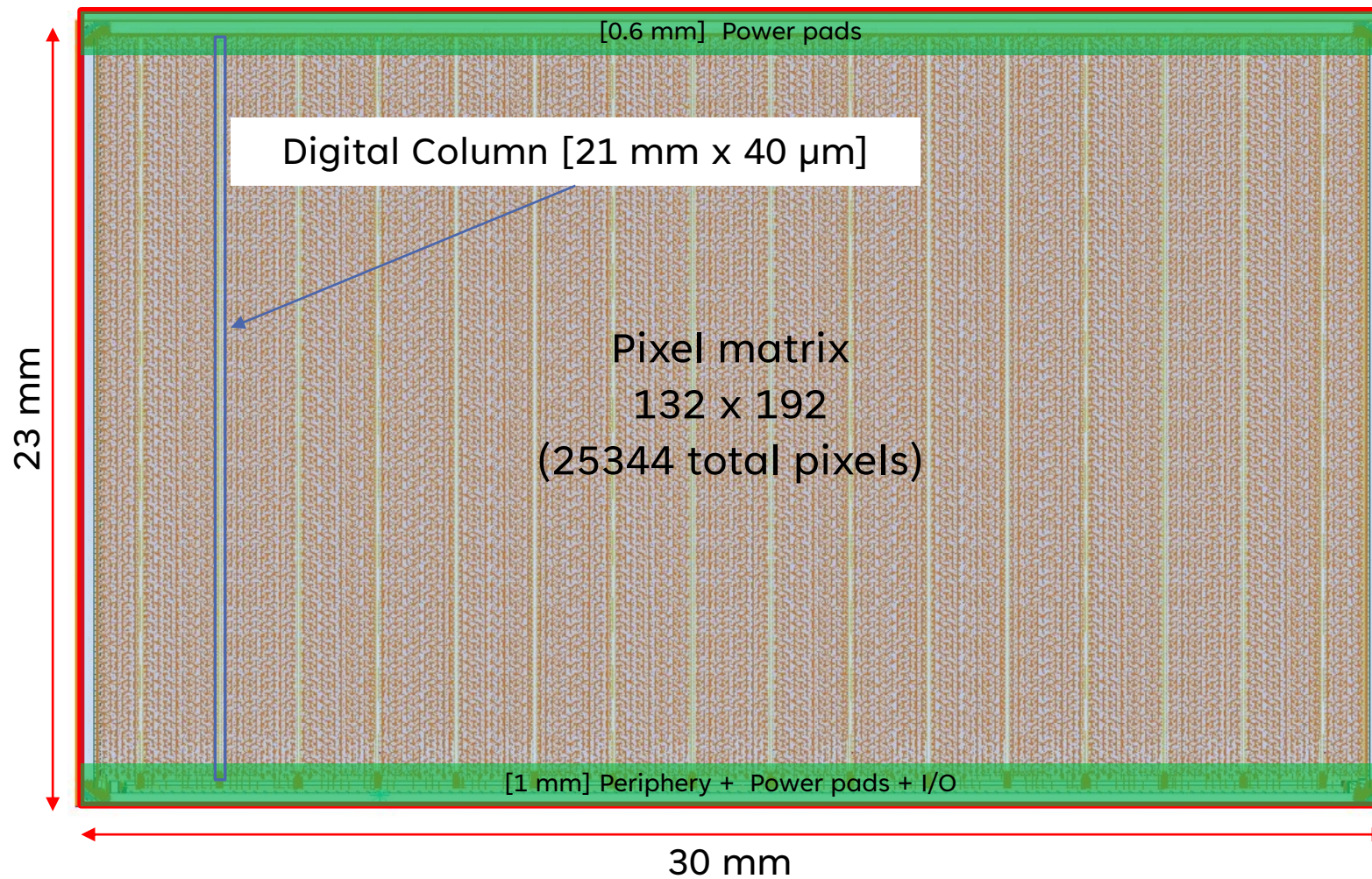
F. Cadoux et al.

Event size	1 cluster (< (5x5) pixels)
Event rate	10 kHz/cm ²
Equivalent Noise Charge (ENC)	~250 e ⁻
Operation Threshold	[2000 / 3000] e⁻
Time resolution RMS (Q _{in} > 7 ke ⁻)	200 ps
ToA	Yes, seed pixel
ToT	Yes, for time-walk correction
Power consumption	< 100 mW/cm²
Pixel pitch	150 μm

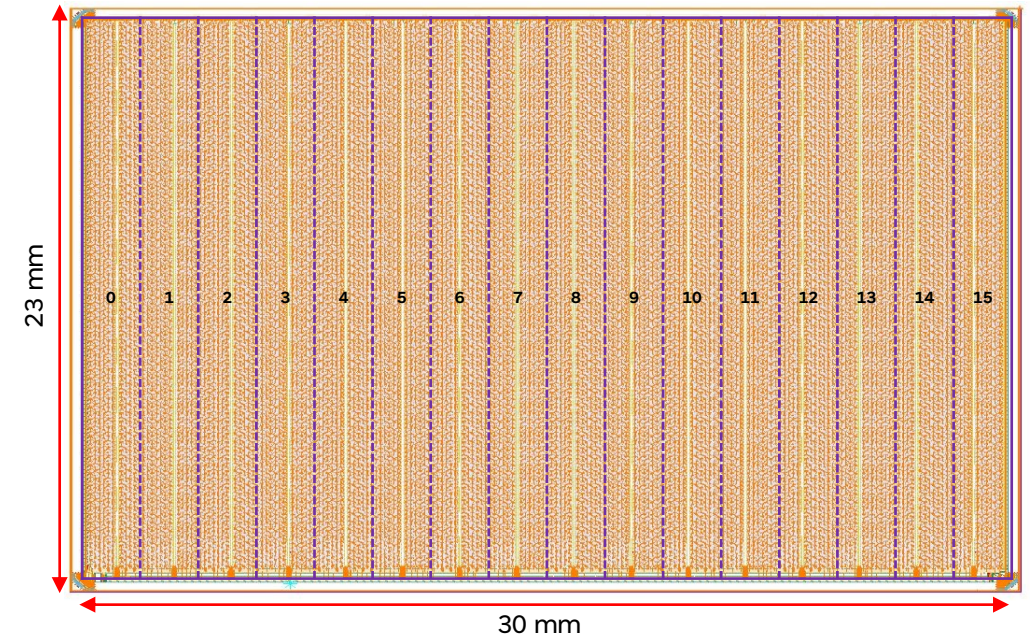
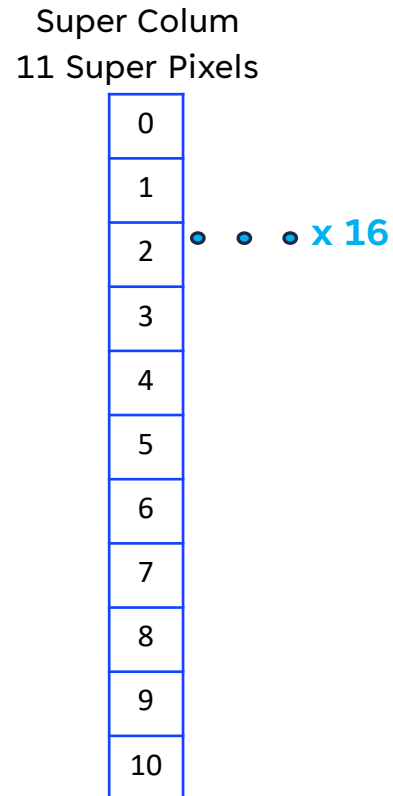
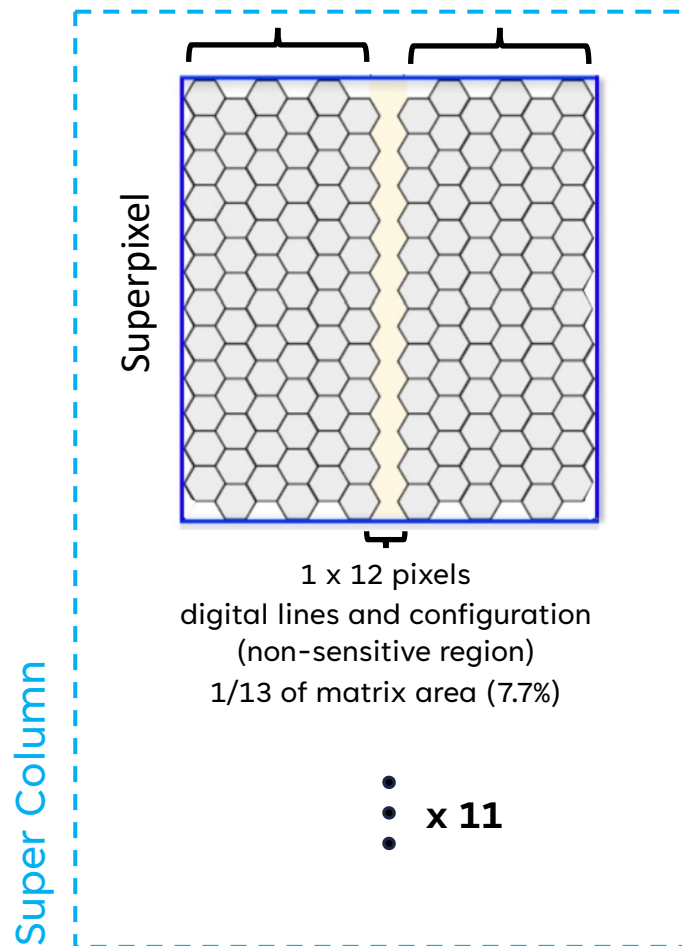
CMOS monolithic sensor in IHP SG13G2 BiCMOS, **130 nm** process featuring **SiGe HBT**

