

# signal processing in cryogenic experiments

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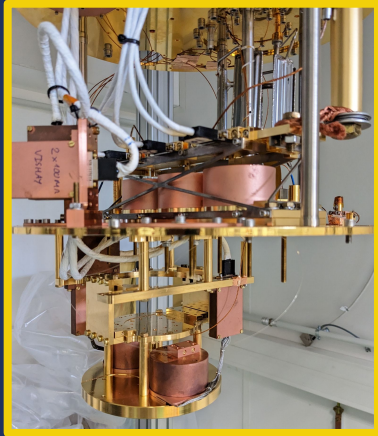
# What is “a signal” ?

“everything can be a signal”

Signal and noise distinction is arbitrary, the signal is the information we care about, the noise is the sum of all other perturbations.

“a signal is what is not a noise”

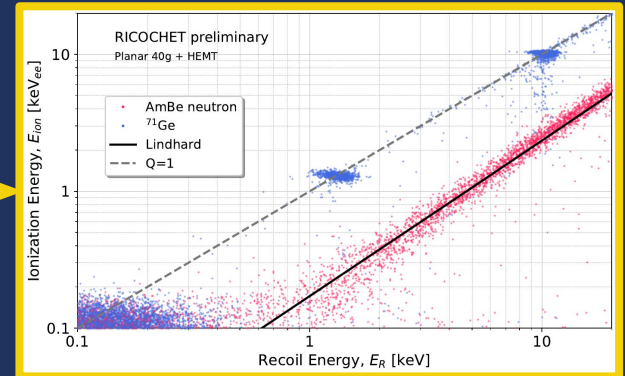
# A little bit of context



detector in a fridge



electronics / DAQ

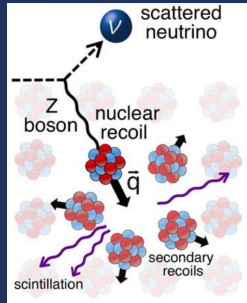


discoveries (at best)  
fun stuff (at least)

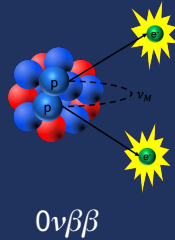
# Signal processing **extracting information**

## science/experiment interplay

the science determines the detection process  
the experiment adds noises

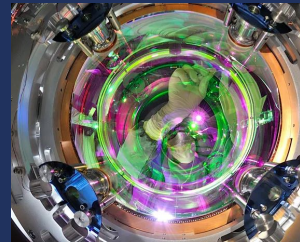


CEνNS

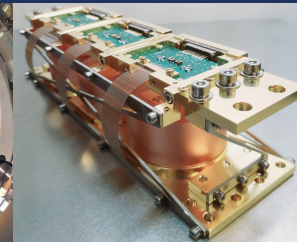


## general methods

same algorithms  
different experiments



LIGO, GW



RICOCHET, CEνNS

# Outlook of this lecture

## Basics (15 min)

basic knowledge  
useful definitions  
tips

## Case study (25 min)

~~bolometer~~ calorimeter experiment processing pipeline  
(RICOCHET like)

# Basics of spectral analysis

Fourier transform

Windowing

Convolution and correlation

Power Spectral Density estimation

# Fourier transform definition

The diagram shows the Discrete Fourier Transform equation:  $X_k = \sum_{n=0}^{N-1} x_n \cdot e^{-i2\pi \frac{k}{N} n}$ . The equation is enclosed in a yellow rectangular box. Arrows point from labels to specific parts of the equation: 'frequency signal' points to  $X_k$ ; 'time signal' points to  $x_n$ ; 'size of signal' points to  $N$ ; 'frequency index' points to  $k$ ; and 'time index' points to  $n$ .

$$X_k = \sum_{n=0}^{N-1} x_n \cdot e^{-i2\pi \frac{k}{N} n}$$

frequency signal      time signal      size of signal

frequency index      time index

# Fourier transform cheat sheet

## Parseval's theorem

$$\sum_{n=0}^{N-1} x_n y_n^* = \frac{1}{N} \sum_{k=0}^{N-1} X_k Y_k^*$$



## Plancherel's theorem

$$\sum_{n=0}^{N-1} |x_n|^2 = \frac{1}{N} \sum_{k=0}^{N-1} |X_k|^2$$

signal energy is **conserved** through FT

## Convolution theorem

$$\{u * v\}(x) = \mathcal{F}^{-1}\{U \cdot V\}$$

Dirac  $\delta$  is the **neutral element** of convolution

## Correlation

$$\mathcal{F}\{f \star g\} = \overline{\mathcal{F}\{f\}} \cdot \mathcal{F}\{g\}$$

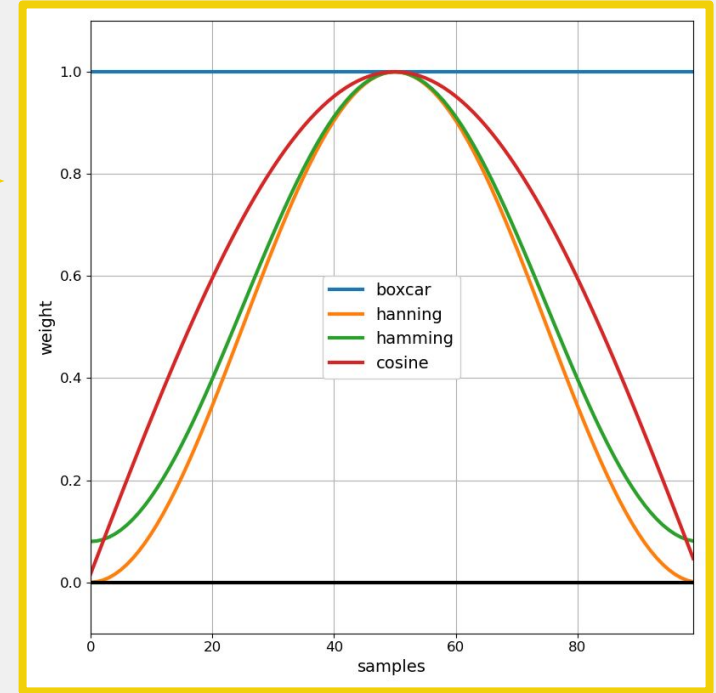
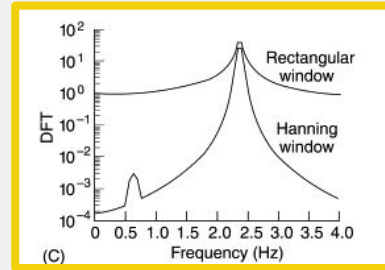
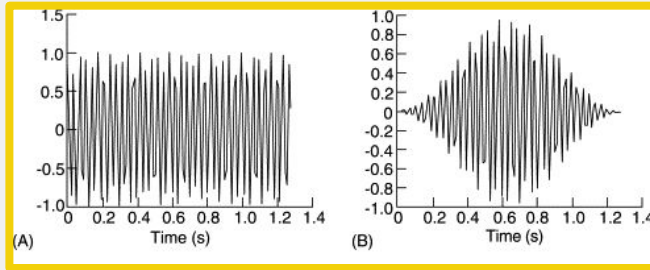
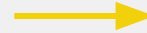
**measure of similarity** between f and g



