



HGCROC: the front-end readout ASICs for the CMS High Granularity Calorimeter

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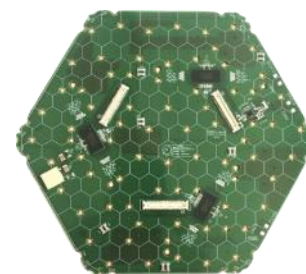
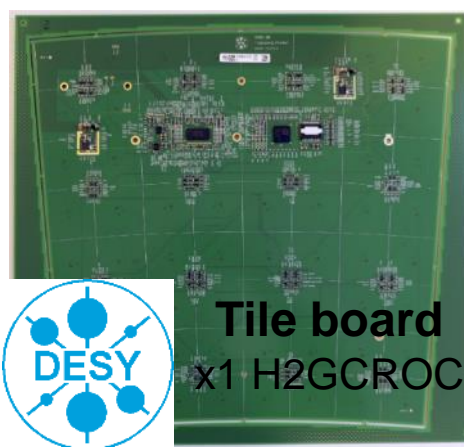
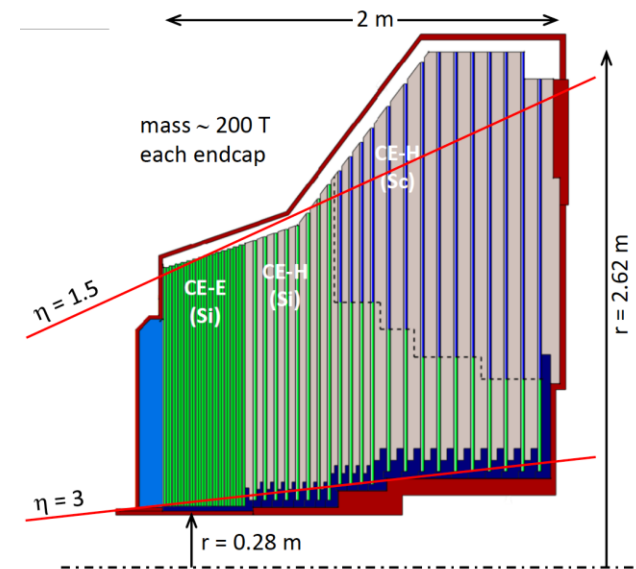
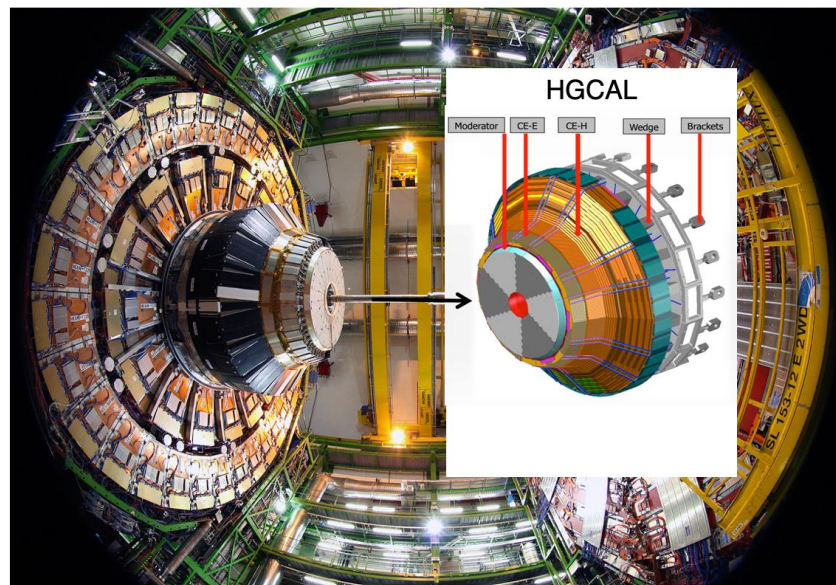
- ❑ HGCAL covers $1.5 < \eta < 3.0$
- ❑ Full system maintained at -30°C
 - ❑ $\sim 640 \text{ m}^2$ of silicon sensors, 6.1M Si channels, 0.5 or 1.1 cm^2 cell size
 - ❑ $\sim 370 \text{ m}^2$ of scintillators, 240k scint-tile channels
- ❑ Data readout from all layers
- ❑ Trigger readout from alternate layers in CE-E and all in CE-H

❑ Active Elements

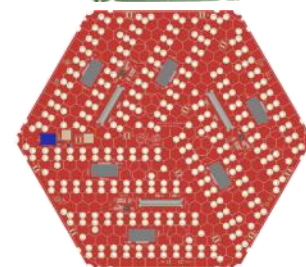
- ❑ Electromagnetic calorimeter (CE-E): Si, Cu/CuW/Pb absorbers
- ❑ Hadronic calorimeter (CE-H): Si & scintillator, steel absorber

❑ New Front-end electronics

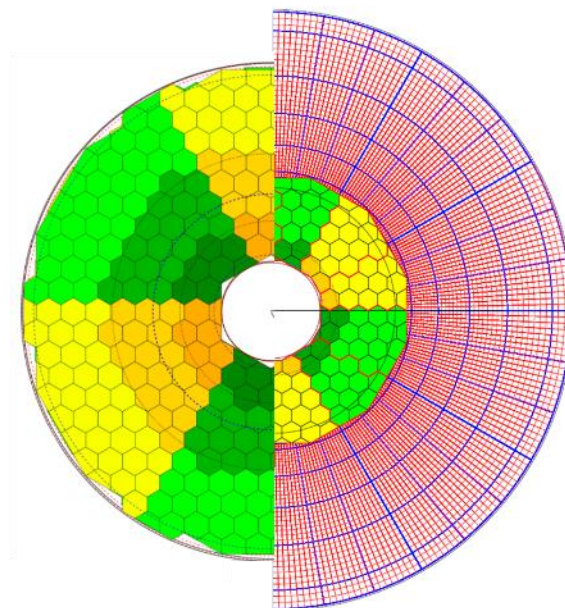
- ❑ Two versions: Silicon and SiPM
- ❑ Rad.tolerant (200 Mrad, $1.1016 \text{ neq} / \text{cm}^2$)
- ❑ Power consumption: 20 mW per channel
- ❑ Noise: 0.4 fC
- ❑ Charge: 0.2 fC to 10 pC
- ❑ Pileup mitigation: Fast shaping (peak $< 25 \text{ ns}$), precise timing capability (25 ps)



LD HB
x3 HGCROC



HD HB
x6 HGCROC



Overall chip divided in two symmetrical parts

- 1 half is made of:
 - 39 channels: 18 ch, CM0, Calib, CM1, 18 ch (78 channels in total)
 - Bandgap, voltage reference close to the edge
 - Bias, ADC reference, Master TDC in the middle
 - Main digital block and 3 differential outputs (2x Trigger, 1x Data)

Measurements

- Charge
 - ADC (AGH): peak measurement, 10 bits @ 40 MHz, dynamic range defined by preamplifier gain
 - TDC (IRFU): TOT (Time over Threshold), 12 bits (LSB = 50ps)
 - ADC: 0.4 fC resolution. TOT: 2.5 fC resolution
- Time
 - TDC (IRFU): TOA (Time of Arrival), 10 bits (LSB = 25ps)

Two data flows

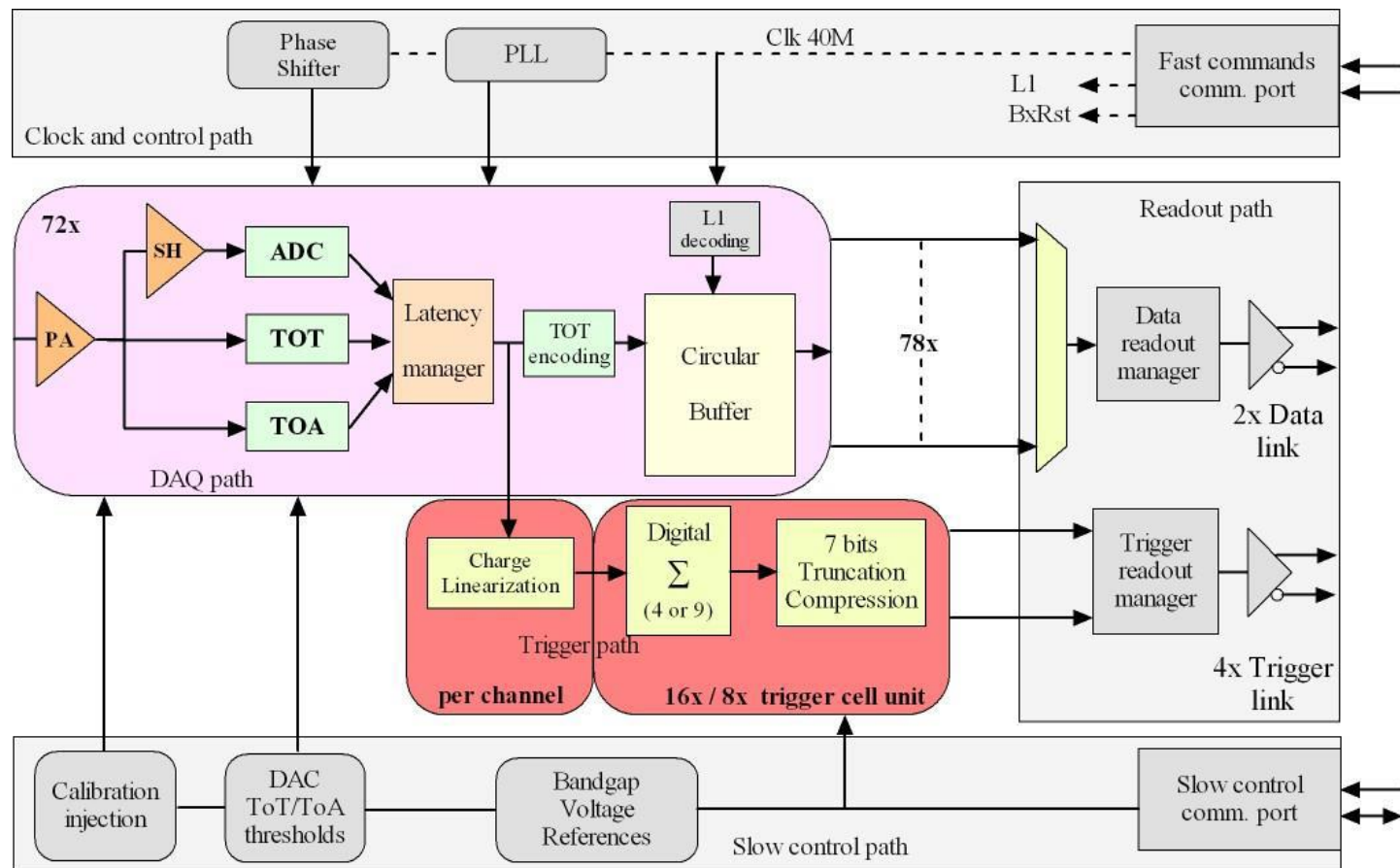
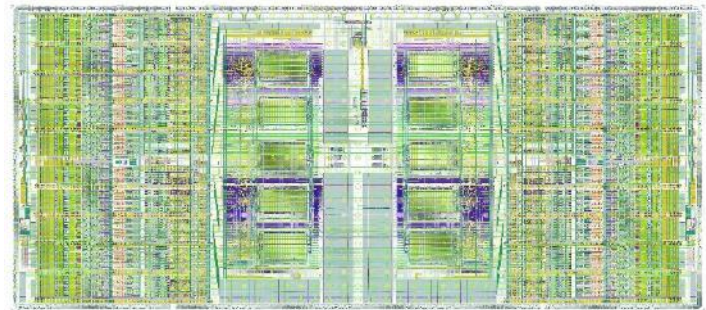
- DAQ path
 - 512 depth DRAM (CERN), circular buffer
 - Store the ADC, TOT and TOA data
 - 2 DAQ 1.28 Gbps links
- Trigger path
 - Sum of 4 (9) channels, linearization, compression over 7 bits
 - 4 Trigger 1.28 Gbps links