See also talk from L. Paolozzi next Thursday





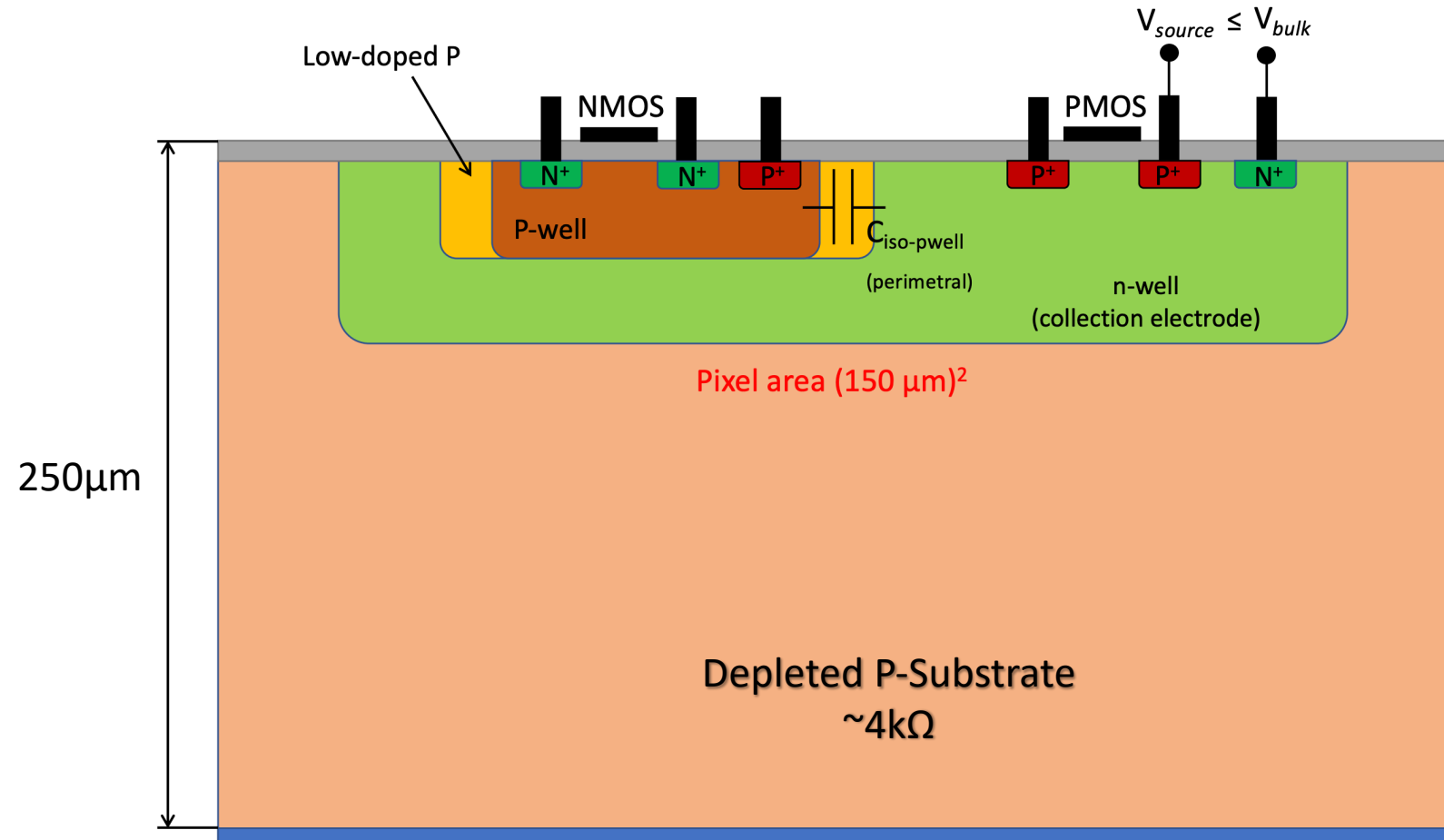
Submitted for production in October 2023



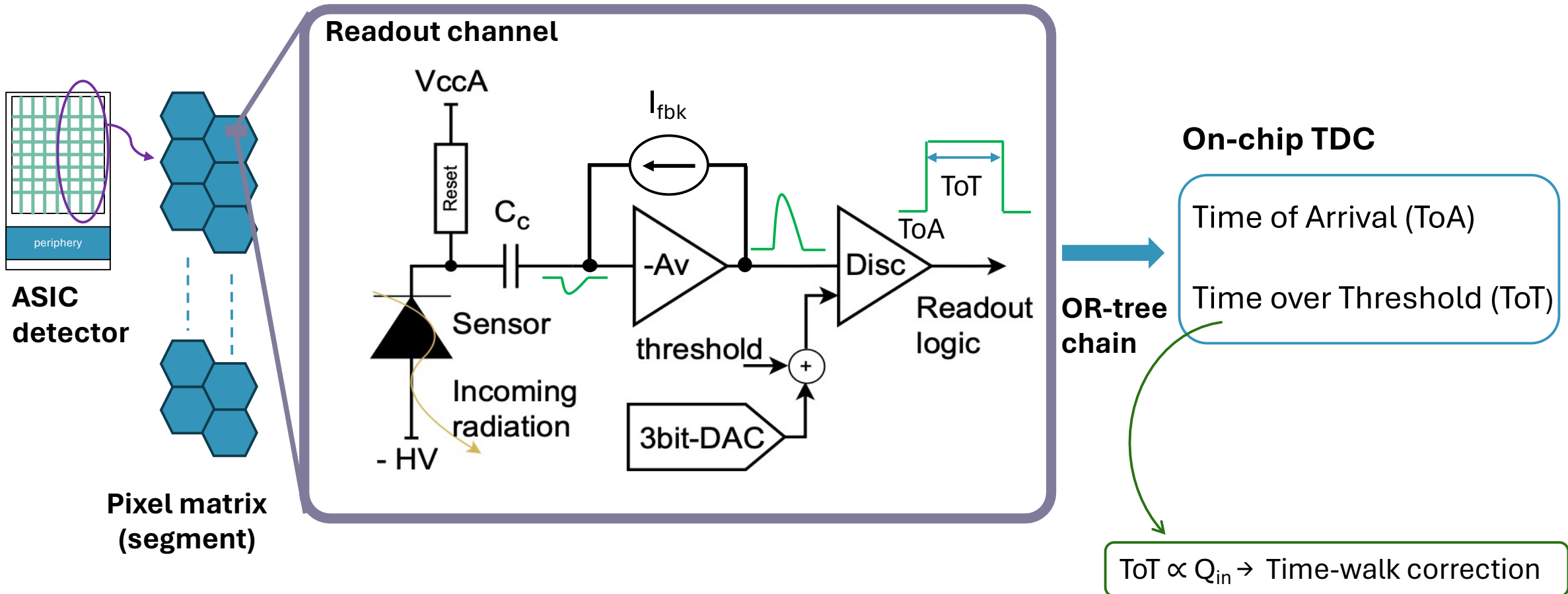
↓
Asynchronous 51-bit bus
No Clock inside the matrix