# Fourier transform **windowing**

windowing forces the signal to be null at the borders of **observation window** → **reduce frequency leakage**

there is a lot of windowing functions  
but there is **no universal rule** to choose one



[Structural Dynamics Fundamentals and Advanced Applications, 2020](https://www.sciencedirect.com/topics/engineering/hanning-window)

<https://www.sciencedirect.com/topics/engineering/hanning-window>

# Spectral estimation definition

## Wiener-Khinchin

$$S(\omega) = \frac{1}{2\pi} \sum_{k=-\infty}^{\infty} r_{xx}(k) e^{-i\omega k}$$

## Autocorrelation

$$r_{xx}(\tau) = \mathbb{E}[x(t)^* \cdot x(t - \tau)]$$

**direct link** between statistical description and system evolution

one can define the **cross power spectral density** as the FT of the **cross-correlation** of two signal x, y

$$S_{xy}(f) = \sum_{n=-\infty}^{\infty} R_{xy}(\tau_n) e^{-i2\pi f \tau_n} \Delta\tau$$

in signal processing we often use the **cross-covariance**

$$R_{XY}(\tau) = \mathbb{E}[X_{t-\tau} \overline{Y_t}]$$

$$K_{XY}(\tau) = \mathbb{E}[(X_{t-\tau} - \mu_X) \overline{(Y_t - \mu_Y)}]$$

# Spectral estimation **coherency**

**Coherency definition**

Power Spectral Density between x and y

$$C_{xy}(f) = \frac{|G_{xy}(f)|^2}{G_{xx}(f)G_{yy}(f)}$$

Coherency estimate is **valid only for linear systems**  
results can be inaccurate in case of a non-linear dynamic

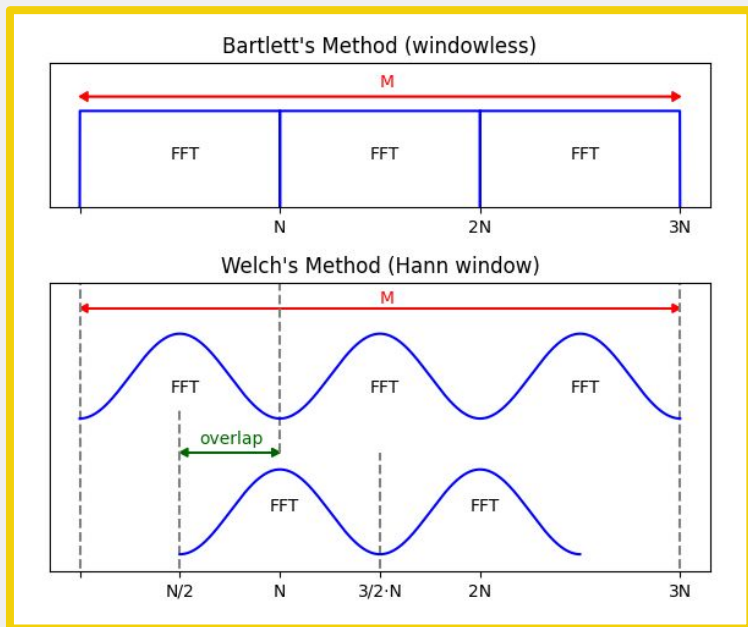
Power Spectral Densities of x and y

# Spectral estimation a lot of methods...

- **Non-parametric methods** for which the signal samples can be **unevenly spaced in time**
- [Least-squares spectral analysis](#), [Lomb–Scargle periodogram](#), [Non-uniform discrete Fourier transform](#)
- **Non-parametric methods** for which the signal samples must be **evenly spaced in time**
  - [Periodogram](#), [Bartlett's method](#), [Welch's method](#) a windowed version of Bartlett's method that uses overlapping segments
  - [Multitaper](#) is a periodogram-based method that uses multiple tapers
  - [Singular spectrum analysis](#)
  - [Short-time Fourier transform](#)
  - [Critical filter](#) is a nonparametric method based on [information field theory](#) that can deal with noise, incomplete data, and instrumental response functions
- **Parametric techniques**
  - [Autoregressive model](#) (AR) estimation, [Moving-average model](#) (MA) estimation, [Autoregressive moving-average](#) (ARMA)
  - [MUltiple Signal Classification](#) (MUSIC) is a popular [superresolution](#) method.
  - [Maximum entropy spectral estimation](#) is an *all-poles* method useful for SDE when singular spectral features are expected.
- **Semi-parametric techniques**
  - Sparse Iterative Covariance-based Estimation (SPICE) estimation, and the more generalized  $(r, q)$ -SPICE.
  - Iterative Adaptive Approach (IAA) estimation
  - [Lasso](#), similar to [least-squares spectral analysis](#) but with a sparsity enforcing penalty.

# Spectral estimation tips

## The Bartlett/Welch methods



IEEE Access. PP. 1-1. 10.1109/ACCESS.2021.3058744.

## Median / mean

**median** is **less sensitive to outliers** than mean

**more robust** than standard Welch method

need to correct the **median bias**

$$\alpha = \sum_{\ell=1}^n \frac{(-1)^{\ell+1}}{\ell}$$

**median** can be used to estimate  $\sigma$  (RMS)

$$\sigma = k \cdot \text{MAD}$$

MAD = median absolute deviation

$$k \sim 1.4826$$

**In Python:**

```
scipy.signal.welch(..., average="median"),  
scipy.stats.median_abs_deviation(..., scale="normal")
```

# Case study → multi-channel calorimeter



**Goal 1:** electronic noise estimation ( including correlations )

**Goal 2:** automatic pulse detection and amplitude estimation

**Constraints** lowest energy possible + “noisy” environment

Colas, J., Billard, J., Ferriol, S. *et al.*

**Development of Data Processing and Analysis Pipeline for the Ricochet Experiment.**

*J Low Temp Phys* 211, 310–319 (2023).