Control

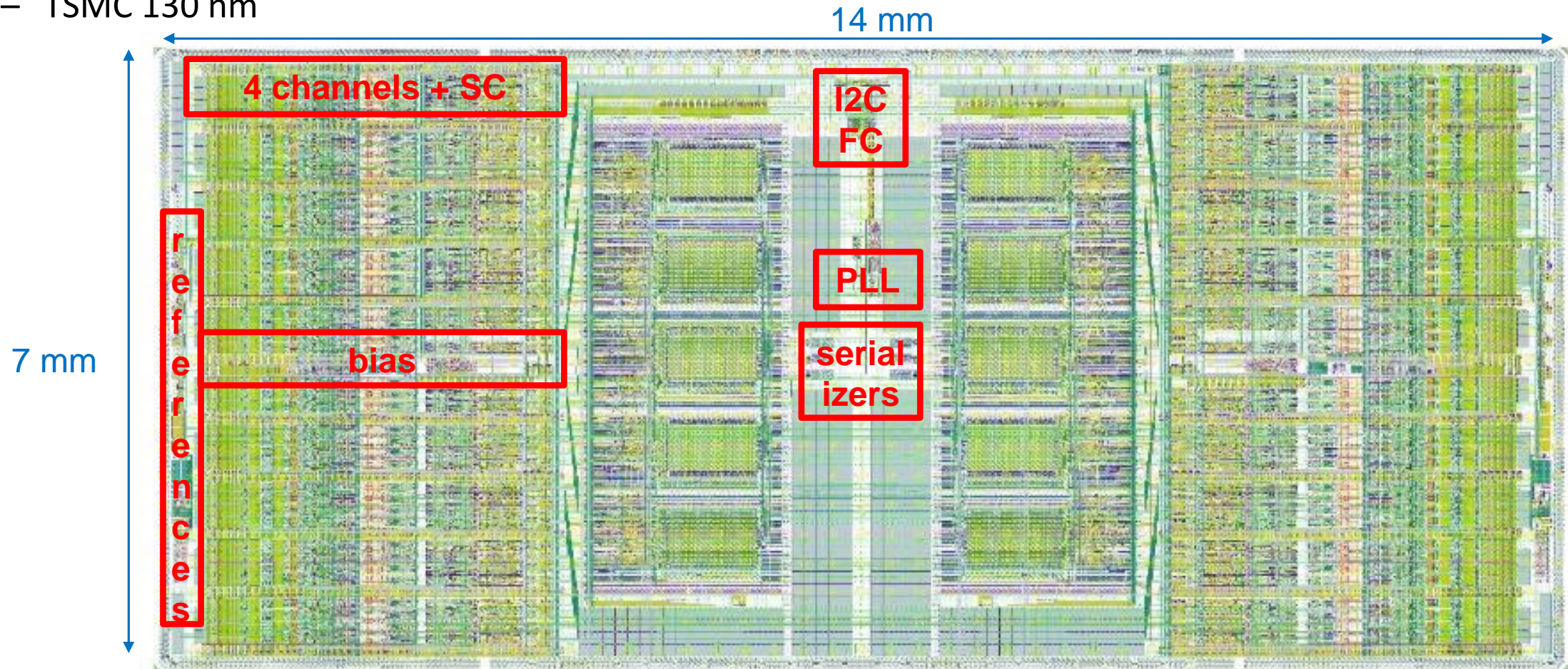
- Fast commands
 - 320 MHz clock and 320 MHz commands
 - A 40 MHz extracted, 5 implemented fast commands
- I2C protocol for slow control

Ancillary blocks

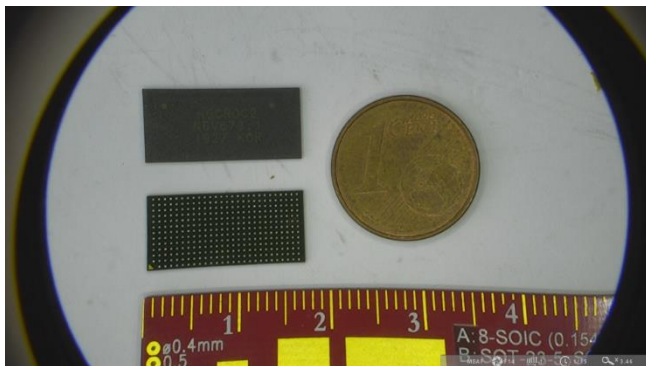
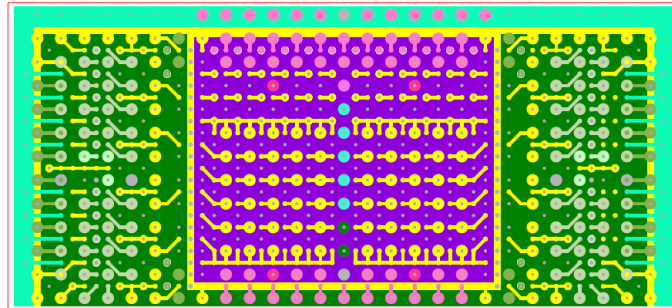
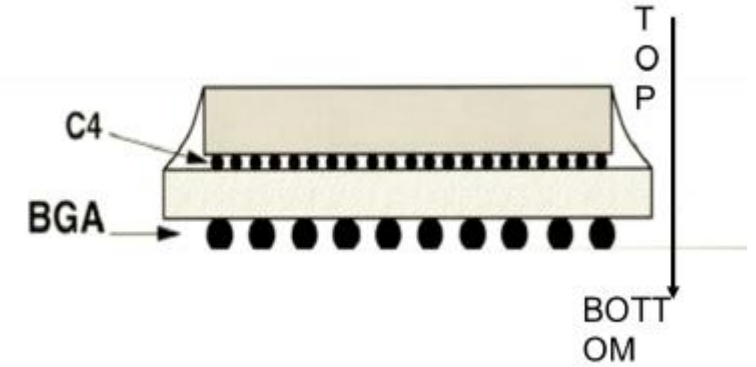
- Bandgap (CERN)
- 10-bits DAC for reference setting
- 11-bits Calibration DAC for characterization and calibration
- PLL (IRFU)
- Adjustable phase for mixed domain



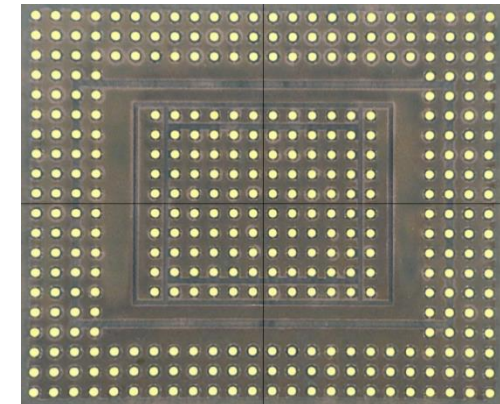
- Two half-chips + common digital blocks (I2C, fast commands, PLL, 1.28 Gb serializers)
 - Each half hosts 8 blocks of 4 channels + 2 common mode blocks
 - Digital data processing : trigger sums and (partial) data storage in DRAM
 - TSMC 130 nm



- C4 bump bonding
- 2 BGA packages fabricated:
 - High Density (0.6 mm pitch) for Hexaboard with ~400 channels (120 μm Si sensors)
 - Low Density (0.8 mm pitch) for Hexaboard with ~200 channels (200 μm and 300 μm Si sensors) and Tileboards with ~70 channels



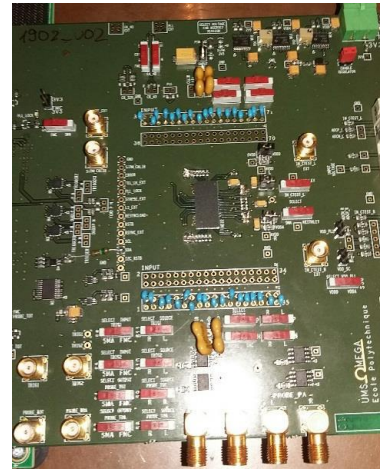
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C	ENC	ENC	ENC	ENC	ADD	SEL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL
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E	ENC	ENC	ENC	ENC	ADD	SEL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL
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G	ENC	ENC	ENC	ENC	ADD	SEL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL	PREL
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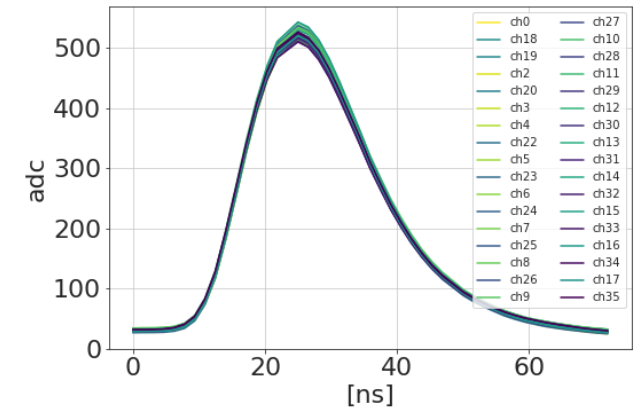
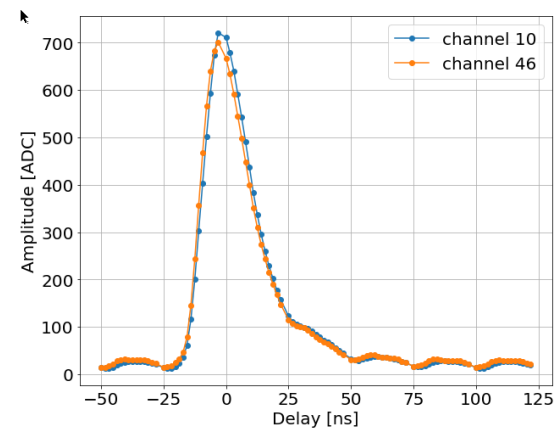
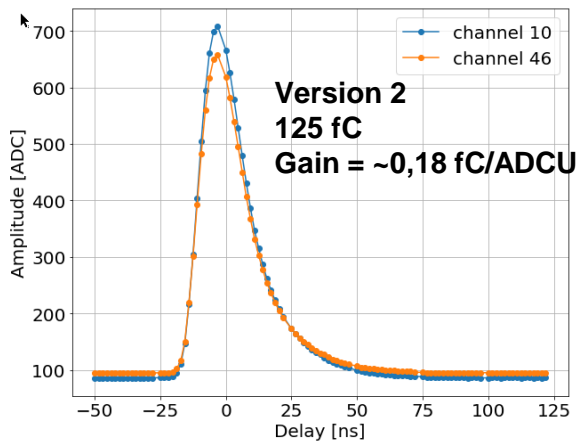
Flip-Chip on mezzanine