Digital Periphery

TDCs, DACs, Bandgap, LVDS drivers, config registers



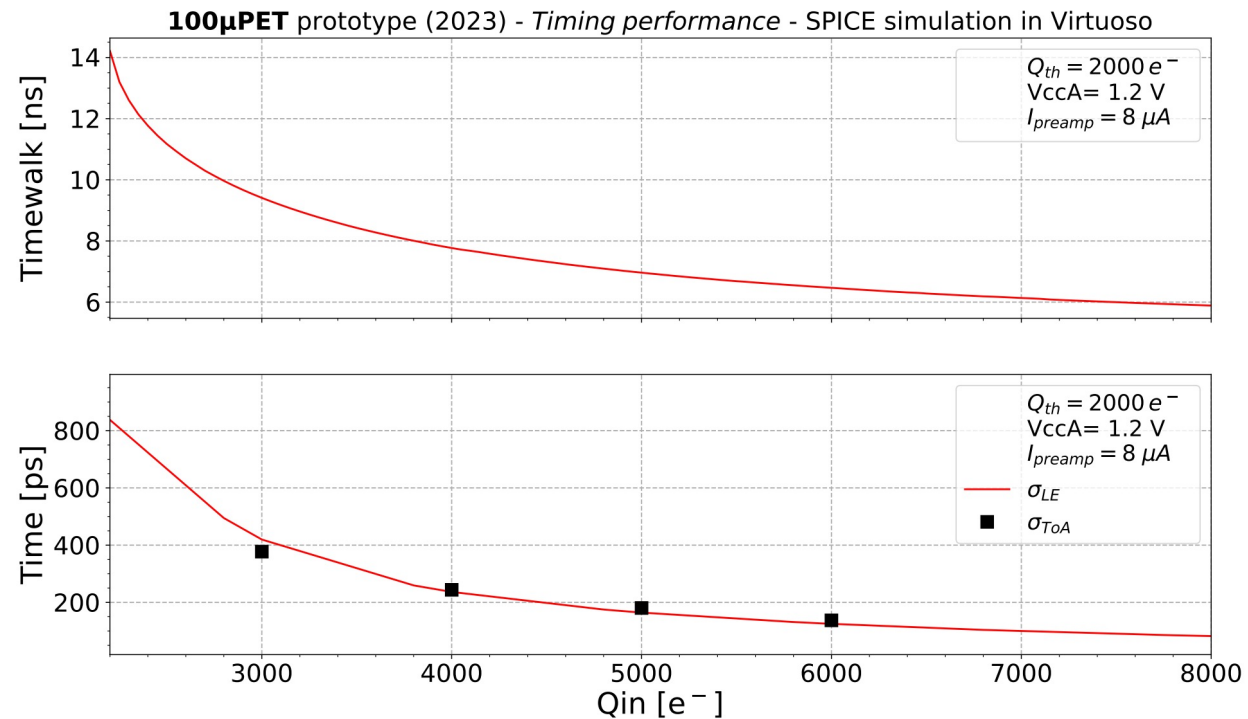
- In-pixel electronics
 - Affects input capacitance
- Non-isolated PMOS
 - Need to reduce in-pixel digital circuits
 - 360fF pixel capacitance



A. Picardi et al.

Good time resolution to select
coincidence events from noise

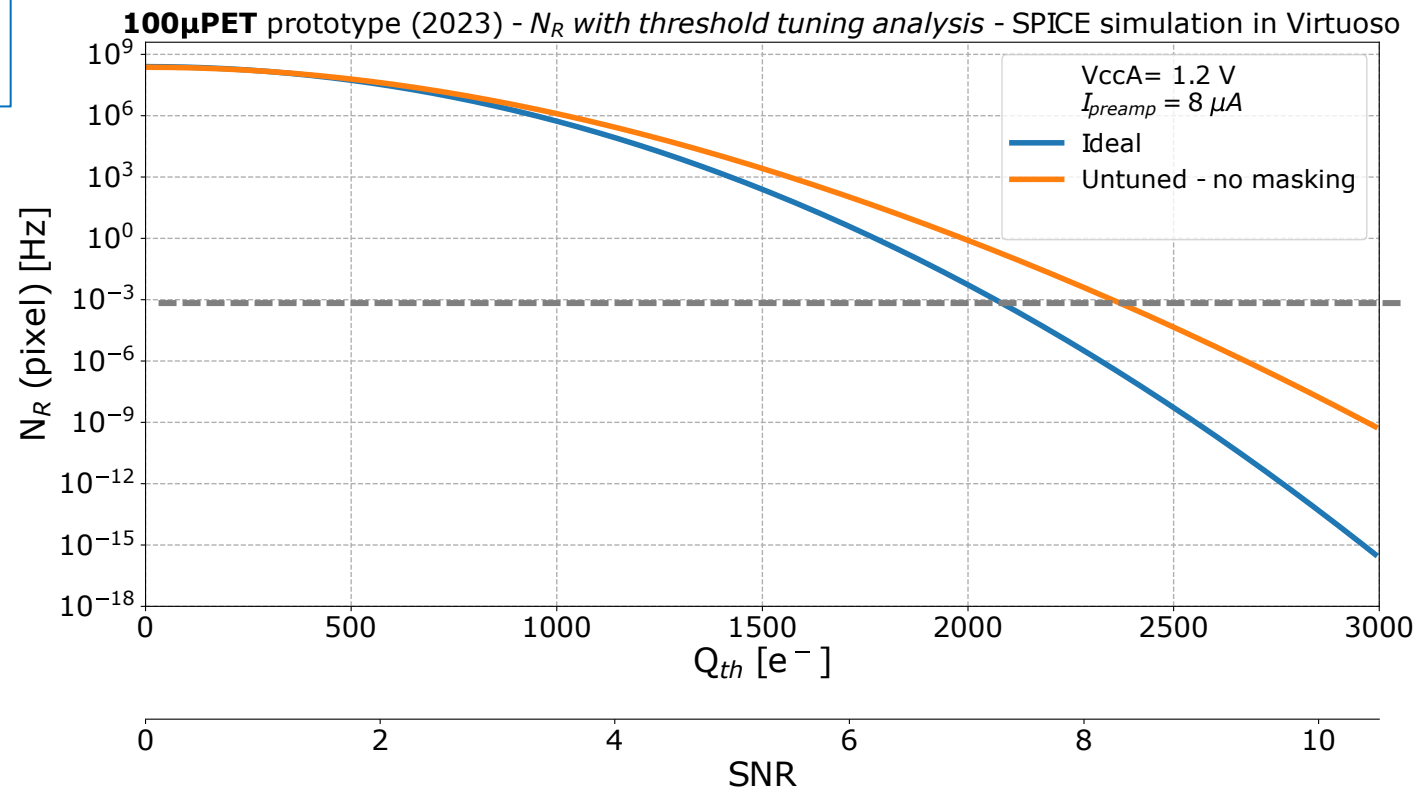
Use 1 ns time window
“4D-Tracking”



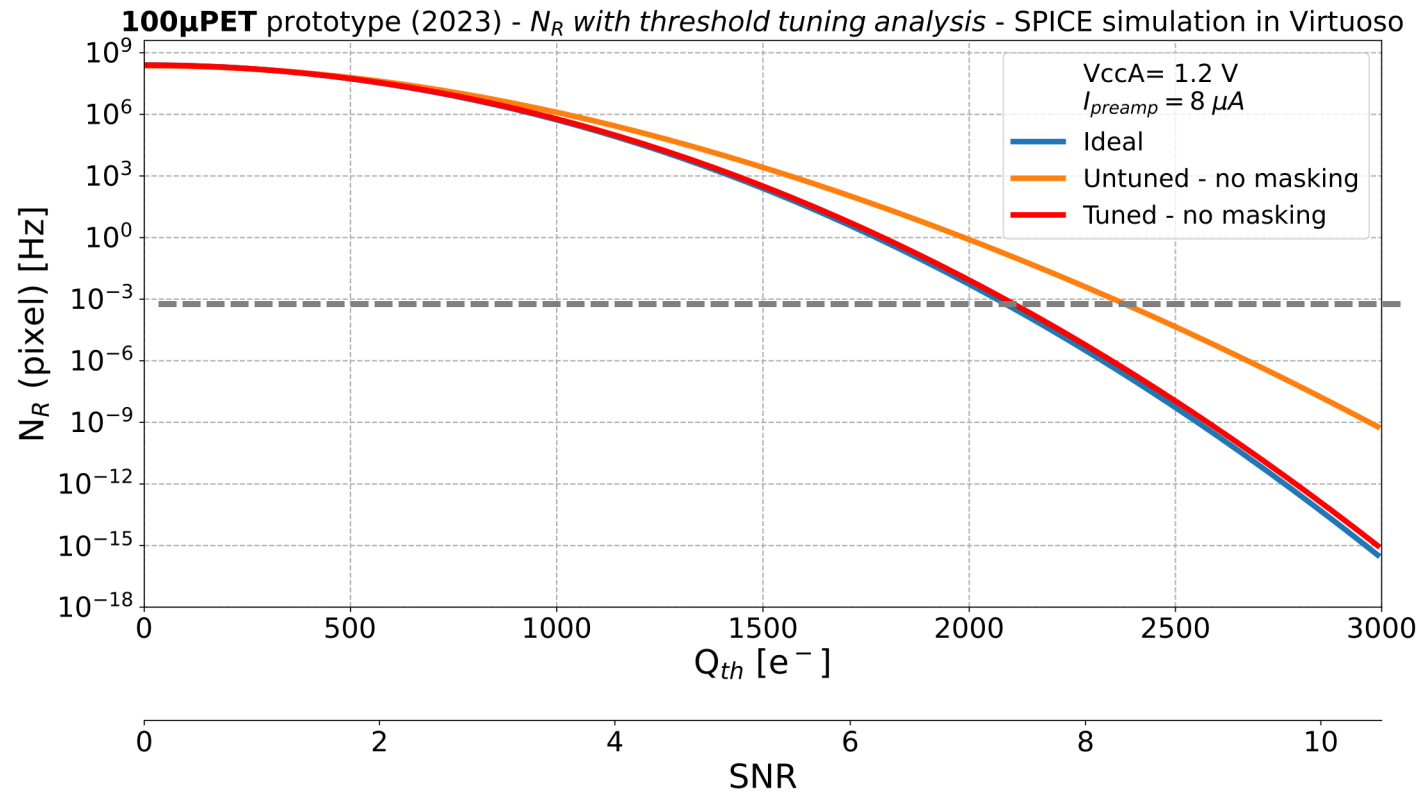
Excellent timing performance with very low power consumption.