<https://doi.org/10.1007/s10909-022-02907-5>

# Degradation model

describes how the **detector measurement line** impacts the **signal of interest**  
we are looking for “u”, but we measure “x”

**additive noise**

$$X = U + B$$

today, we'll go with this one



**cross-talk**

$$X = U + H.Y$$

**linear transformation**

$$X = A . (U + B) + C \quad (\text{example})$$

# Processing pipeline



combiner

downsampler

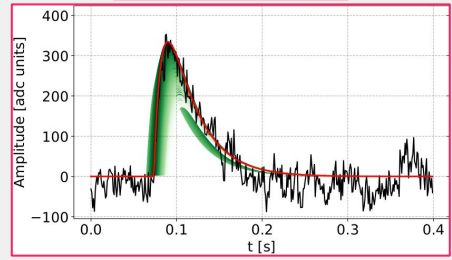
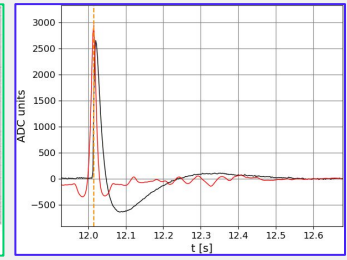
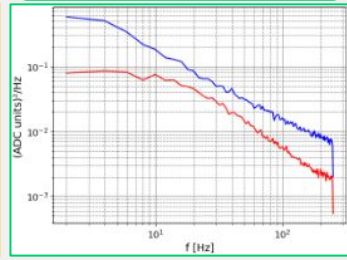
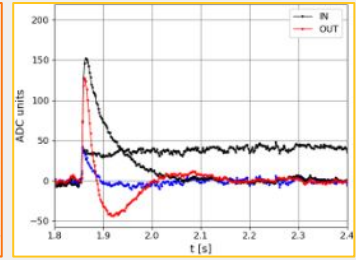
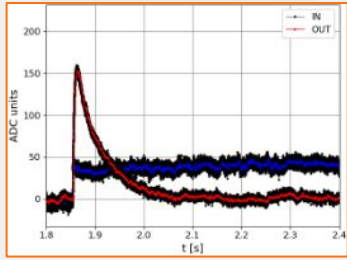
pre-filter