HD BGA board



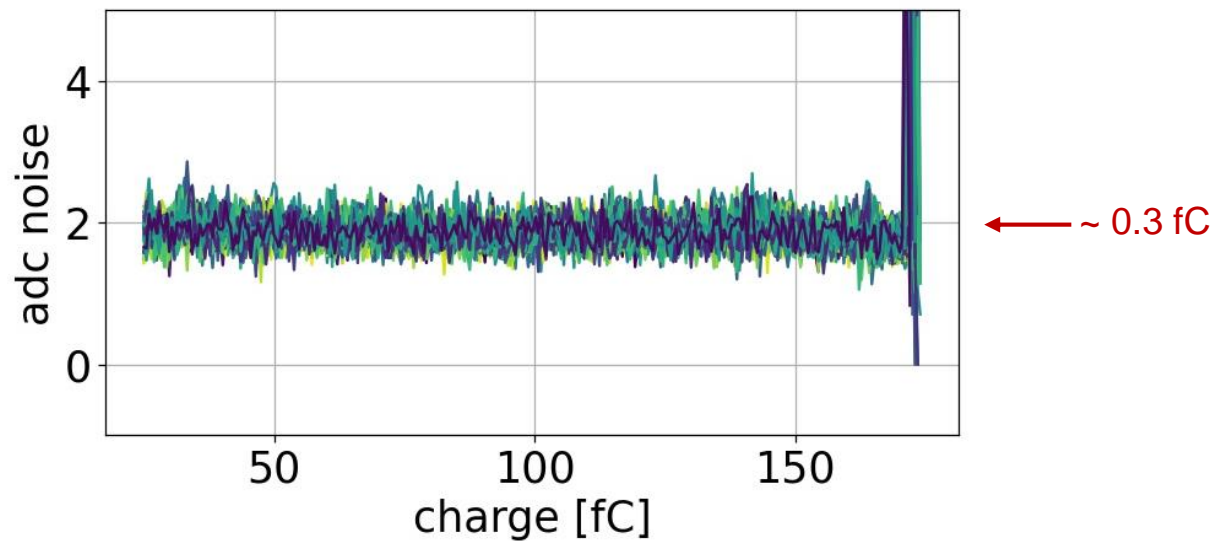
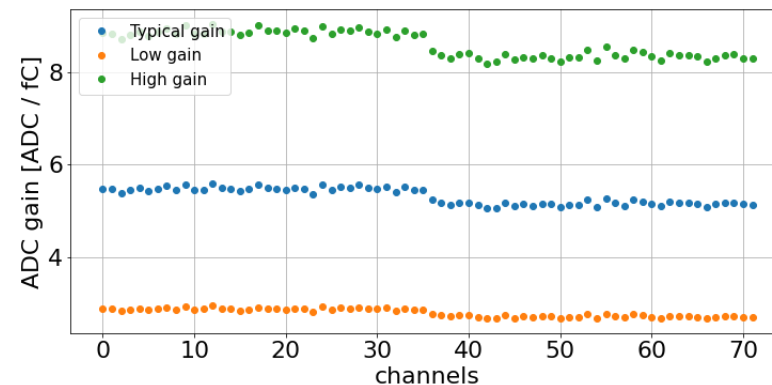
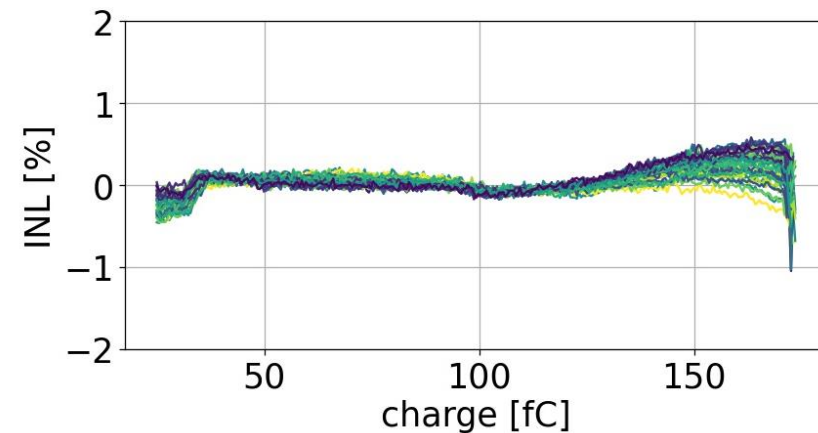
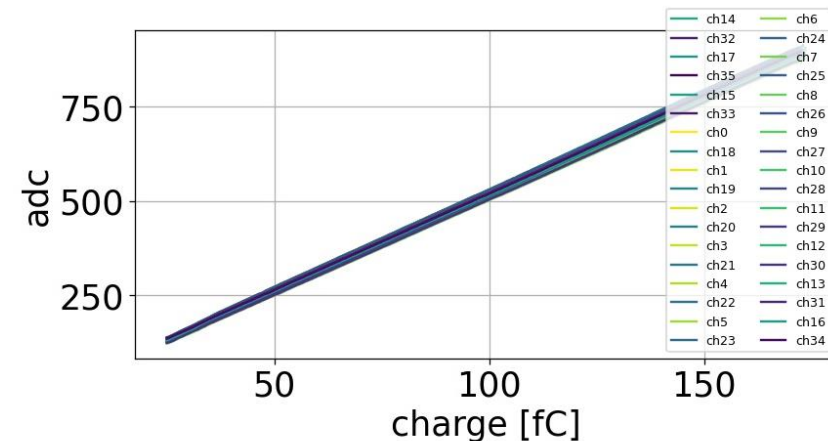
HD BGA on mezzanine



- Separates effects from the BGA substrate and PCB
- Rising time (10-90 %): ~ 15 ns
- Falling time (10-90 %): ~ 30 ns, < 20 % at BX+1
- Good uniformity over the channels
- Digital 40 MHz clock coupling on the analog signal on the HD BGA board: digital noise (slide 11)

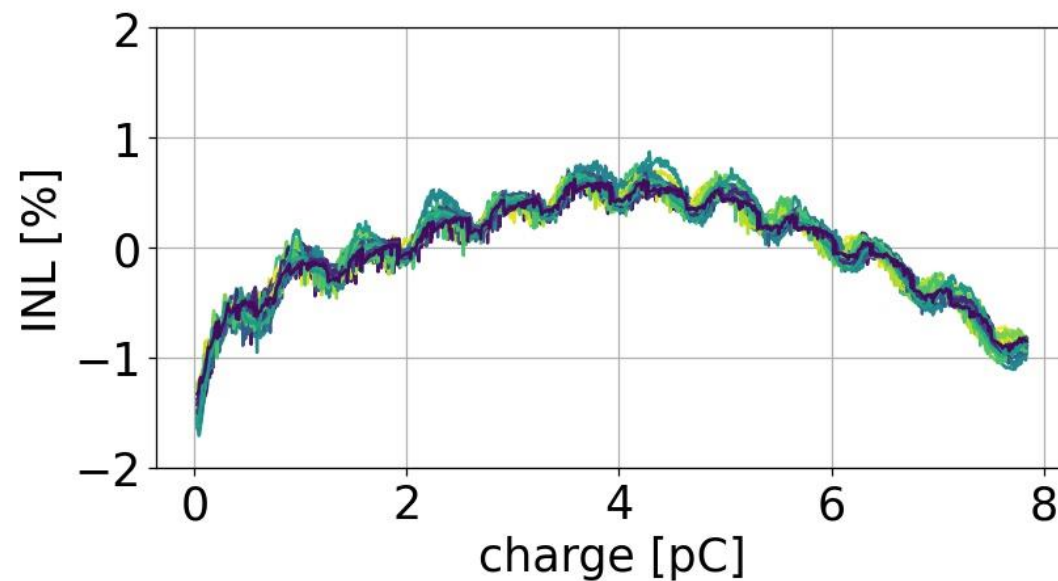
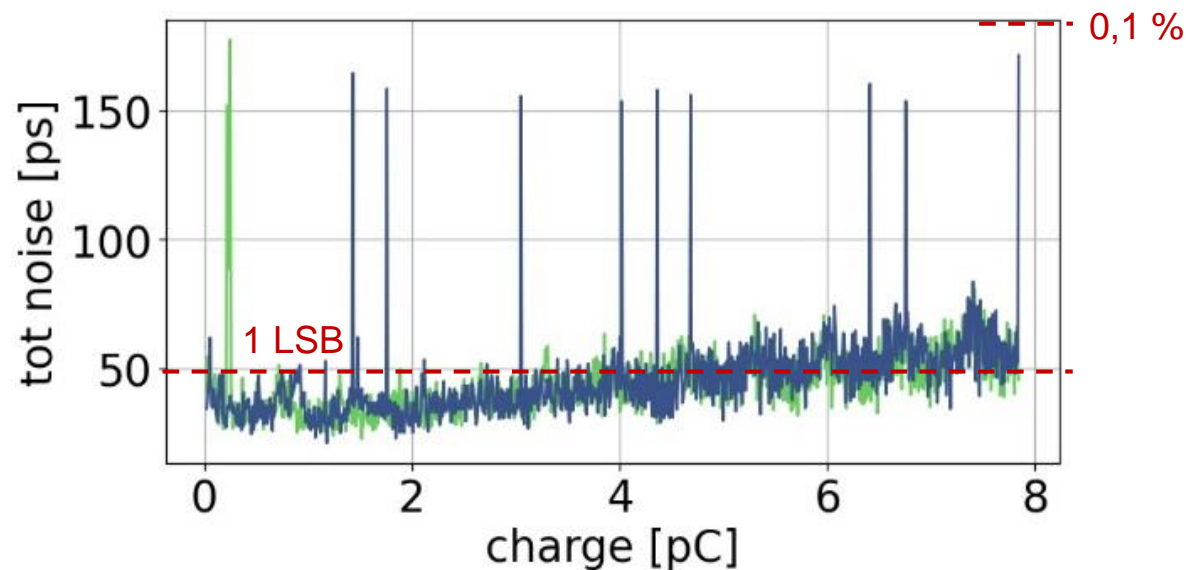
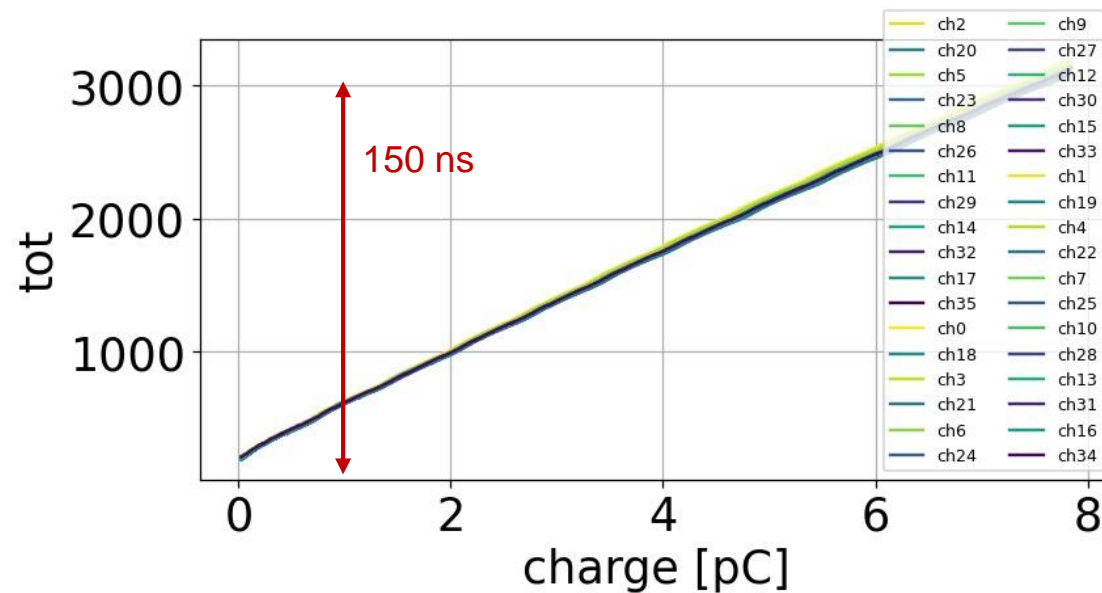
Charge ADC (0 – 160 fC)

- Two 10b-DAC to globally set the pedestal to a wanted level
- 5b-DAC to reduce dispersion per channel
 - From ~ 100 ADCu dispersion to ~ 5 ADCu
- Good linearity within $\pm 0.5\%$
 - 1.6 fC (~ 1 MIP) linearity for the typical gain
- ~ 0.3 fC resolution with 50 pF input capacitor



Charge measurement from TOT when preamplifier saturates

- 160 fC to 10 pC (for the typical preamplifier gain)
 - 12 bits over 200 ns
 - 50 ps binning
- Linearity
 - < 2% linearity
 - Better with input capacitor (as expected)
 - Small residual wiggles on TOT due to digital noise on preamplifier input
- Resolution around the LSB (~ 50 ps)
 - Some peaks due to outliers (understood and fixed)