Want to see what SiGe can do with little increase in power consumption?
 Don't miss the talk from L. Paolozzi next Thursday

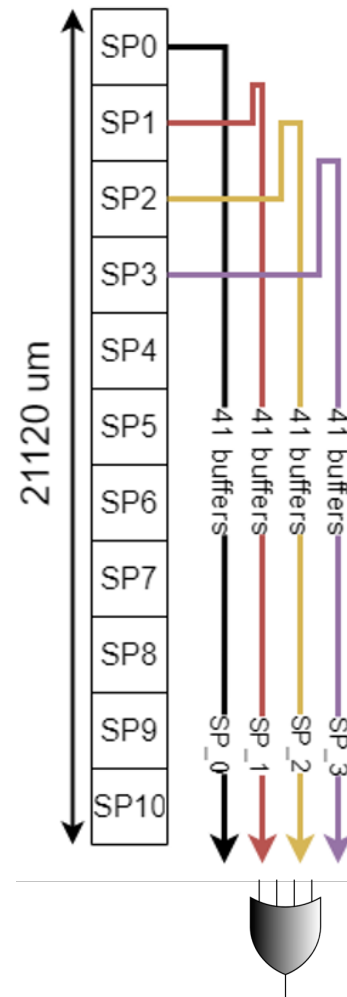
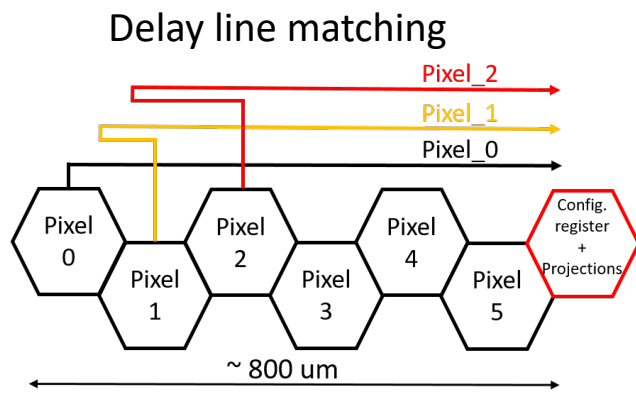
$$N_R = f_{n0} \cdot e^{\frac{-Q_{th}^2}{2 \cdot ENC^2}}$$



Mismatch worsens the Noise Rate. With threshold tuning, we can achieve the required Noise Rate



Mismatch worsens the Noise Rate. With threshold tuning, we can achieve the required Noise Rate



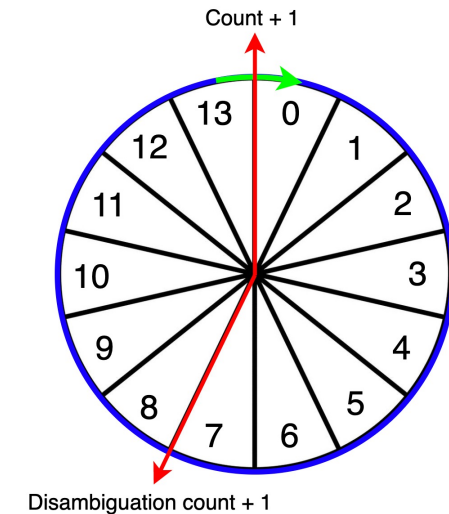
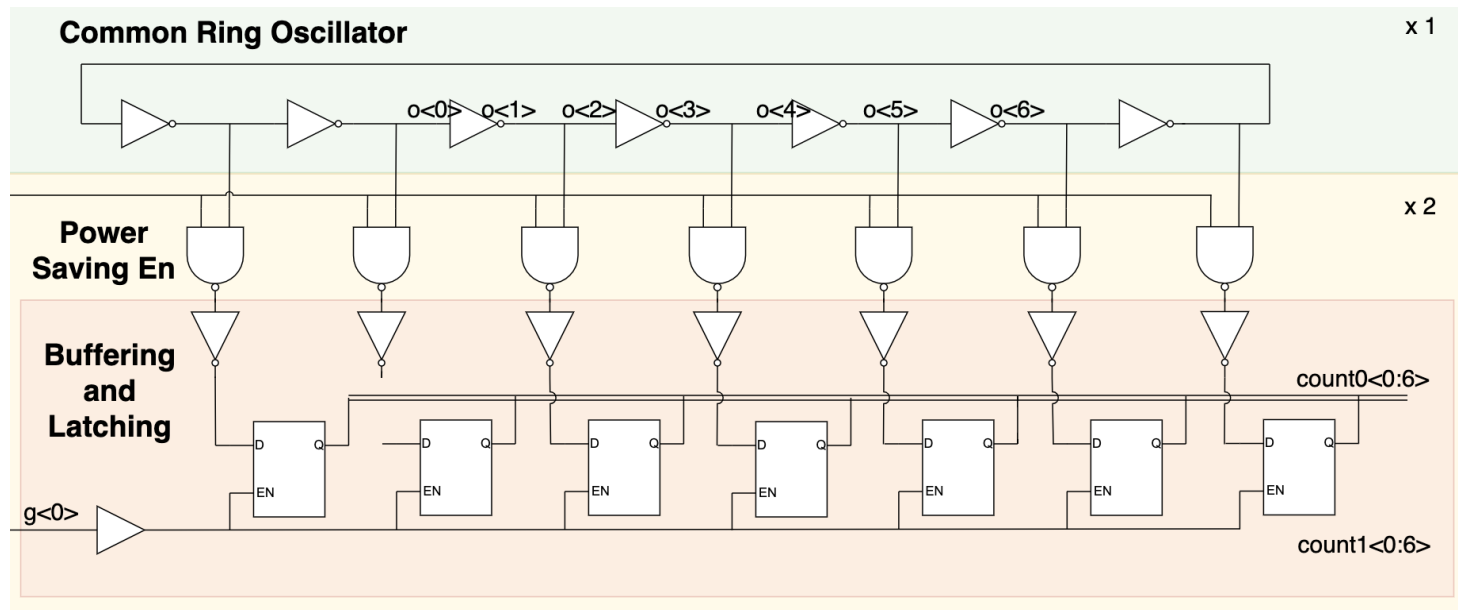
Simulation results

Max skew	67 ps
Mismatch rms	11 ps
Jitter rms	0.5 ps
T. coeff.	0.6 ps/C
V. coeff.	7.3 ps/mV

**No pixel-to-pixel
delay calibration
required**

11 Super-Pixel output fast-OR

1 TDC channel/Super Column



Supercolumn independent Flash-TDC with 7-stages free-running ring oscillator.

- States of the oscillator encode the fine-time: expected **LSB = 150ps**.
- Course timing with a counter on stage 0 and disambiguation bit on stage 1.
- **Reference** and **Calibration's** channel signals are generated per Supercolumn.
- Reference : Synchronization with DAQ for coincidence with an internal counter at readout clock frequency (50MHz).