noise estimator

trigger

minimizer

RQ's



continuous data stream

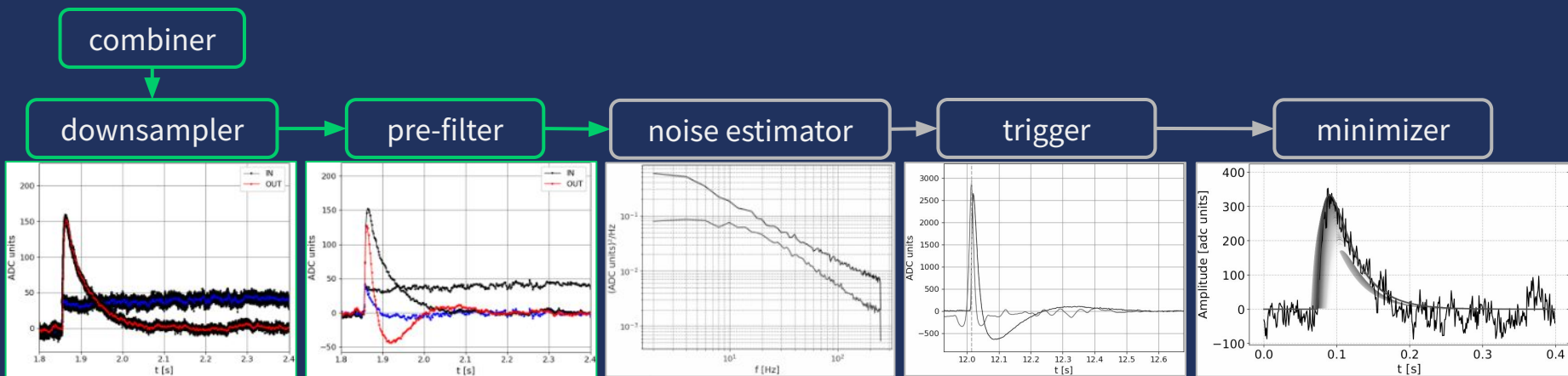


reduced quantities



# Pre-processing

Goal: prepare data for more reliable or faster analysis

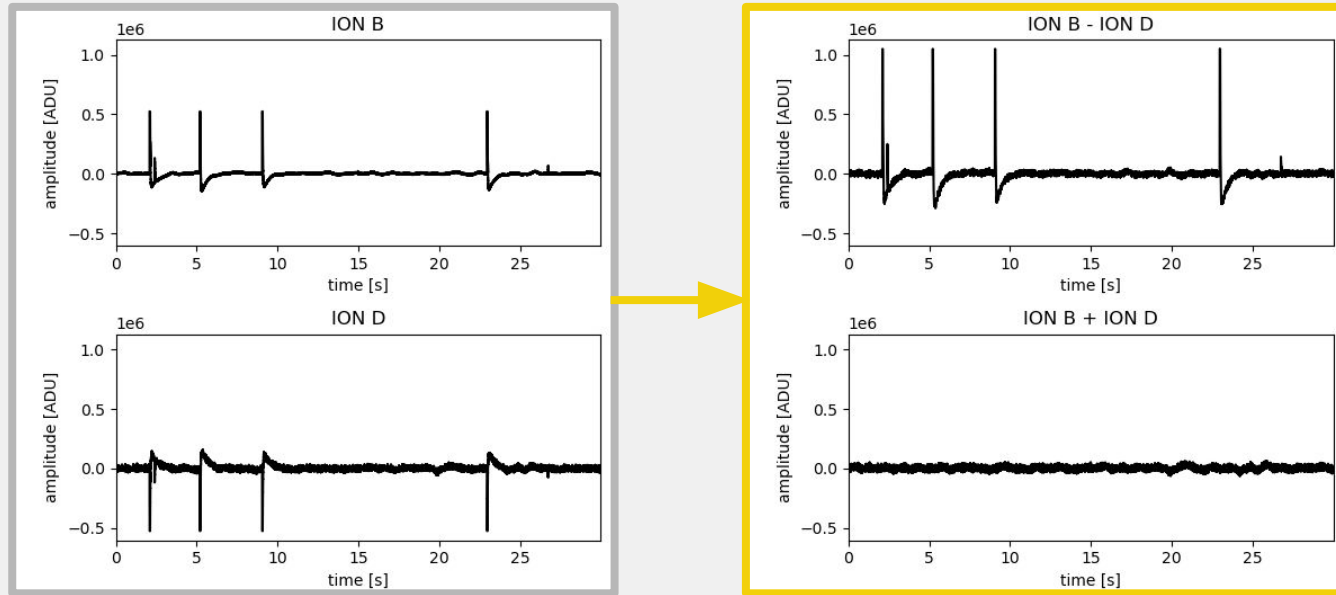


continuous data stream



“clean” data stream

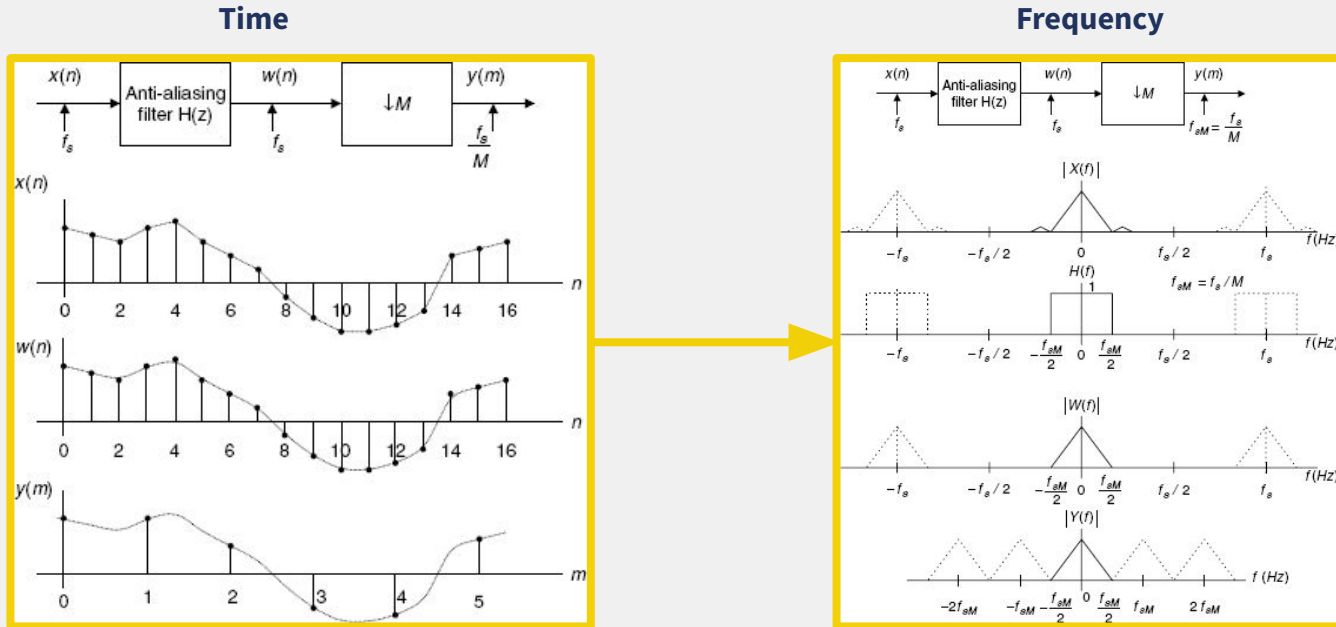
# Pre-processing **combine channels**



**Sometime, signal processing can be trivial**

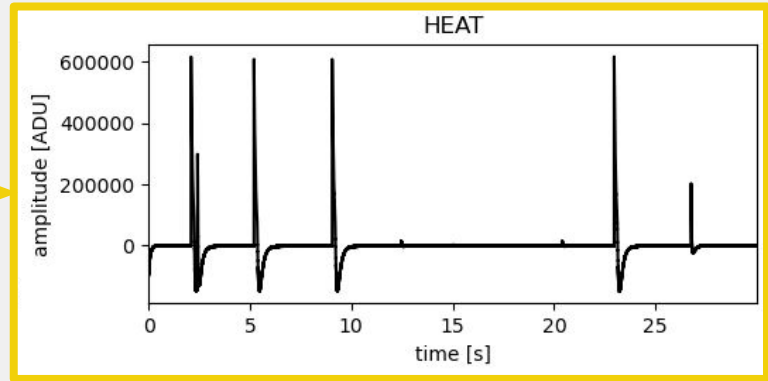
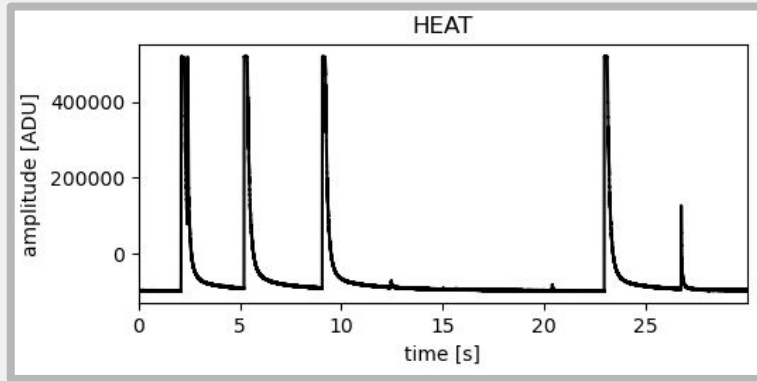
$e^-h^+$  collection  $\rightarrow$  differential signal  $\rightarrow$  summing them will improve SNR by  $\sqrt{2}$

# Pre-processing downsampling



**Downsampling reduces the data size**, thus improves the processing speed and (may) degrade performances reducing the number of sample will induces **aliasing** → you need to filter these aliases before decimation

# Pre-processing high-pass filtering



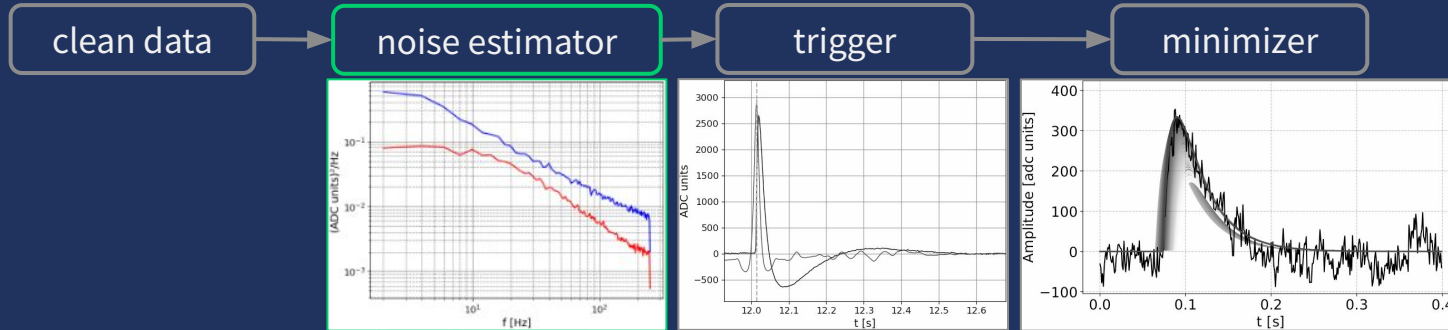
It's necessary to define an observation window of size **tw**, so it's not possible to (**accurately**) observe frequencies below  $2/tw \rightarrow$  **remove them !**