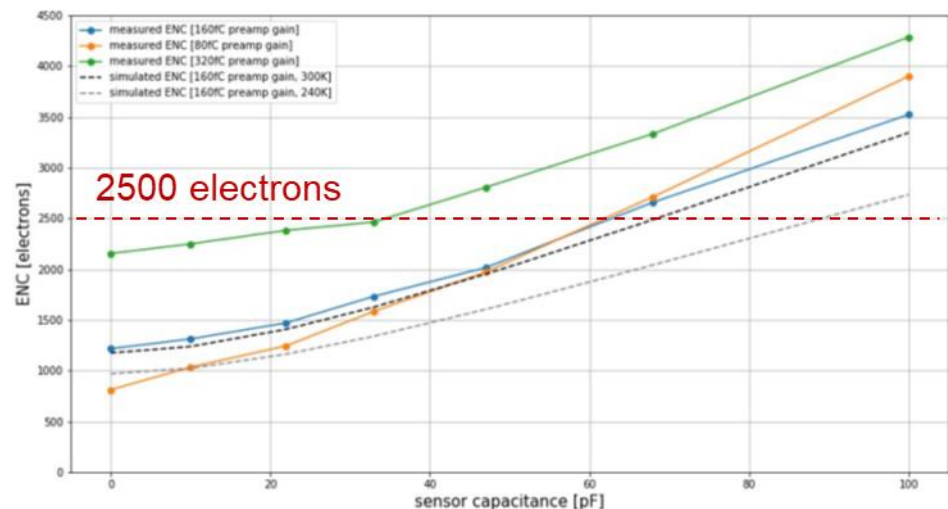


Noise

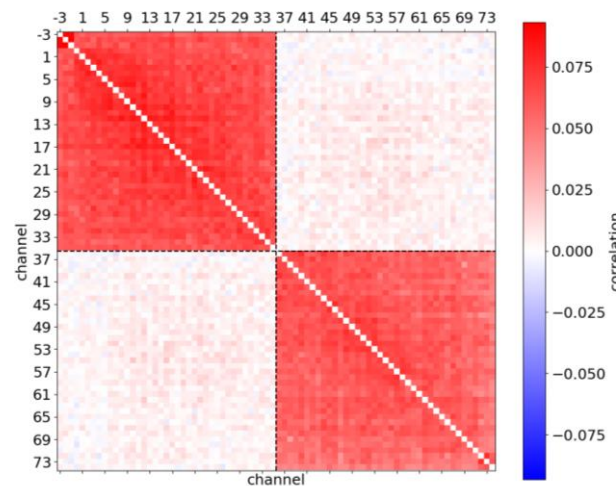
- Measured noise with 50 pF input cap = **0.3 fC** (**~ 2000 electrons**) (0.7 nV / $\sqrt{\text{Hz}}$)
- Very low correlated noise contribution: **~ 0.1**
- Coherent noise extracted by comparing direct and alternate sums on n channels (n = 72):
 - $DS = \sum ped[i]; AS = \sum (-1^i) ped[i]$
 - Incoherent noise $IN = rms(AS)/\sqrt{n}$
 - Coherent noise $CN = \sqrt{var(DS) - var(AS)}/n$

	0 pF Cdet	47 pF Cdet	68 pF Cdet
High gain ENC	900 electrons	2000 electrons	2750 electrons
Typical gain ENC	1250 electrons	2000 electrons	2700 electrons
Low gain ENC	2200 electrons	2800 electrons	3400 electrons
TOA FOM ⁽¹⁾	NA	2.5 ns/fC (FlipChip) 3 ns/fC (BGA)	3 ns/fC (FlipChip)
TOA noise floor ⁽¹⁾	20 ps	25 ps (FlipChip) 25 ps (BGA)	25 ps (FlipChip)
TOA Time-Walk	0.8 ns (FlipChip) 4 ns (BGA)	2.5 ns (FlipChip) 6.5 ns (BGA)	4 ns (FlipChip)

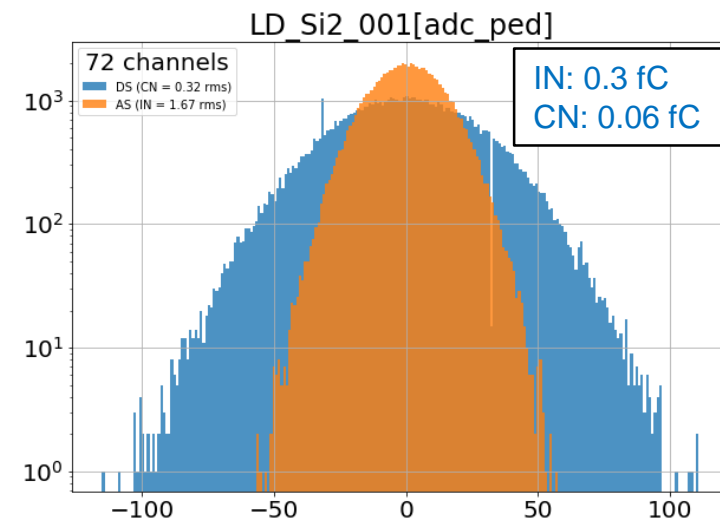
Equivalent Noise Charge wrt. sensor capacitance



Correlation matrix

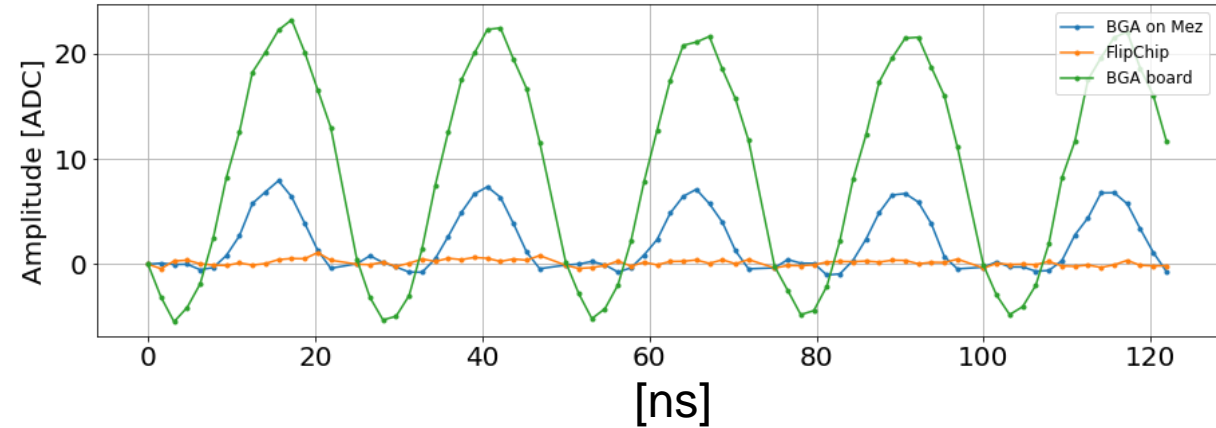


Coherent vs. incoherent noise

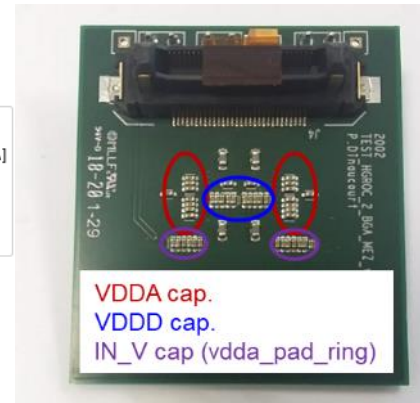
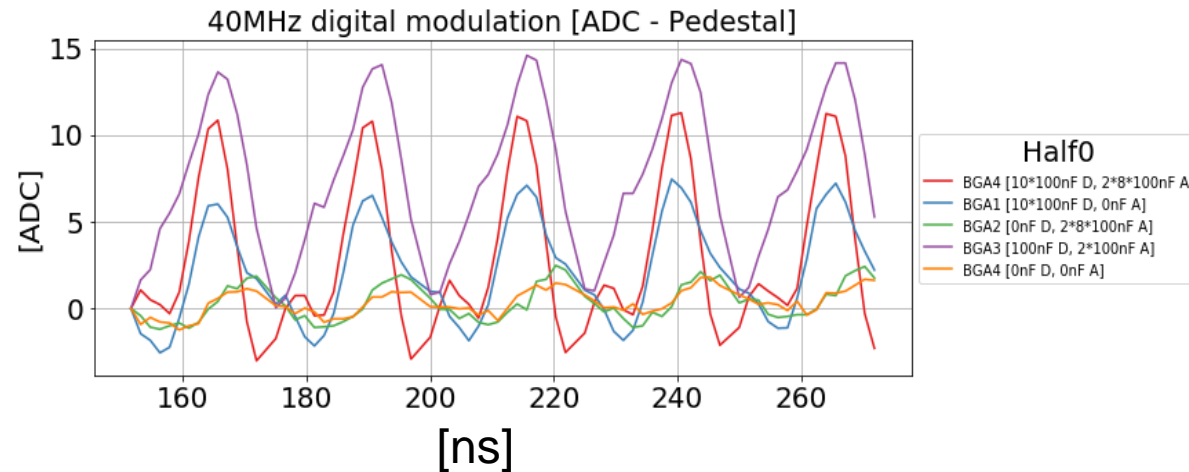


Digital noise

- A 40 MHz modulation is visible on the analog signals
 - Comes from digital current spikes on preamp ground node
 - 10 μV on ground give 1 ADC count
 - BGA worse than flip chip
 - BGA substrate optimised, improvements made by optimizing decoupling & the pcb ground impedance
 - Further improved by removing decoupling caps !
 - Reduces digital current spikes (inductance)

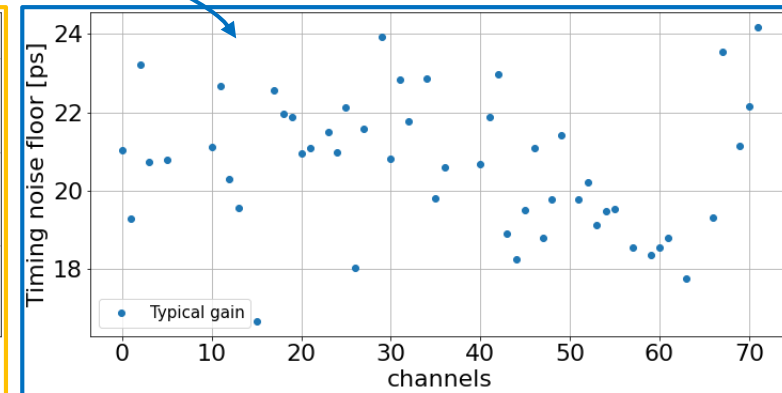
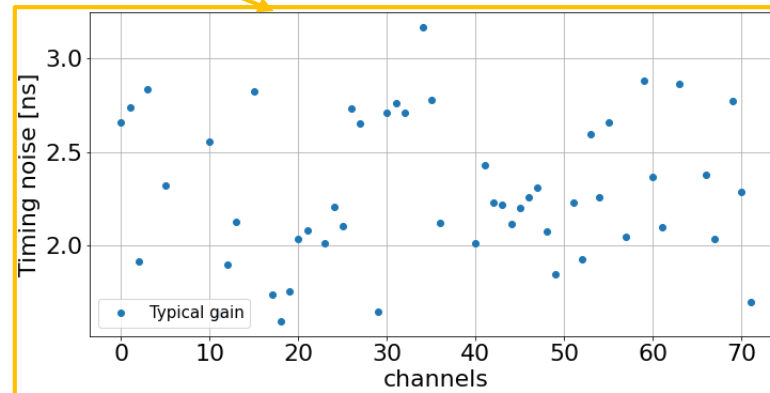
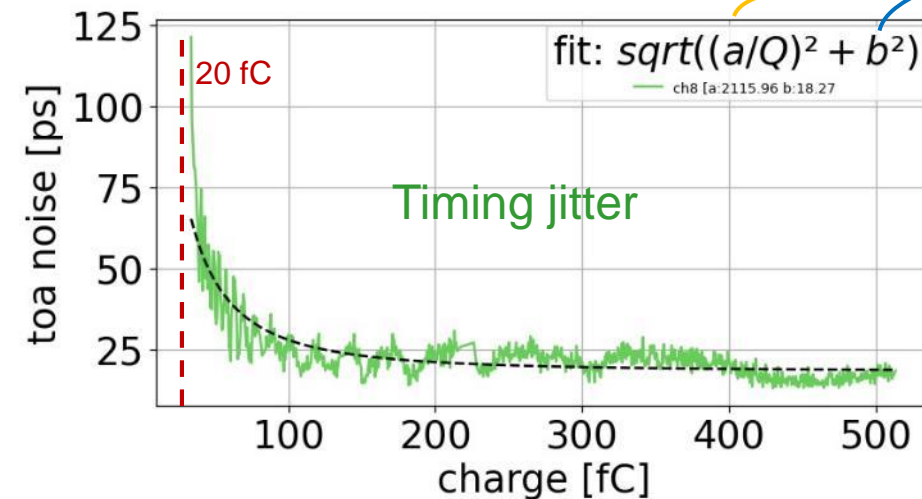
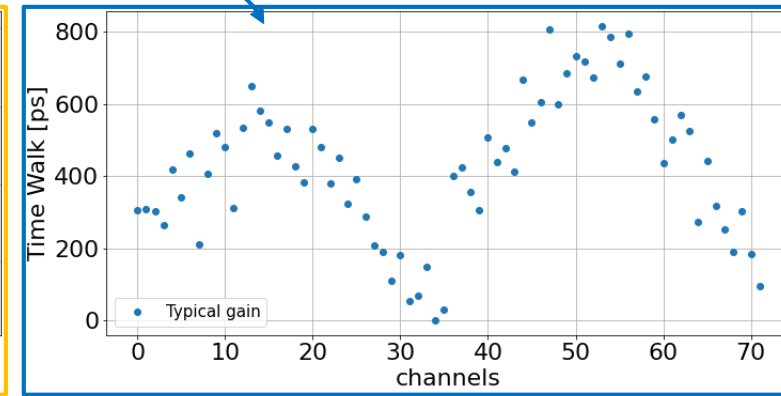
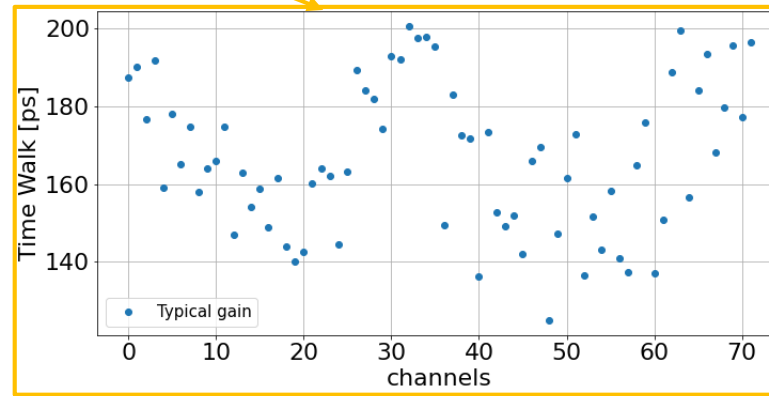
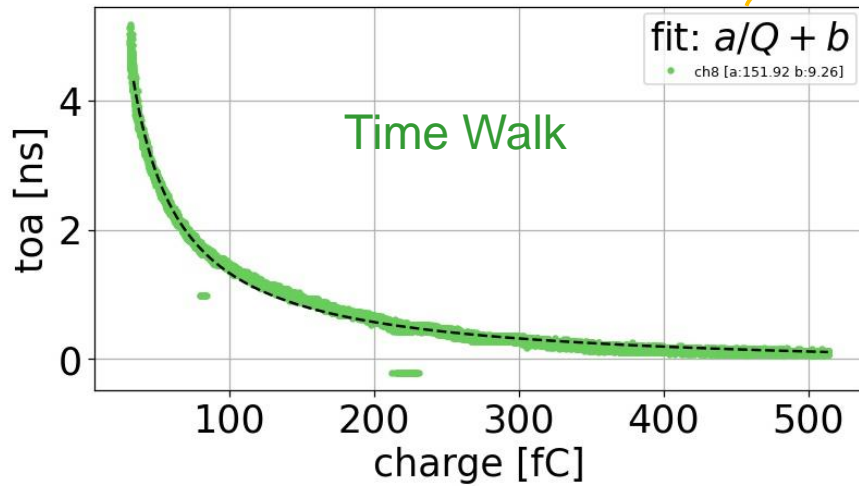