- We are building a small-animal PET with ultra-high resolution
 - All silicon (**MAPS**): 960 sensors to build the full scanner
 - Design validated with simulations
 - **Volumetric** resolution between **0.01 mm³** and **0.02 mm³**
 - **Sensitivity** from **3%** to **5.5%**
- Currently building a **thermal Mock-up scanner** to validate cooling and assembly
 - NCP assembly validated with mock-ups
- MAPS submitted in October 2023
 - 130nm SiGe BiCMOS
 - Asynch readout
 - 150μm pitch, 200 ps time resolution, 2ke⁻ ENC 280-e @ <100mW cm²

The people

The 100μPET project



Giuseppe Iacobucci
 • project P.I.
 • System design



Yannick Favre
 • Board design
 • RO system



Roberto Cardella
 • ASIC Design (Digital)
 • Sensor Design



Sergio Gonzalez-Sevilla
 • System integration
 • Laboratory test



Carlo A. Fenoglio
 • ASIC Design (Digital)
 • Firmware



Matteo Milanesio
 • Laboratory test
 • Data analysis



Jordi Sabater
 • Detector simulation
 • Laboratory test



Andrea Pizarro
 • Laboratory test
 • Data analysis



Lorenzo Paolozzi
 • Sensor design
 • Analog electronics



Stéphane Débieux
 • Board design
 • RO system



Mateus Vicente
 • System integration
 • Laboratory test



Terry Baltus
 • Board design
 • RO system



Théo Moretti
 • Laboratory test
 • Data analysis



Chiara Magliocca
 • Laboratory test
 • Data analysis



Rafaella Kotitsa
 • Sensor simulation



Ivan Semendyaev
 • Laboratory Tests
 • Data Analysis



Didier Ferrere
 • System integration
 • Laboratory test



Franck Cadoux
 • Mechanical design



Jihad Saidi
 • Laboratory test
 • Data analysis



Sébastien Cap
 • Board design
 • RO system



Stefano Zambito
 • Laboratory test
 • Data analysis



Leonardo Cecconi
 • Chip Design
 • Firmware



Antonio Picardi
 • ASIC Design (Analog)
 • Front-end



Thanushan Kugathasan
 • Lead chip design
 • Analog



Luca Iodice
 • ASIC Design (Analog)
 • TDC



Martin Walter
 • P. I.



Pablo Jané
 • Nuclear Medicine
 • PET imaging
 • Translational imaging



Vincent Taelman
 • Molecular biology
 • Radiopharmacy



Michäel Unser
 • P. I.



Aleix Boquet-Pujadas
 • Signal/image processing
 • Physical modeling



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 INFN Rome2 & UNIGE



Holger Rucker
 IHP Mikroelektronik



Marzio Nessi
 CERN & UNIGE



Matteo Elviretti
 IHP Mikroelektronik

Funded by:

Swiss National
Science Foundation



European Research Council
 Established by the European Commission



Thank you for your attention



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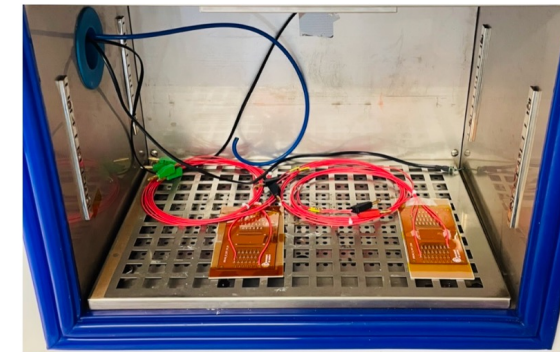
Backup slides



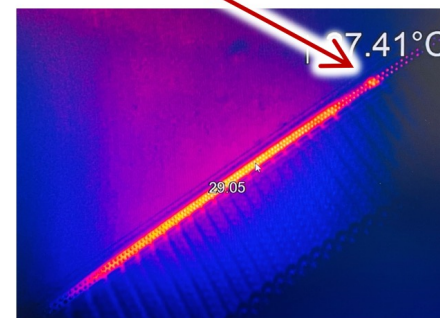
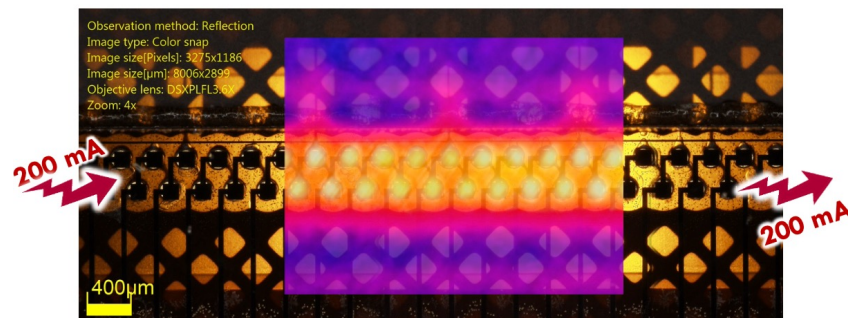
Au studs with NCP bonding was chosen for the assembly

- 100% yield connection among hundreds of pads
- (Direct) Current stress-tests to verify bonding failure
 - Limit DC to 200 mA, avoiding local heating exceeding T_g of the glue (60-80 °C)
 - Visible permanent defects (hot spots – higher bonding resistance) after reaching 300 mA
- 100 TCs from +5 to +60 °C
 - No effect on connections