Ex: If **tw** = 1 s, **fc** =  $2/tw$  = 2 Hz

# Noise estimation

**Goal**: estimate electronic noise floor PSD/CSD

**Constraints**: lot of pulses (wrt. det. dynamic), stationary noises (ergodicity), ...

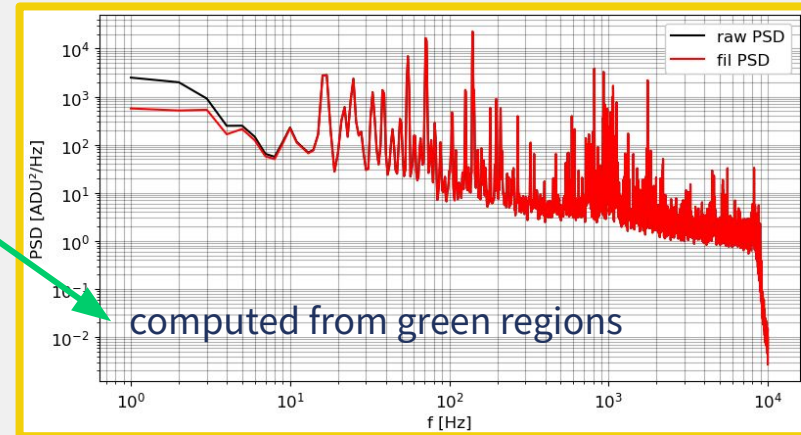
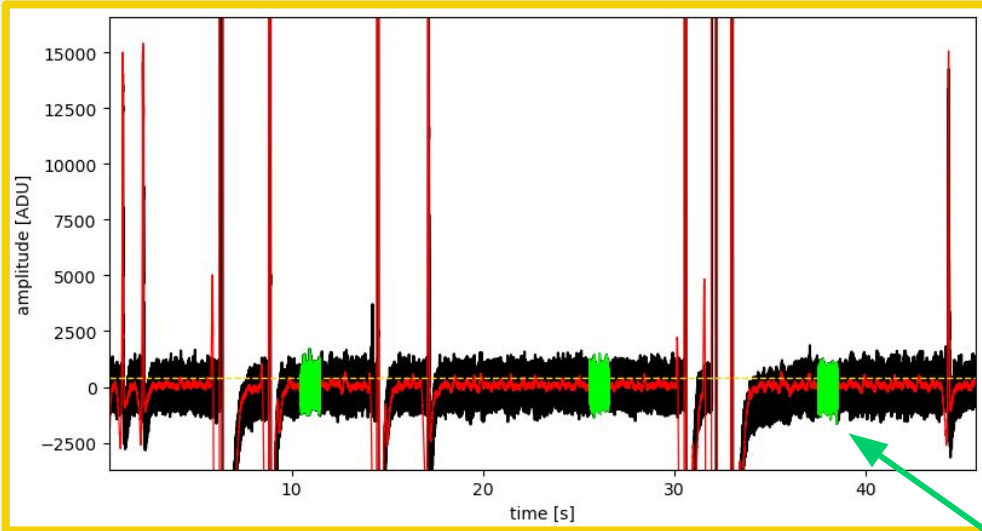


# Noise estimation

## Principle

we know the signal we are looking for: **pulses**

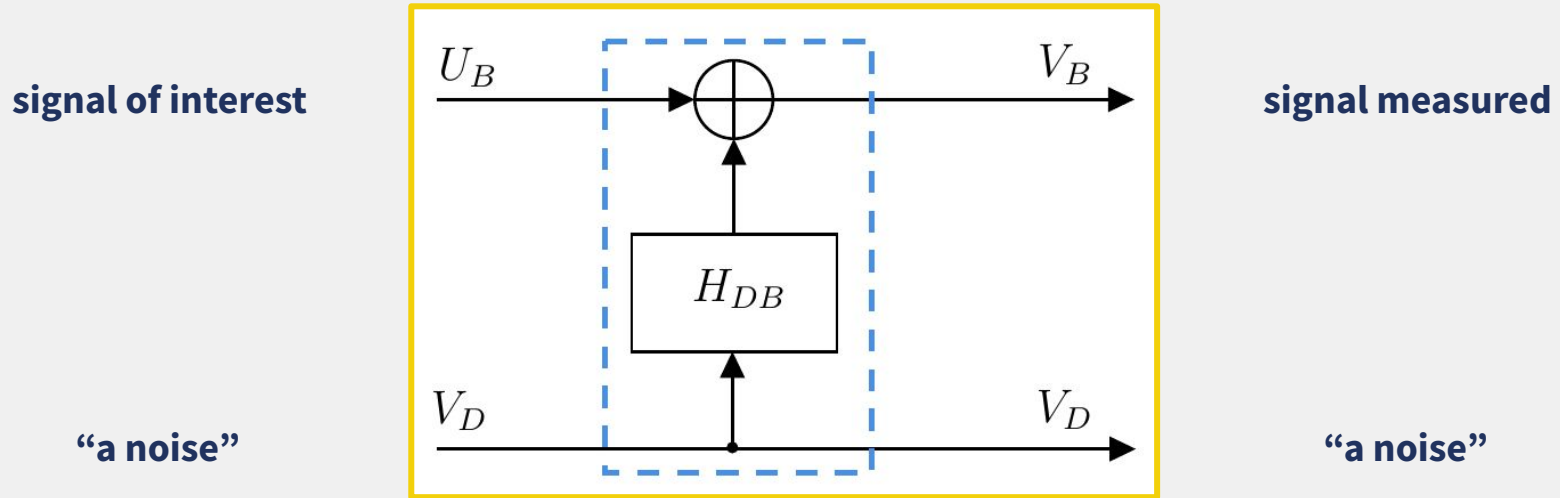
we can use a **similarity** measurement to identify the sections of data which are “**as different as possible of a signal**”



**Note:** for a more precise description of this algorithm see my Phd thesis

# Exercise 1: noise decorrelation

We can use many approaches to clean our data, let's talk about noise decorrelation