- This provides recommendations for the Hexaboard design
 - Very delicate design!

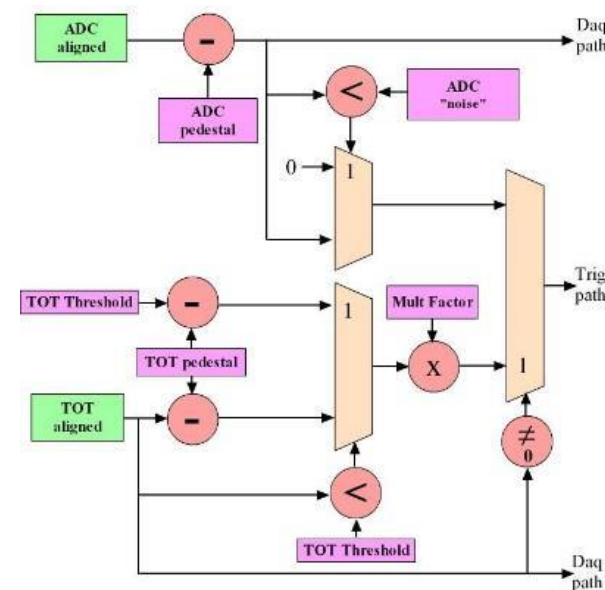
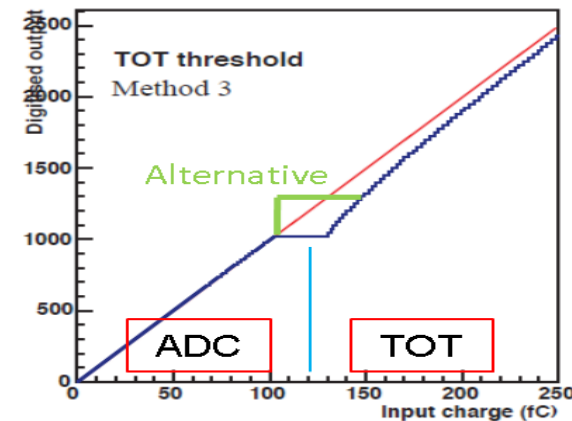
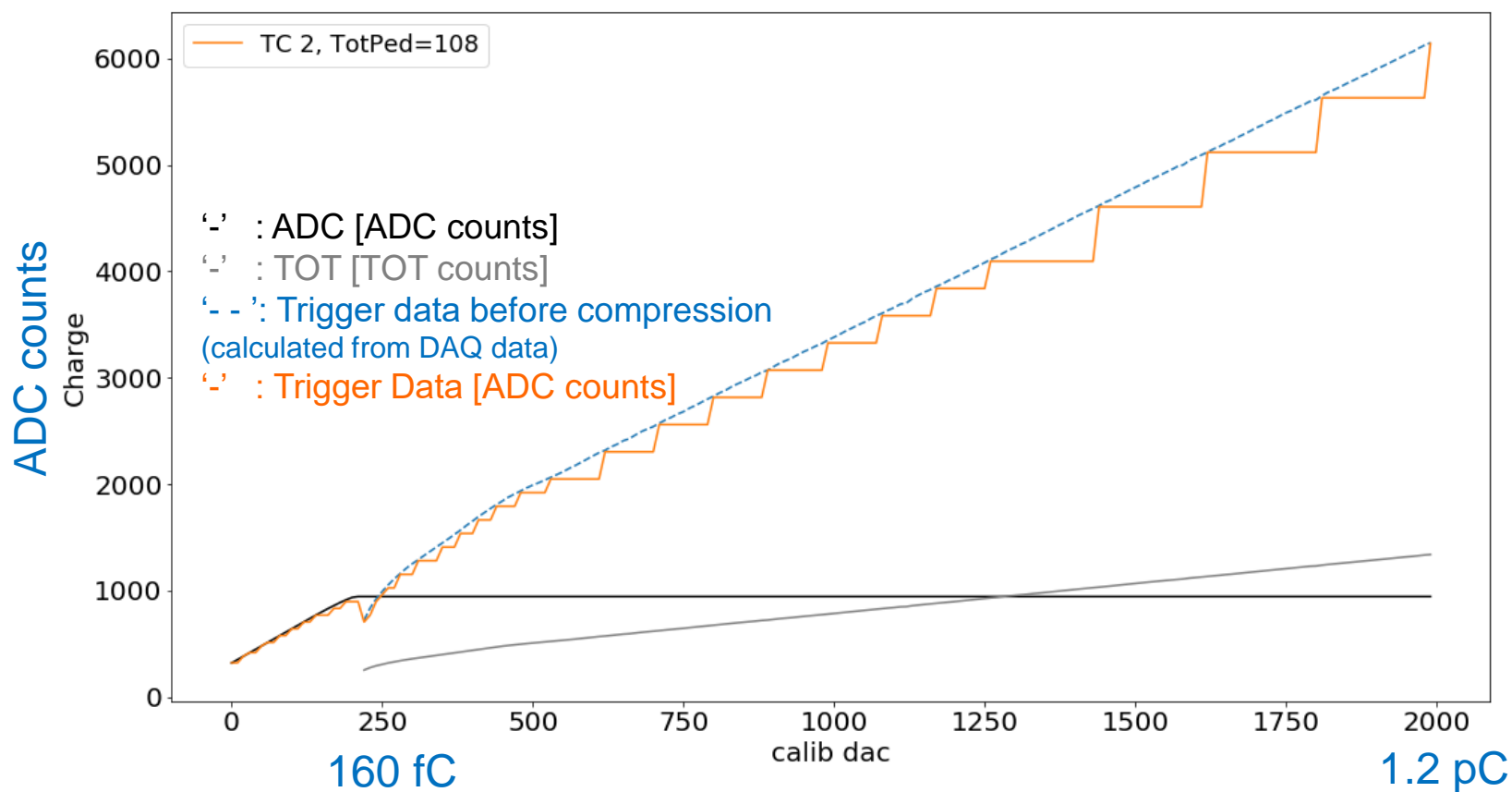


Timing performance

- Minimum Charge providing a TOA measurement is 20 fC, limited by the digital coupling
- $Time\ Walk = \frac{170 [ps]}{Q [fC]} + 400 [ps]$: 2.5 ns w/o detector cap., ~ 5 ns with 50 pF input capacitor
- $Jitter = \frac{2[ns]}{Q[fC]} (+) 22 [ps]$



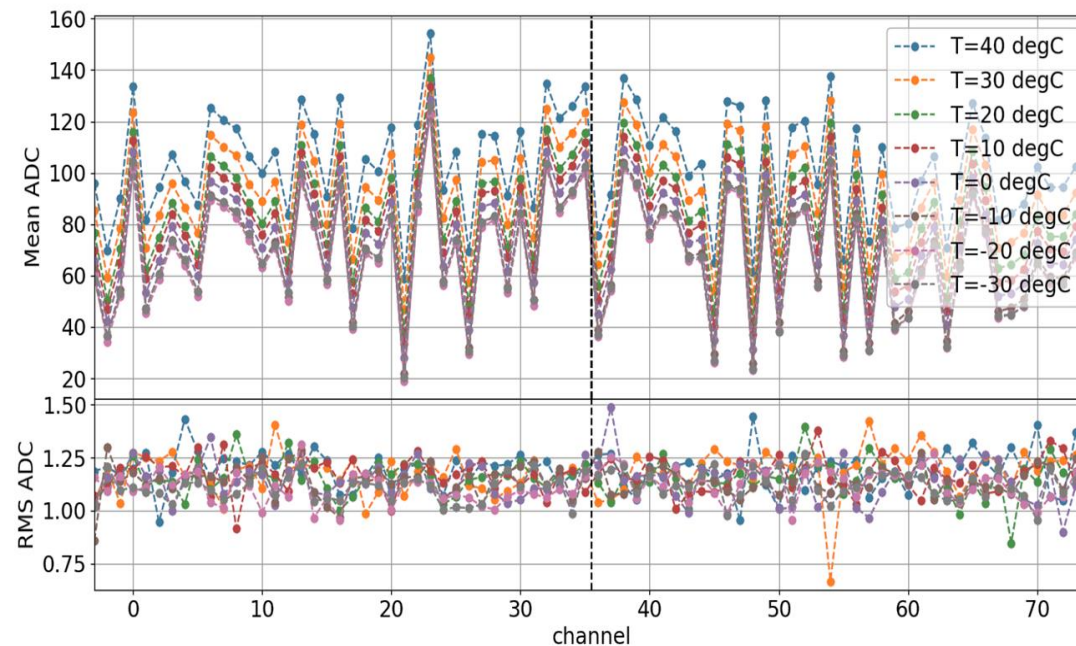
- Charge linearization over ADC and TOT range
- Sum by 4 / 9 (19b / 21b)
- Compressed to 7 bits: 4b exponent + 3b mantissa