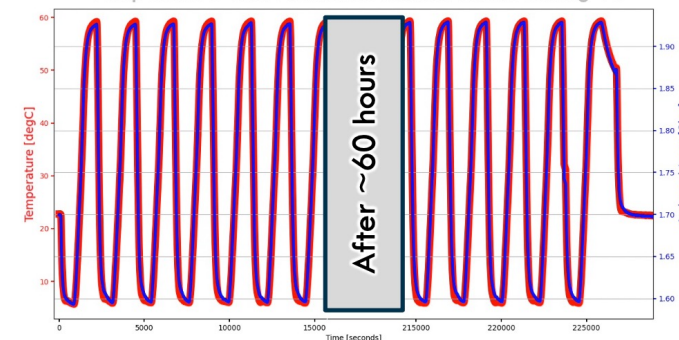
Bonded samples inside climate chamber



Bonded assembly and IR thermal image during DC injection



Temperature and resistance measurement during TC



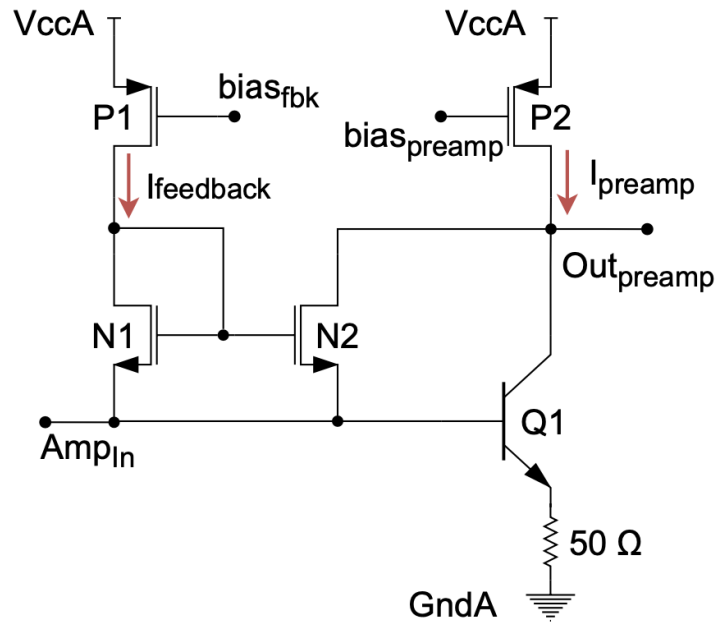
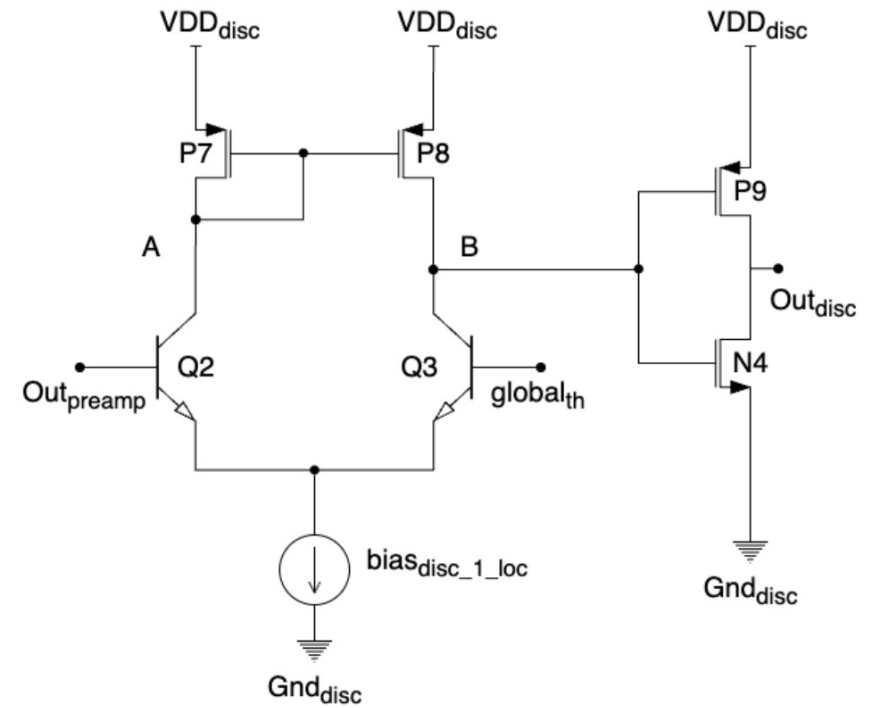
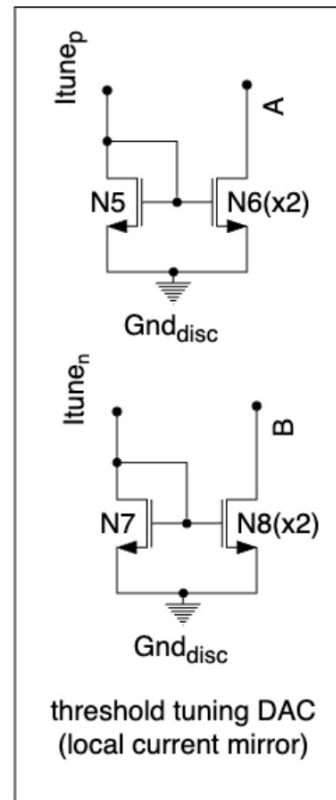
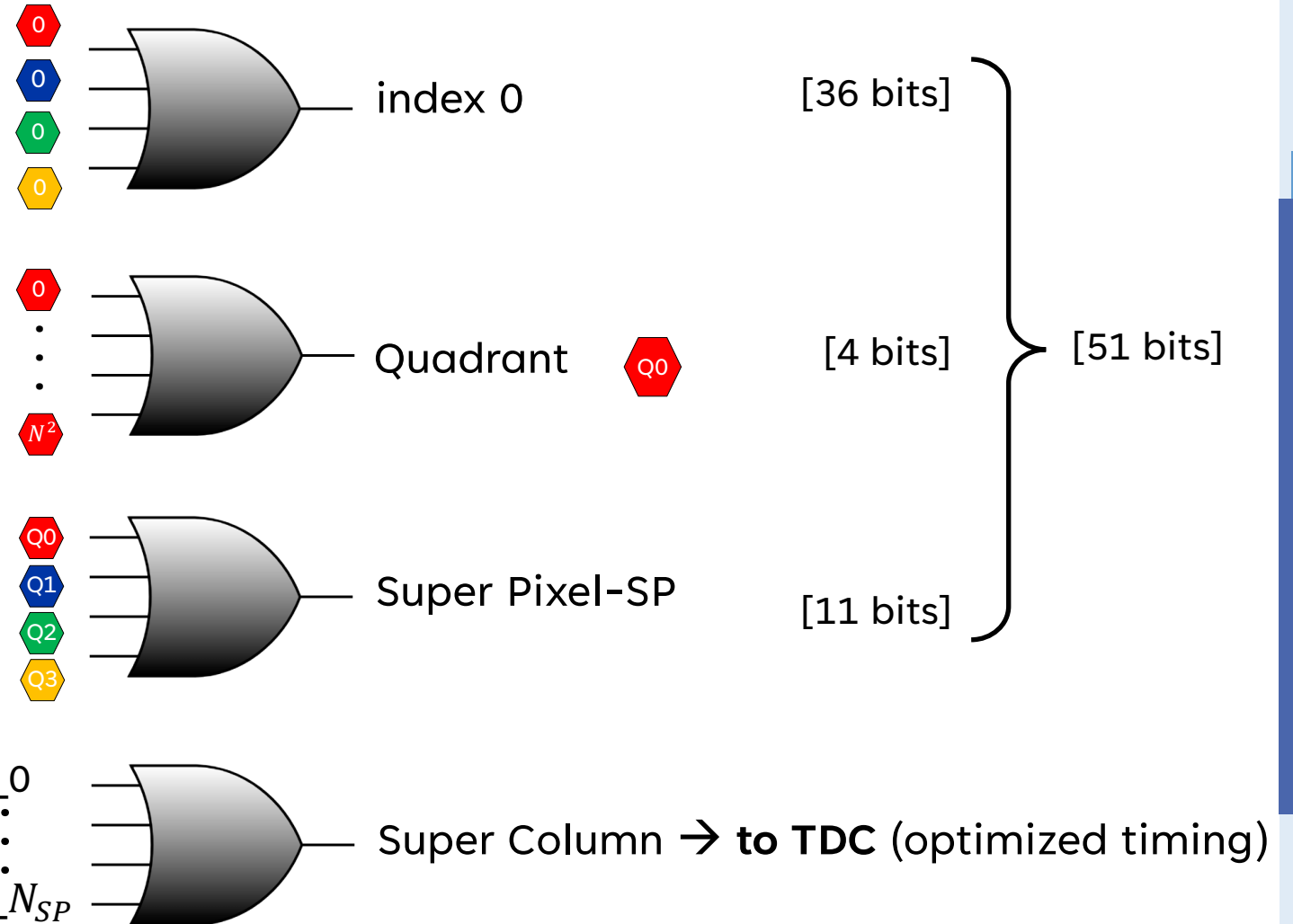
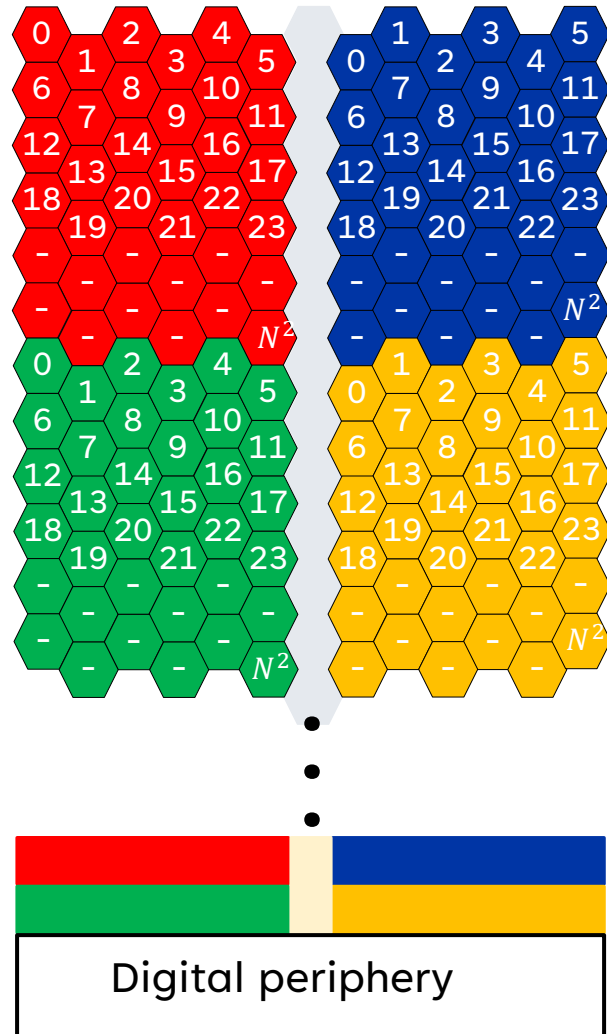
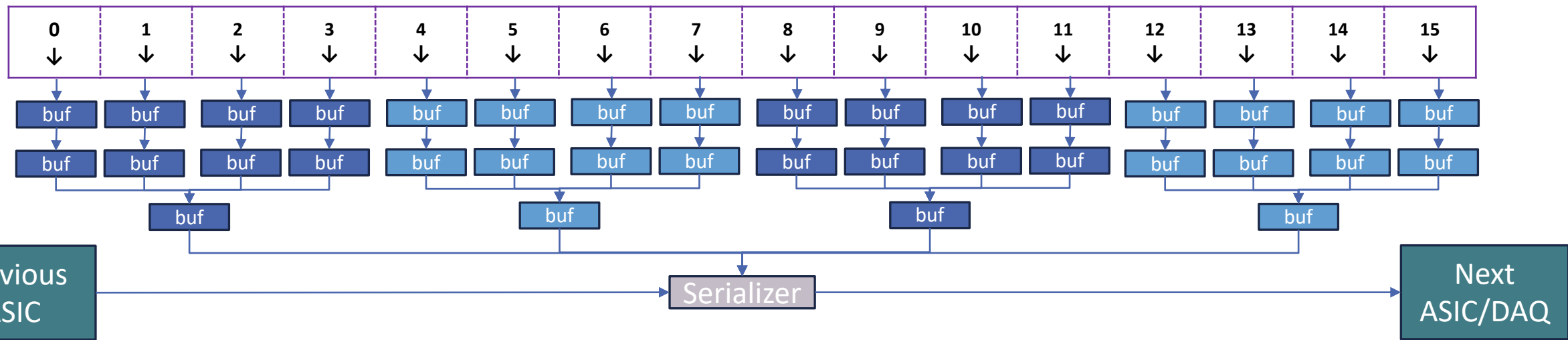


Figure 72 – Pre-amplifier schematic.