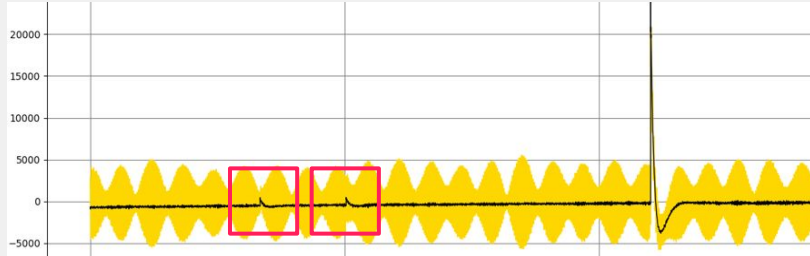
**Question 1:** find the coupling transfer function that minimize the energy of decorrelated signal

**Question 2:** determine the relation between the PSD of  $U_B$  and  $V_B$

# Exercise 1: noise decorrelation (results)

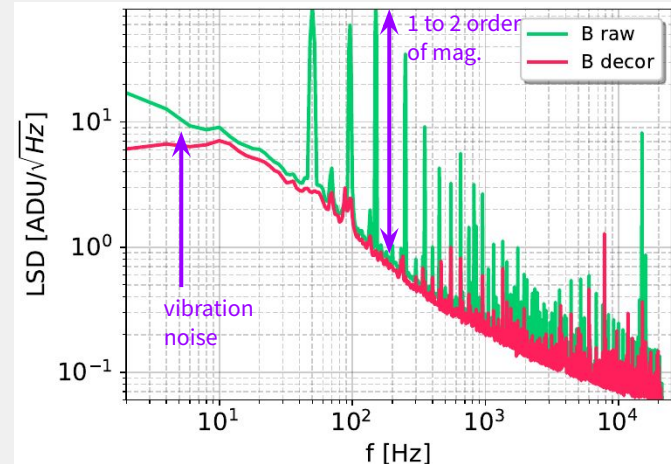
## Question 1

$$H_{DB}(f) = \frac{\Re(V_B^*(f)V_D(f))}{|V_D(f)|^2}$$



## Question 2

$$\text{PSD}(U_B) = (\mathbf{1} - \mathbf{C}_{BD}) \cdot \text{PSD}(V_B)$$

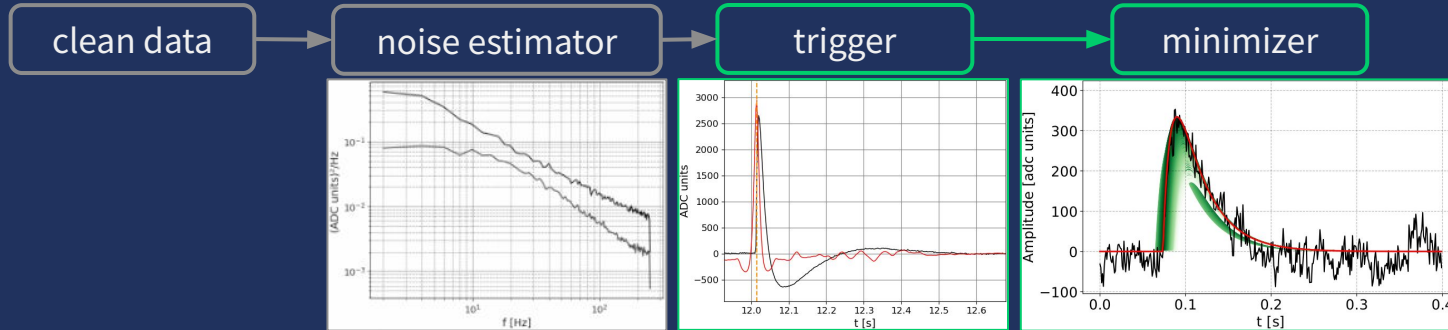




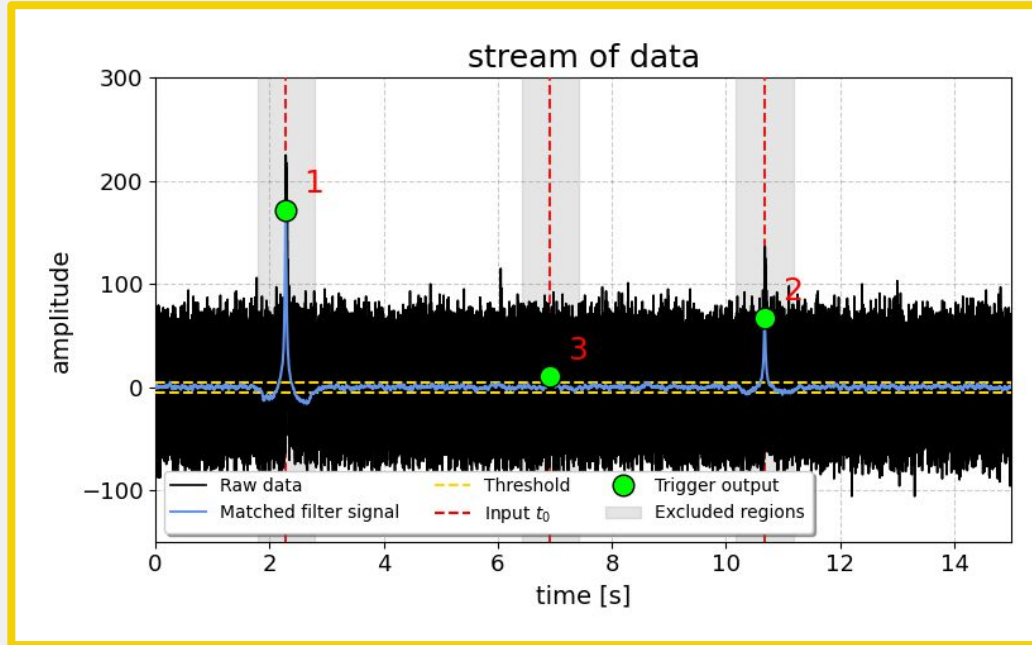
# Automatic pulse detection

**Goal 1:** detect pulses at the lowest energy possible

**Goal 2:** estimate the amplitude of each individual pulse



# Automatic pulse detection



matched filter

$$H = c \frac{P^*}{J}$$

normalization factor

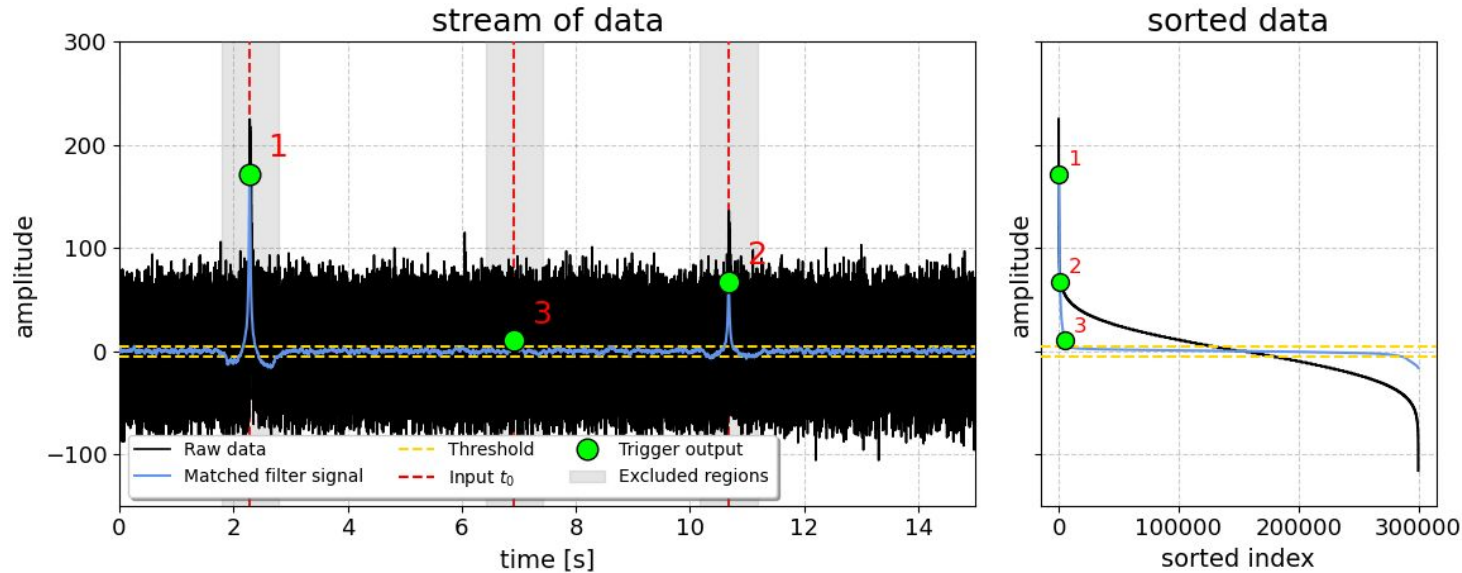
pulse template

electronic noise psd

the matched filter **optimizes the SNR**

it implies a **lower detection threshold**  