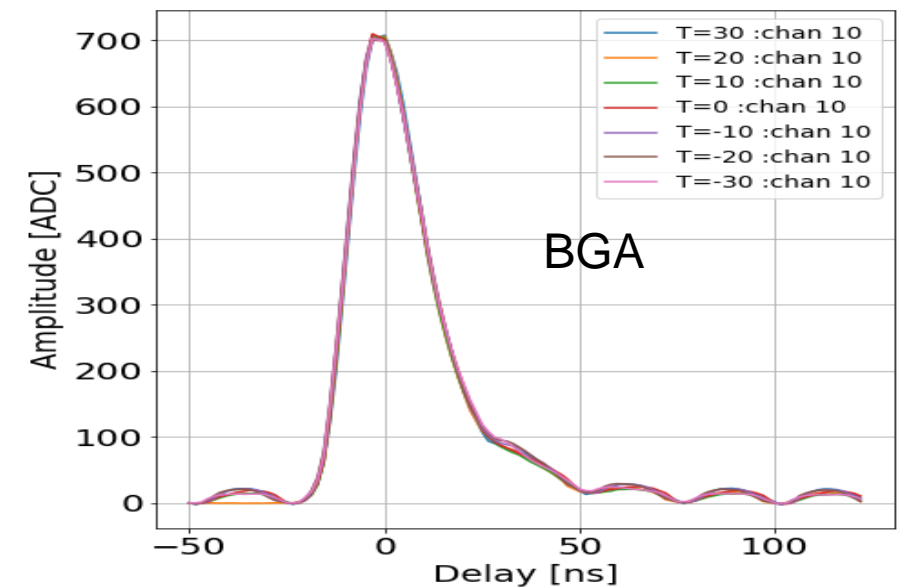
Bandgap and temperature measurements

- Readout is DC coupled to the shaper output → temperature sensitivity mitigation mandatory
- Bandgap value moves over 2mV between -40°C and +40°C, but lower shift around -20°C
 - Increasing to ~ 6-8 mV after 300 Mrad
- Pedestals
 - 1 ADC/°C at room temperature
 - ~ 0.4 ADC/°C around -30°C (Would be around 5 ADC/°C without mitigation)

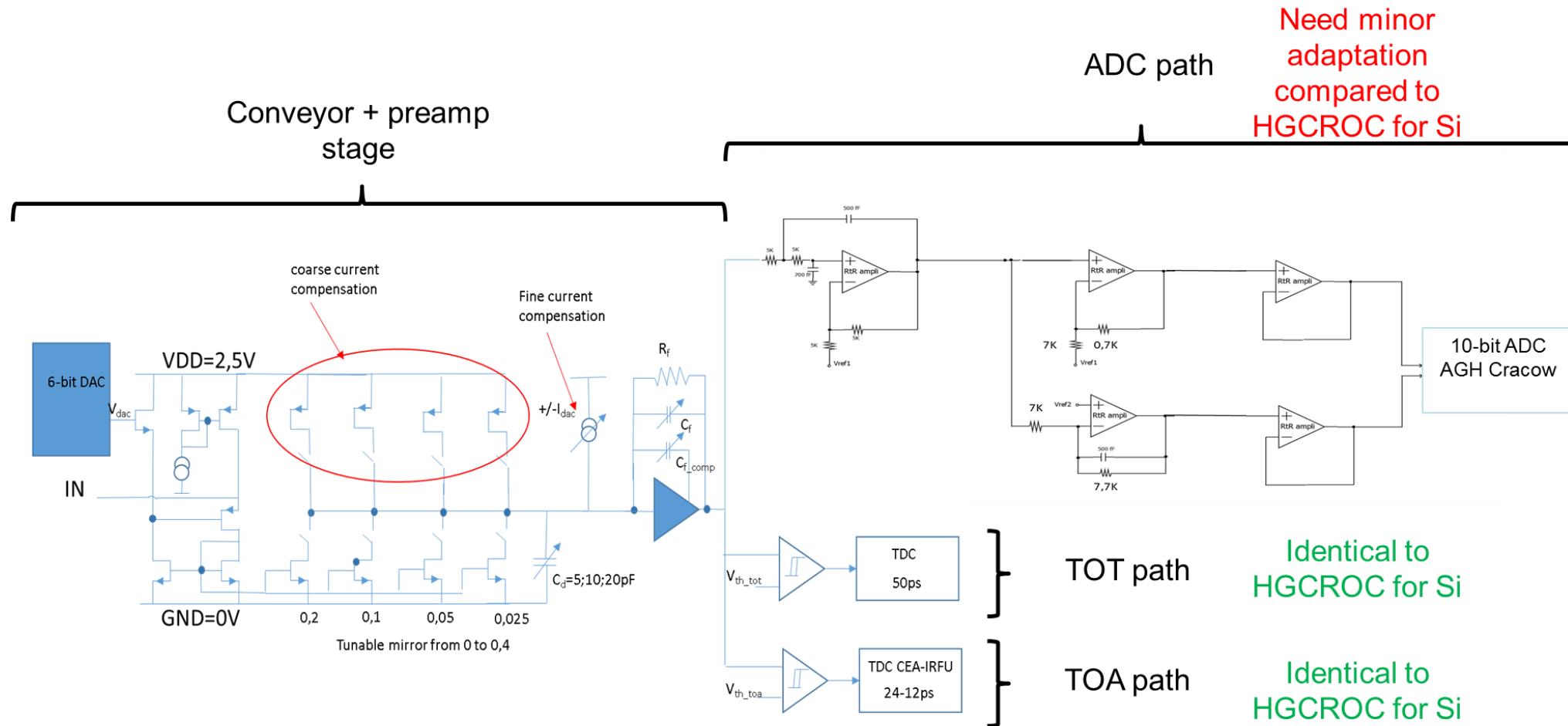
Pedestals value as a function of temperature



Gain as a function of temperature

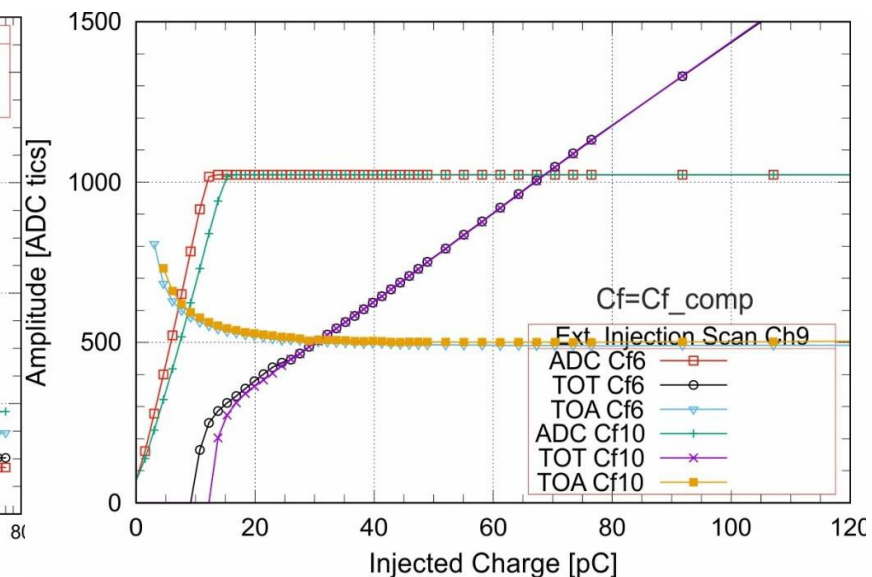
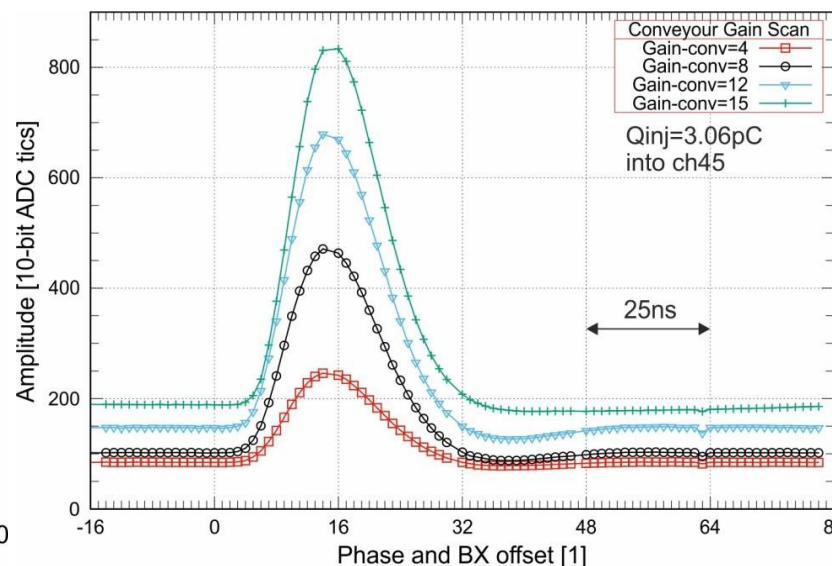
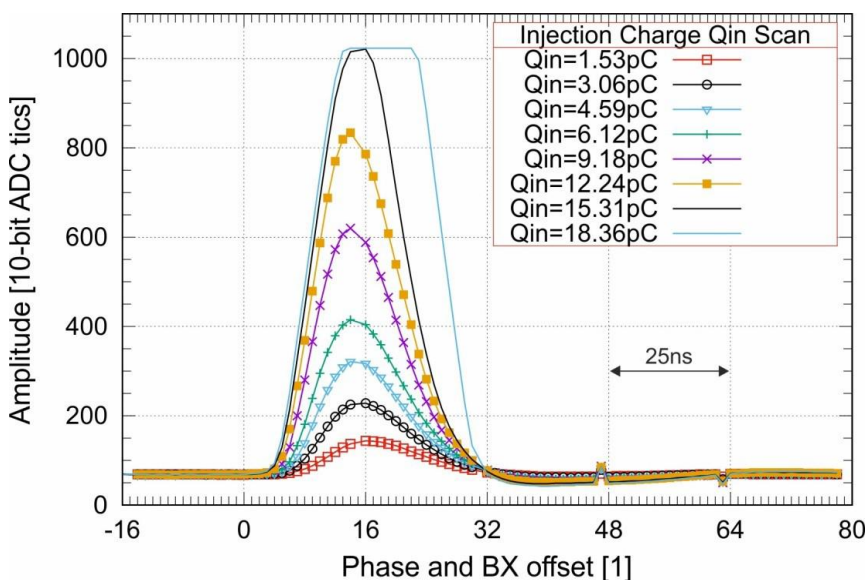
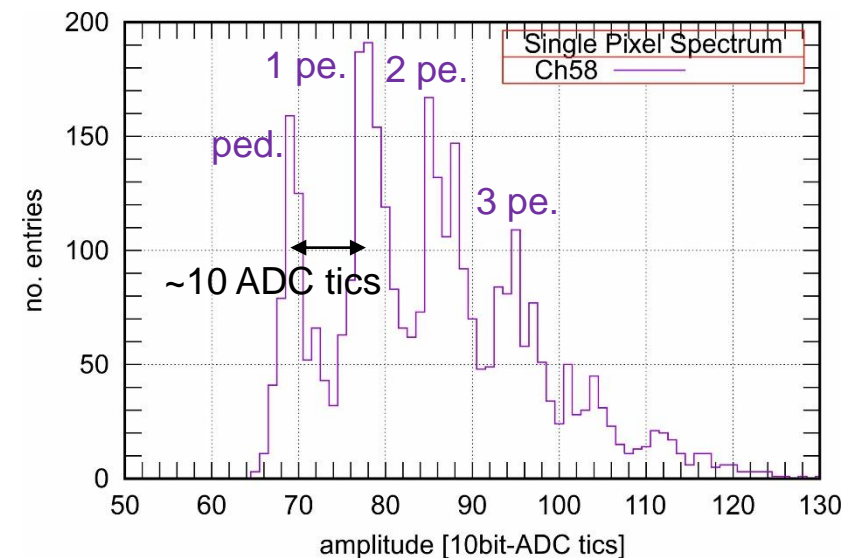


- H2GCROC2 (SiPM version) = HGCROC2 + input current conveyor
 - Heidelberg university design
- 2.5 V input stage for overvoltage adjustment