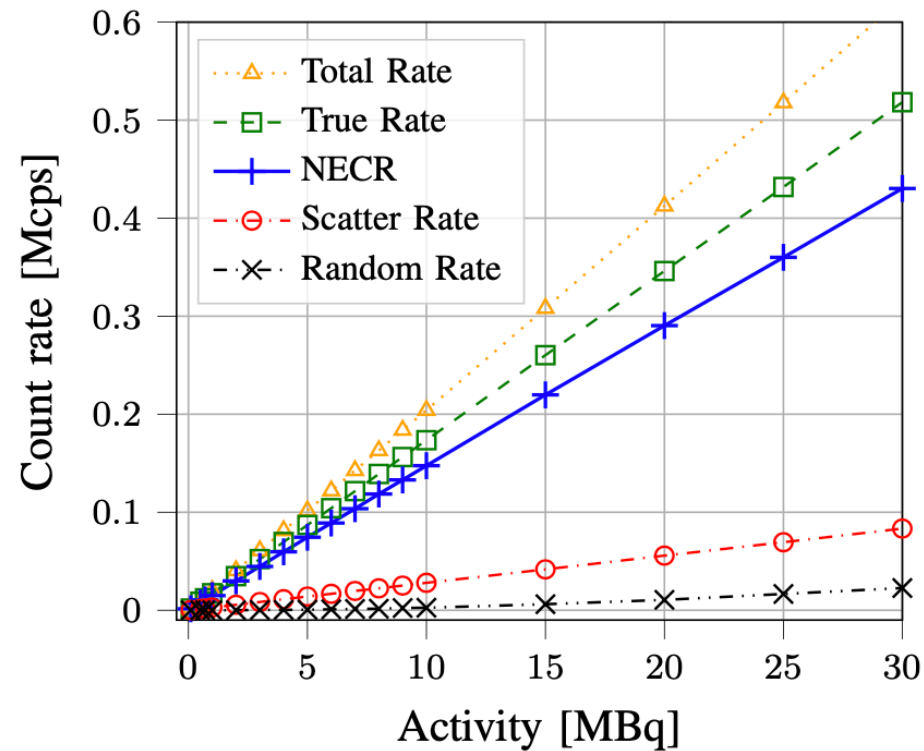


Each super-column provides pixel address and TDC data:
event size of **143 bits**.

Each super-column data is processed **in parallel** and **independently** in the periphery
3-level buffer minimizes data loss at the target event rate.

Read-out speed 50 Mb/s (with **50 MHz clock**)

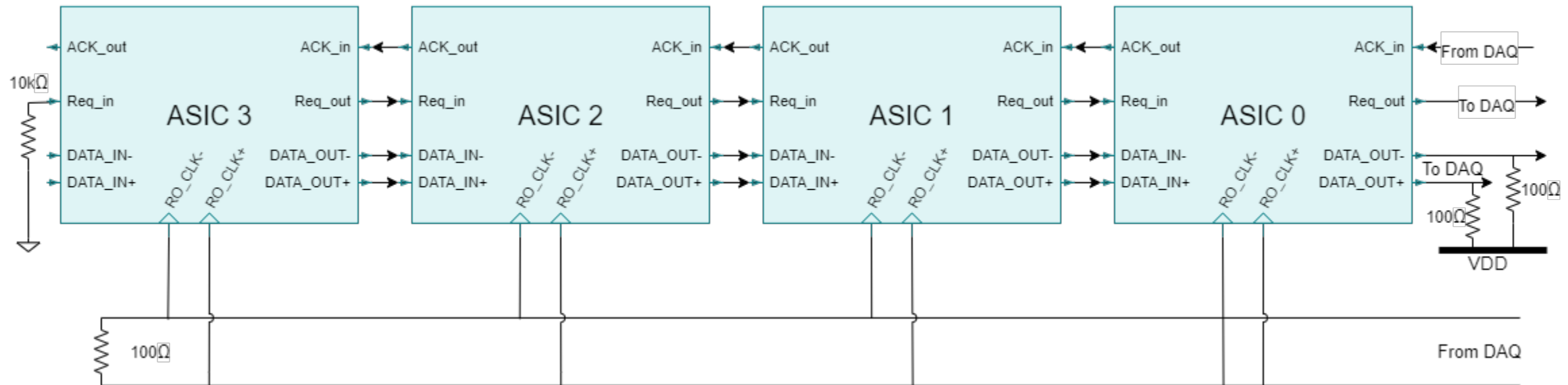
- Max read-out event rate 350 k events/s,
- Chips can be daisy-chained to reduce the number of outputs to DAQ.
- Maximum expected rate for quad module ($24 \text{ cm}^2 \times 10 \text{ kHz/cm}^2$) = 240 k events/s



$$NECR = \frac{T^2}{S+T+KR}$$

Fig. 17. Count rate as a function of activity with a window of 1 ns. (The timing resolution is of 0.3 ns.)



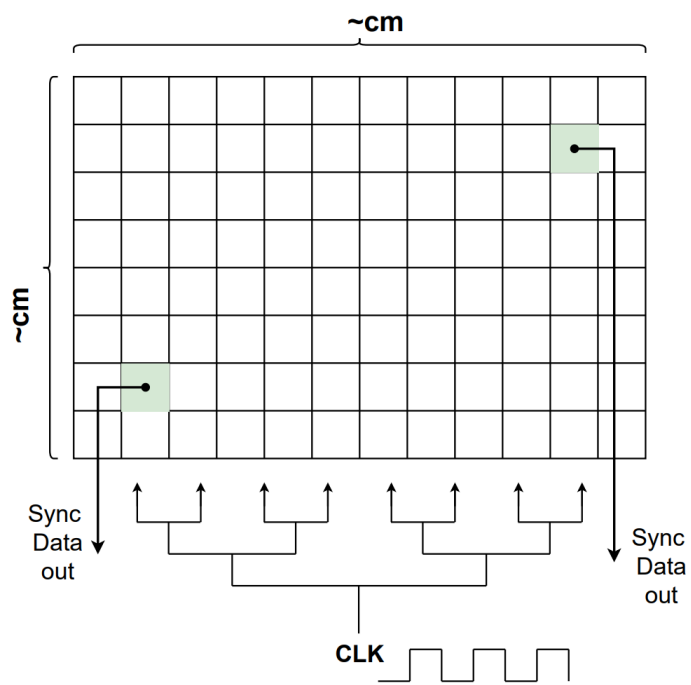


- 4-chip daisy chain
- Priority given to the left-most chip (furthest from DAQ)
- First chip request to GND
- DAQ acts as last unit in the chain.

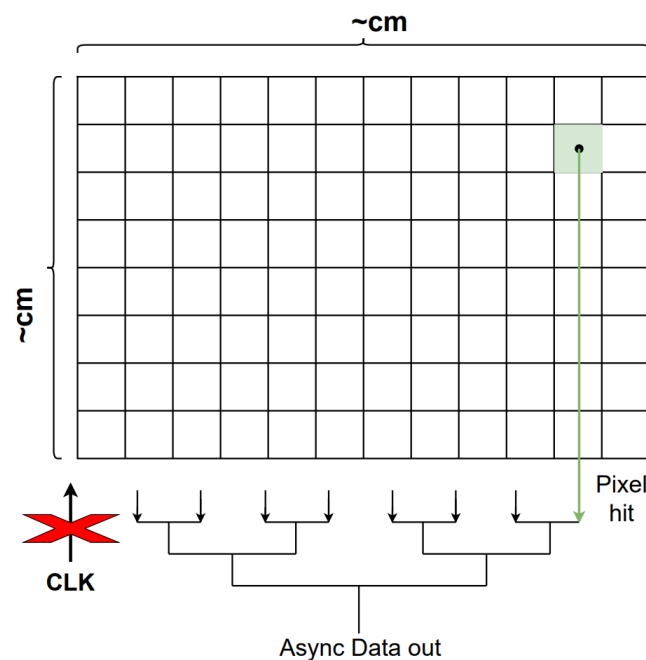
- In pixel sensor applications, especially for **large area pixel sensors** with a density of thousands of pixels, one of the challenges is an efficient information readout that meets the recent trends of stringent timing specifications.