→ useful for low energy physics

# Automatic pulse detection



1. sort by amp.
2. take first sample
3. exclude in time
4. take next
5. go back to 3

# Exercise 2: amplitude estimation

We identified the approximate position of pulses → **what are their amplitudes ?**  
we want an **objective function to minimize** →  $\chi^2$

**Question 1: how to estimate the amplitude “a” from  $\chi^2$  expression ?**

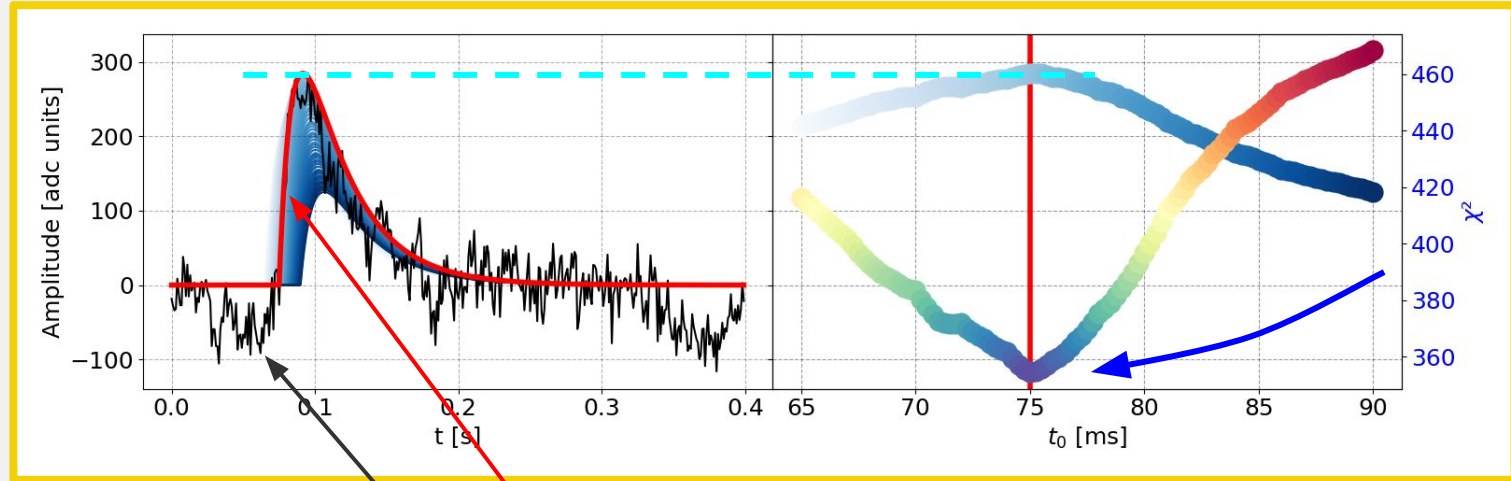
$$\chi^2 = \sum_f \frac{|D - a \cdot M|^2}{J}$$

This expression suppose a gaussian variance likelihood

in short, minimizing  $\chi^2$  is equivalent to maximizing the likelihood

# Exercise 2: amplitude estimation

## Frequential pulse fitting illustration



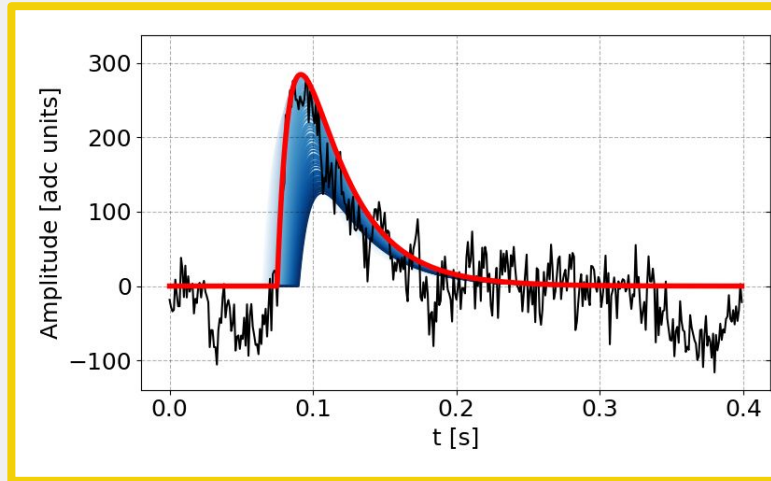
$$\chi^2 = \sum_f \frac{|D - a \cdot M|^2}{J}$$

$\chi^2$  can be seen as a measure of **goodness of fit**

$\min \chi^2 \leftrightarrow \max \hat{a}$

# Energy resolution

What is the energy resolution ?