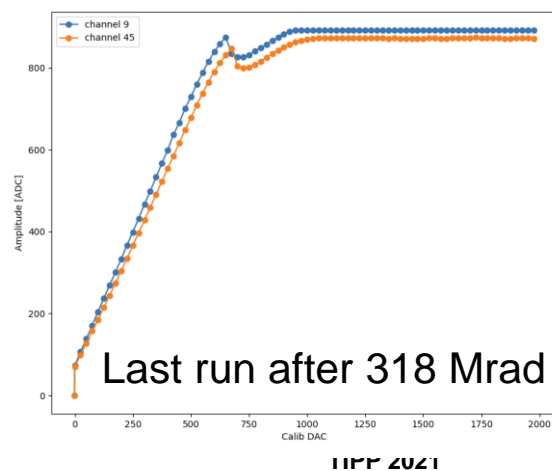
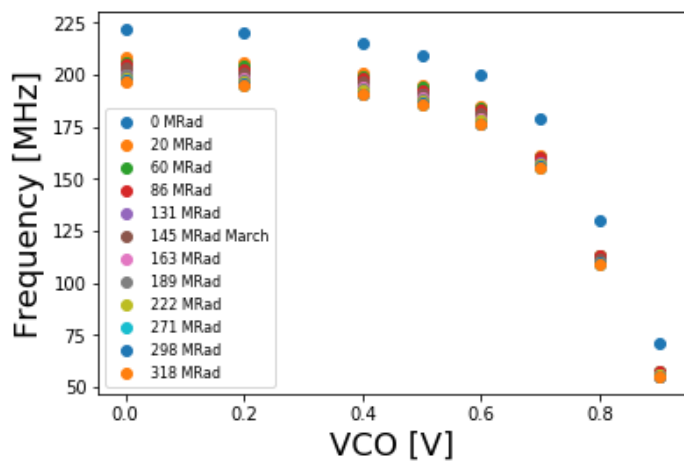
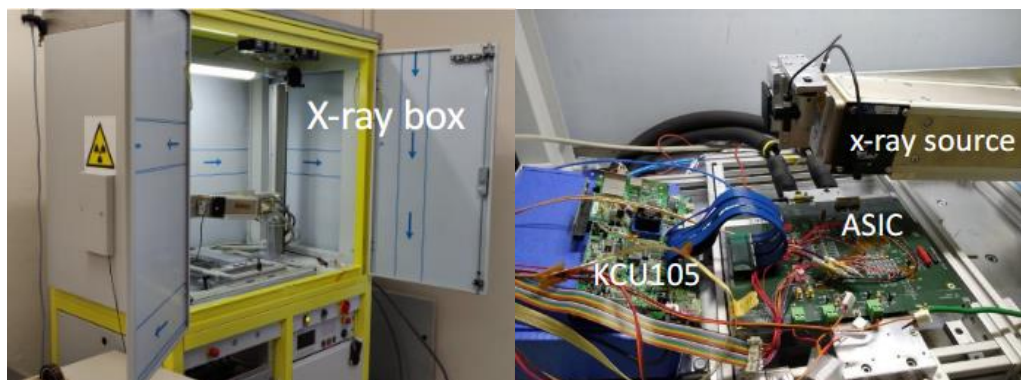
H2GCROC2 measurements:

- Good shape as expected
- Adjustable conveyor gain
- Dynamic range as expected: TOT (12bit) – 300pC aimed
- Single-Photon Spectra (SPS):
 - 4V overvoltage, conveyor gain = 12
 - Gain: ~ 10 ADC counts



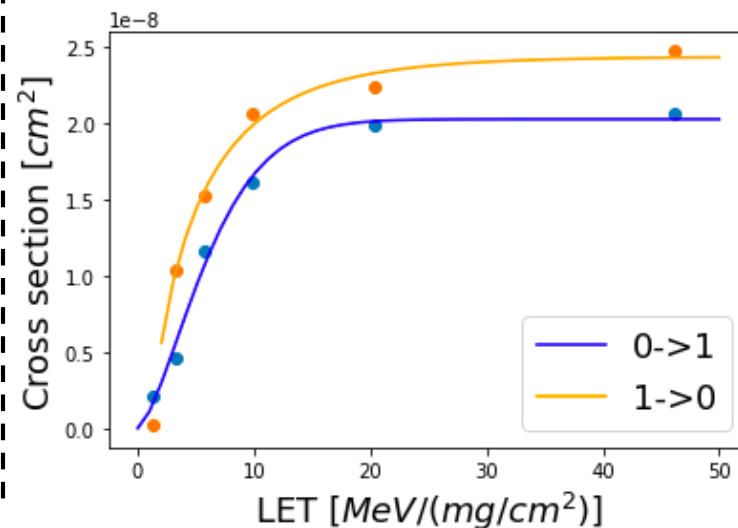
TID: thinned chip (70 μm)

- 3 campaigns at CERN:
 - Si-version at room temperature (Oct 2019)
 - Si-version at cold (Mar & Jun 2020)
 - SiPM-version at room temperature (Aug 2020)
- Irradiation up to 310 Mrad (5 Mrad for SiPM)
- The chip still works after annealing

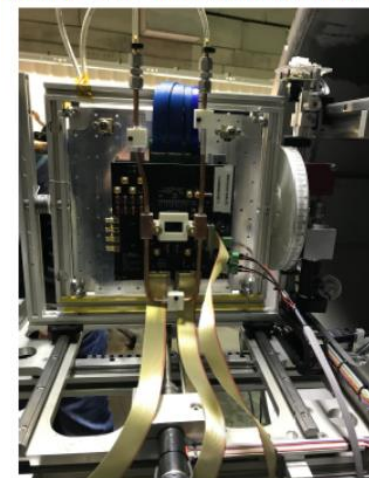


SEE: 2 campaigns at Louvain (Nov 2019 & Feb 2020)

- **Heavy ions LET 5 – 45 MeV**
- I2C acted as expected: cross section follows the usual curve and **no errors recorded when auto-correction set. Flips < 2 E-7 Hz/chip**
- **Bit shifting in the DAQ link (no triplicated Serializer in HGCR0Cv2)**
- No signs of latch-up effects
- No flips seen in ADC/TOA/TOT data



Inside the vacuum chamber



Analog Upgrades:

- Increased sensor leakage current compensation
- Increase resolution (12bit) of DAC and phase shifter for calibration
- TOA calibration with a Randomized Pulse Generator

Triplicated Logic

- Memory pointers and calculation logic

High Speed Serial links

- SEE tolerant

Control

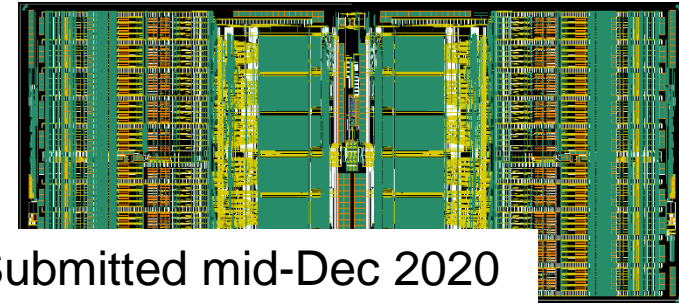
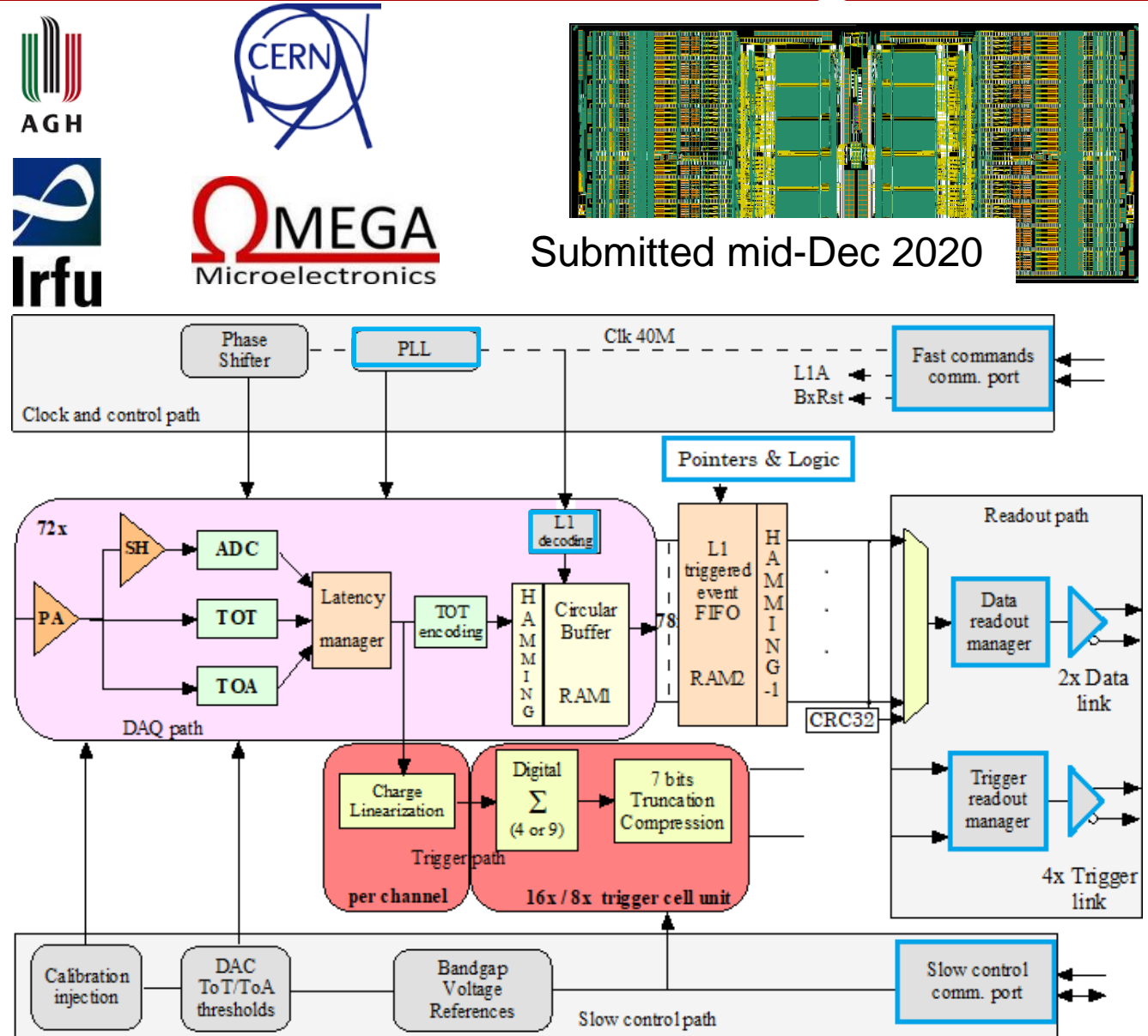
- SEU tolerant Fast commands
- SEU tolerant I2C Module for slow control

Memory blocks

- Adds trigger derandomizing buffer for DAQ path
- Custom Hamming Encoding at the entry of Circular Buffer
- Hamming Decoding after the L1 FIFO