→ Two different approaches:

Synchronous



Asynchronous



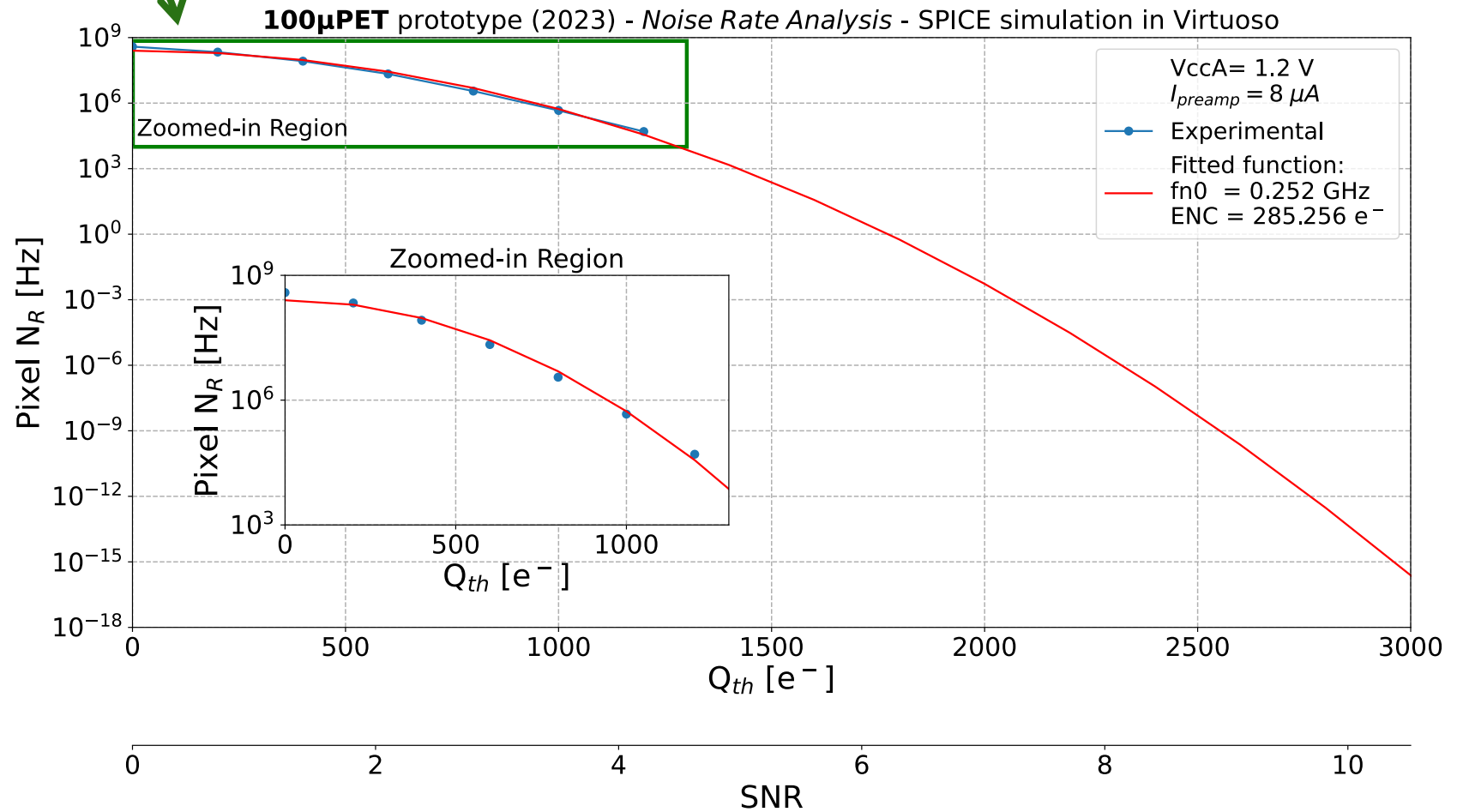
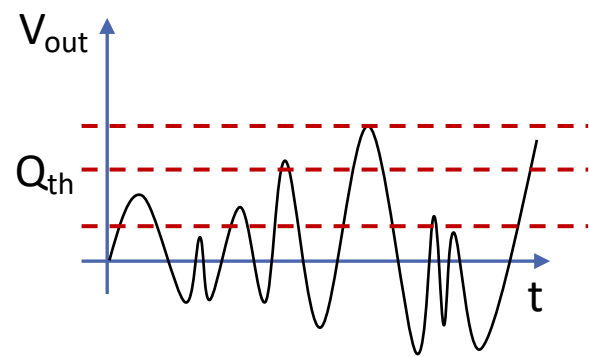
Synchronous

- Clock distributed across the chip
 - High power dissipation
 - Noise on the power supply
 - Stringent skew constraints on the clock tree

Asynchronous

- No clock distributed across the chip
 - Lower power dissipation
 - Less noise on the power supply
 - Stringent skew constraints on the outputs

$$N_R = f_{n0} \cdot e^{\frac{-Q_{th}^2}{2 \cdot ENC^2}}$$



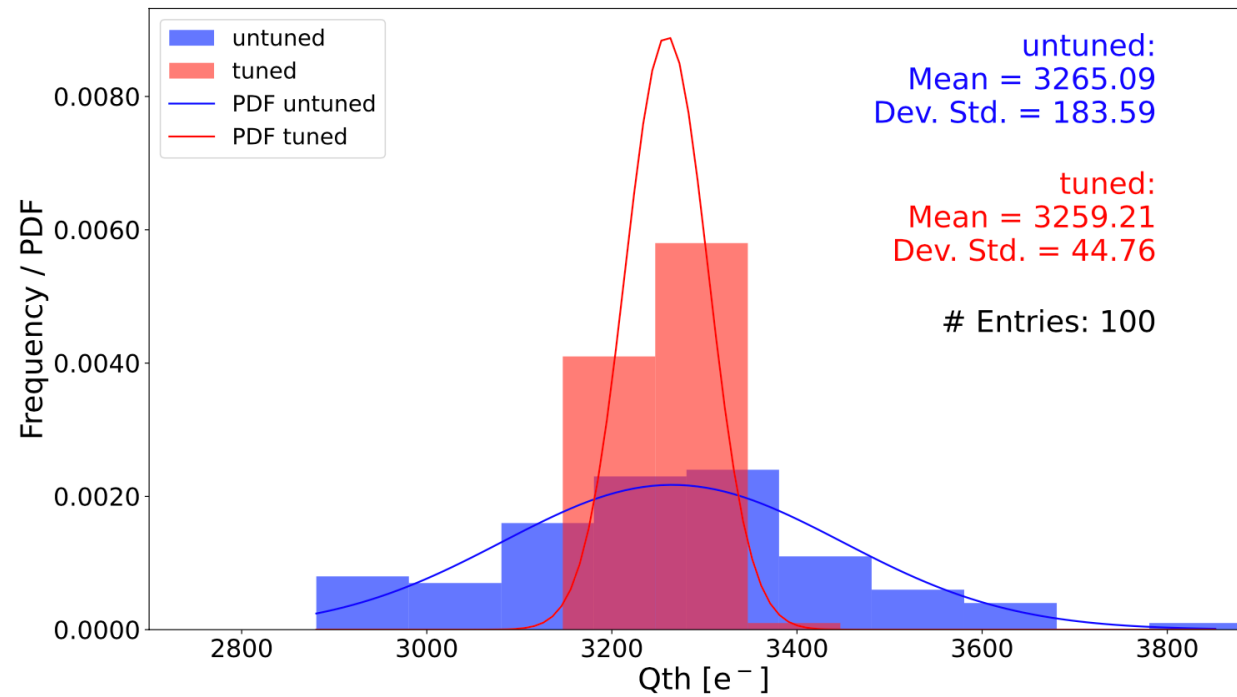


Figure 74 – Simulation - Mismatch compensation technique: frequency and the probability density function (PDF) of charge threshold mismatch of both untuned and tuned configurations. (Qth=3000 e⁻, # montecarlo = 100)