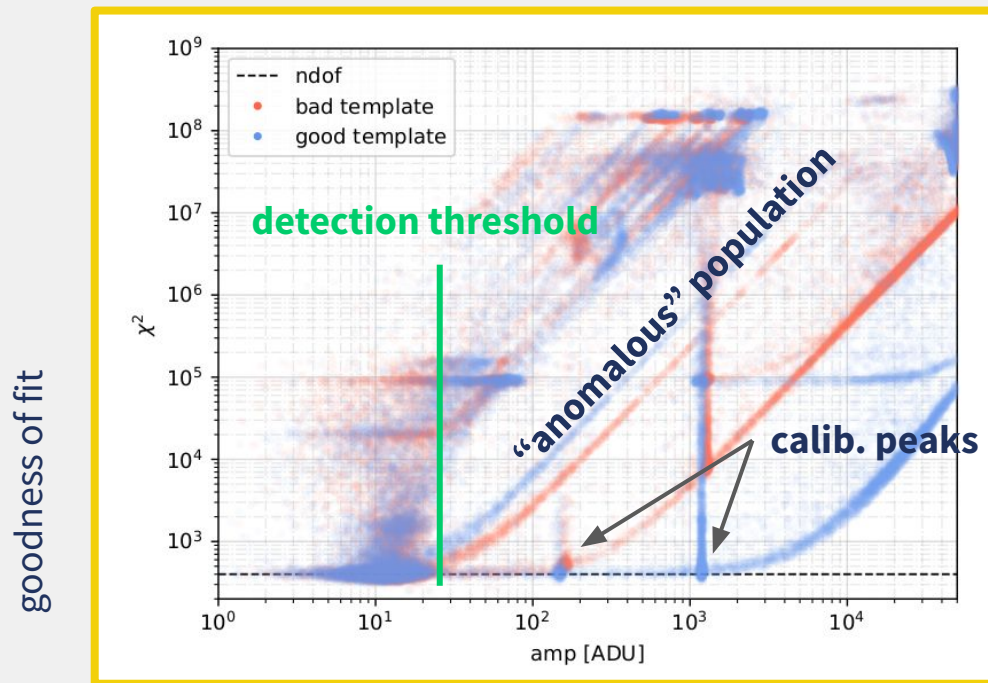


Formula

$$\sigma^2 = \left( \frac{1}{2} \frac{\partial^2 \chi^2}{\partial \hat{a}^2} \right)^{-1}$$

The timing resolution can also be derived from  $\chi^2$

# What we get at the end ...



amplitude

# Conclusion

- Signal processing (SP) **can be simple** (ex: addition of two signals)
- There is a bunch of **unused and interesting methods**
  - interact with SP experts ! (it's a very wide field)
- A cryogenic detector is a **complex system** which combine hardware and software
  - SP is a crucial step in the detection pipeline
- SP method to use depends on the **sensor conditioning**
  - detector design
  - amplifiers type

*“ Information is the resolution of uncertainty. ”* Claude Shannon



# Literature

## Non-linear methods:

“Improving the Performance of Cryogenic Calorimeters with Nonlinear Multivariate Noise Cancellation Algorithms”  
[arXiv:2311.01131v2 \[physics.ins-det\]](#) 6 Feb 2024

## Decorrelation methods:

“Automatic cross-talk removal from multi-channel data”  
[arXiv:gr-qc/9909083v1](#) 27 Sep 1999

“Noise correlation and decorrelation in arrays of bolometric detectors”  
[arXiv:1203.1782v1 \[physics.data-an\]](#) 8 Mar 2012

## Threshold reduction:

“Lowering the energy threshold of large-mass bolometric detectors”  
[arXiv:1012.1263v1 \[astro-ph.IM\]](#) 6 Dec 2010

“Processing the signals from solid-state detectors in elementary-particle physics.”  
*Riv. Nuovo Cim.* **9**, 1–146 (1986) <https://doi.org/10.1007/BF02822156>

## Parameter/resolution estimation:

“Optimum filter-based analysis for the characterization of a high-resolution magnetic microcalorimeter towards the DELight experiment”  
[arXiv:2310.08512v1 \[hep-ex\]](#) 12 Oct 2023

“When “Optimal Filtering” Isn’t”  
[arXiv:1611.07856v1 \[physics.data-an\]](#) 23 Nov 2016

+ variational minimization, adaptative filtering (active denoising), wavelets, non-gaussian noise hypothesis, machine learning, ...

# Likelihood vs. chi<sup>2</sup>

N pulses fitted simultaneously with beta\_n the “local” parameters (t0, a, ...) and theta the “global” shape parameter

$$\mathcal{L}(\vec{\theta} | D_0, \dots, D_N) = \prod_f \prod_{n=1}^N e^{-\frac{1}{2} \frac{|D_n(f) - \bar{D}_n(f)|^2}{J(f)}}$$

$$-\log \mathcal{L}(\vec{\theta}) = \frac{1}{2} \sum_{n=1}^N \chi^2(\vec{\theta}; \vec{\beta}_n)$$

$$\begin{aligned} \left. \frac{\partial \log(\mathcal{L}(\vec{\theta} | D_0, \dots, D_N))}{\partial \vec{\theta}} \right|_{\vec{\theta}=\hat{\vec{\theta}}} &= 0 \\ \equiv \sum_{n=1}^N \left. \frac{\partial \chi^2(\vec{\theta}; \vec{\beta}_n)}{\partial \vec{\theta}} \right|_{\vec{\theta}=\hat{\vec{\theta}}} &= 0 \end{aligned}$$

**minimizing** the chi<sup>2</sup> = **maximizing** the likelihood

# Generalization

mono-channel

$$\chi^2 = \sum_f \frac{|D - a \cdot M|^2}{J}$$

noise PSD

multi-channels

$$\chi^2(\vec{a}; t_0) = \sum_f [\vec{Y} - \vec{a} \vec{P}(\vec{\theta}; t_0)]^\dagger J^{-1} [\vec{Y} - \vec{a} \vec{P}(\vec{\theta}; t_0)]$$

cross-covariance  
matrix