Checksum

- CRC-32 Checksum Encoding before serialization

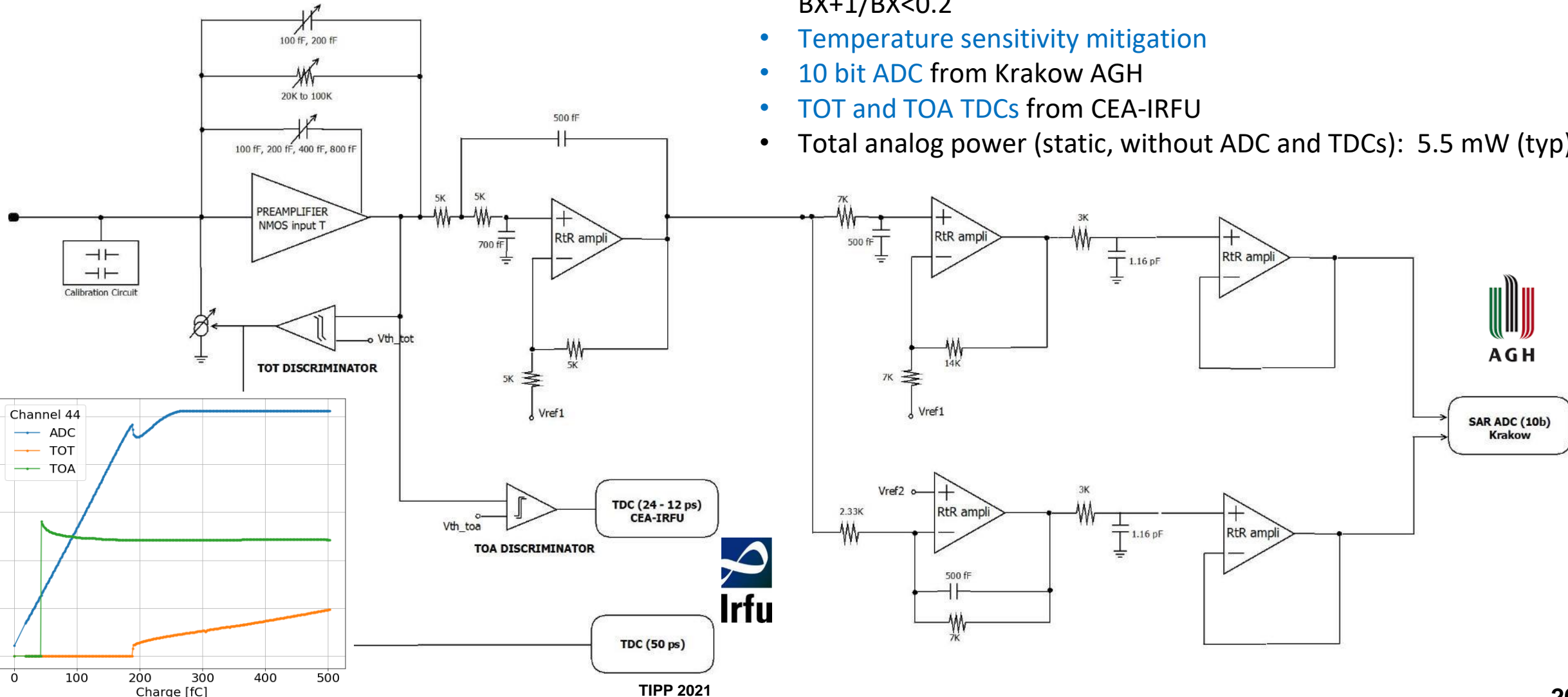


: Triplicated Module

- With HGCROC2, a big step has been taken to reach the HGCal's requirements. It is probably one of the most complex chips ever designed for imaging calorimetry: high dynamic range, precision timing measurements, high speed links, a lot of digital, harsh radiation environment...
- Measurements now well advanced on both Silicon and SiPM versions @CERN, LLR, IRFU, Desy and OMEGA
 - Charge performance reaches the specification: 1 % linearity, for both ADC and TOT
 - Timing performance: time walk calibration feasible, jitter below 25 ps
- Digital coupling
 - Good performance at chip level
 - Stringent requirement on PCB design: low ground impedance and optimized power decoupling
 - First studies on the Hexaboard and Tileboards started
- TID and SEE campaign show very good performance
 - SEEs appear only in the non-triplicated parts of the chip (as expected)
 - SEU errors on the slow-control parameters $< 1e-7$ Hz
- HGCROC3 will be the final version...
 - Now at packaging, awaited end-May

- Calibration pulser, 0.5pF and 8 pF calibration cap.
- Preamp : adjustable gains for 80, 160 and 320 fC ranges
- Tunable TOT threshold

- Compensation for the leakage current, 10 μ A max. (not shown in the figure)
- Sallen Key shaper RC4, $t_p < 25$ ns, tunable ($\sim 20\%$) with 2 bits, $BX+1/BX < 0.2$
- Temperature sensitivity mitigation
- 10 bit ADC from Krakow AGH
- TOT and TOA TDCs from CEA-IRFU
- Total analog power (static, without ADC and TDCs): 5.5 mW (typ)

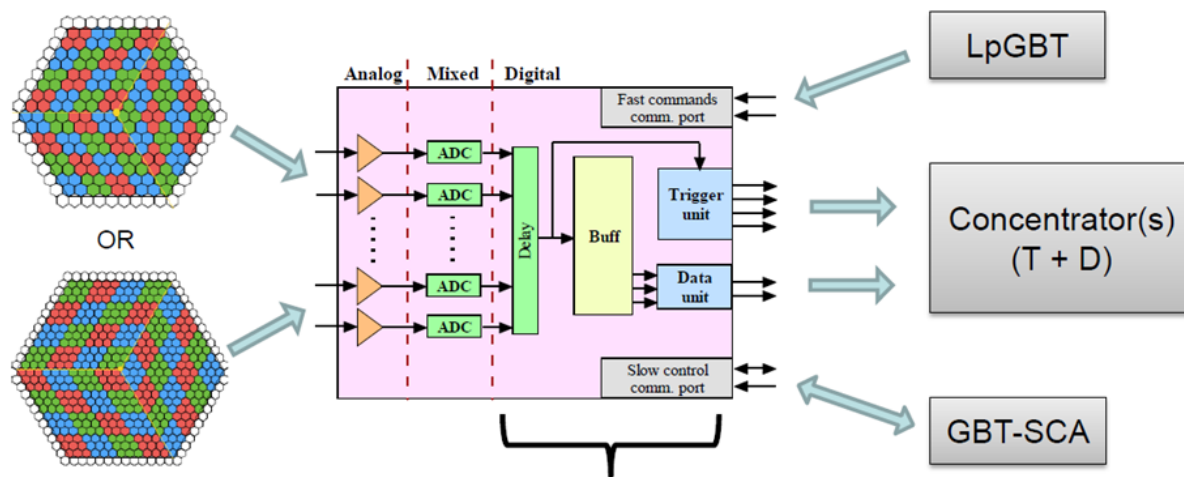


SAR ADC (10b)
Krakow



- The chip provides compressed data of the charge to the L1 trigger processing

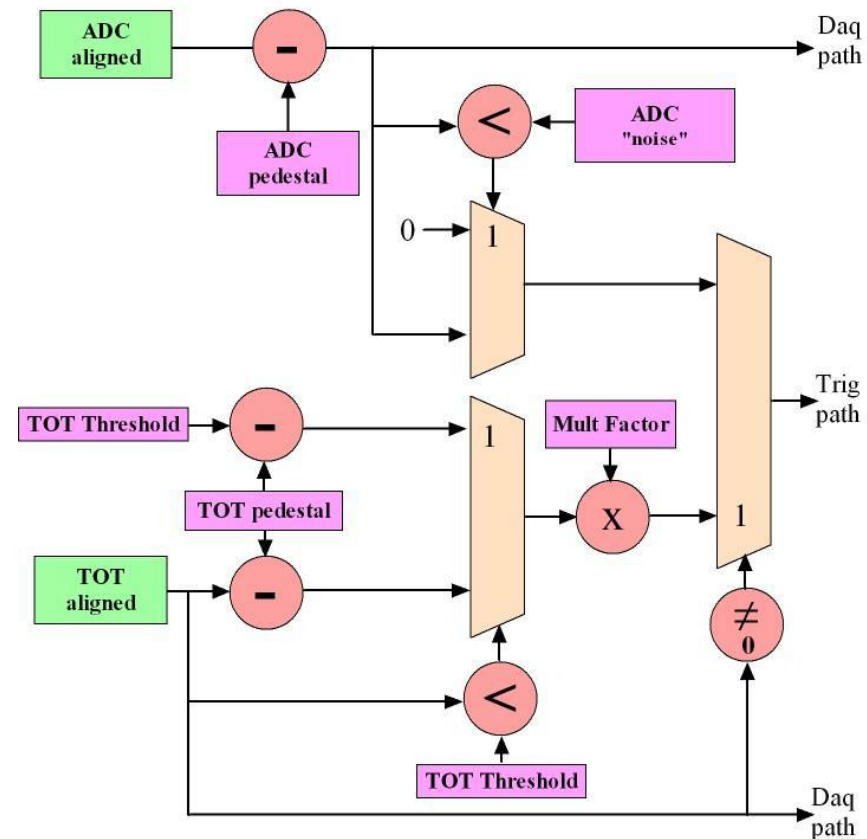
- Charge linearization over ADC / TOT range
- Sum of 4 (or 9) channels (depending on the sensor)
- Charge compression to fit the bandwidth (4b Exponent + 3b Mantissa)
- 4 E-link transmitters, CLPS @ 1.28 Gbps



HGCROC2 integrates 72 channels to readout

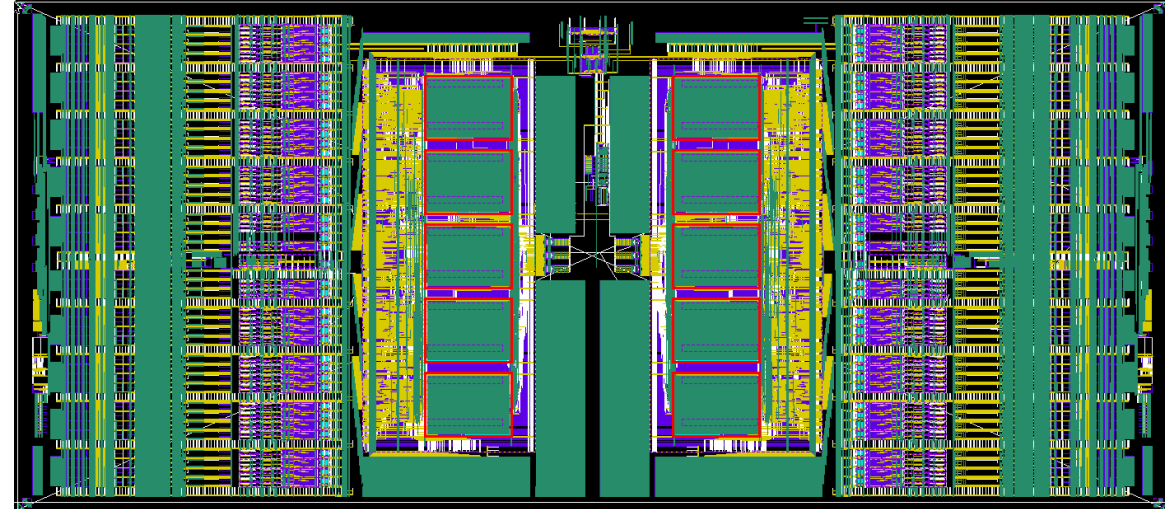
- 192 channels sensor with a 64-ch configuration
- 432 channels sensor with a 72-ch configuration

TIPP 2021

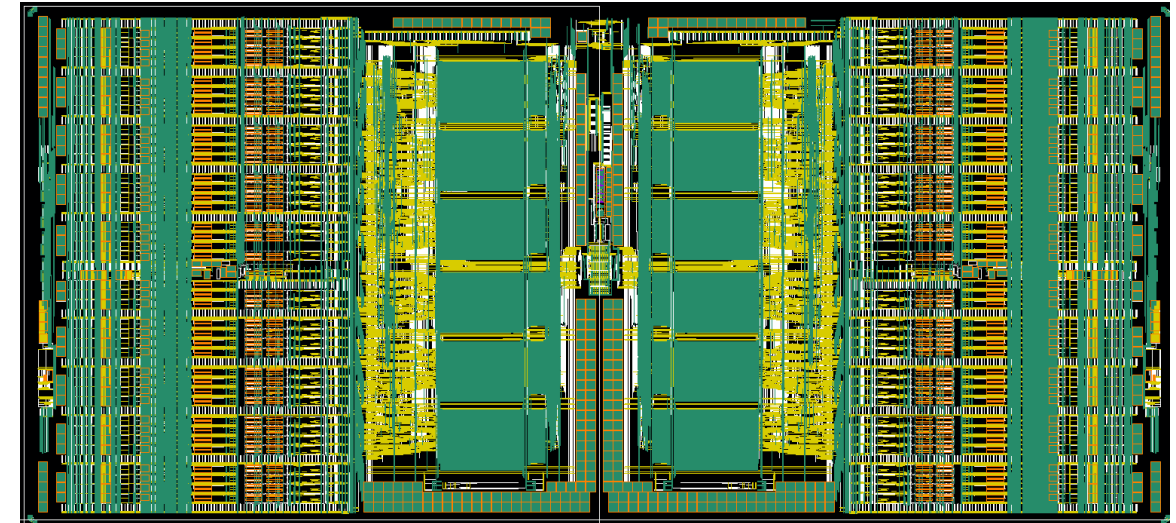


- Sharp eye needed to distinguish the differences!
 - Same size/pinout/package
 - Small improvements in analog part
 - Adds derandomizer after L1 DRAM
 - SEU mitigation done in digital part
 - More robust I²C and fast command decoder
 - Final functionality
- **Completed end-october**
 - Reviewed Dec 1st
 - **Submitted mid-Dec**

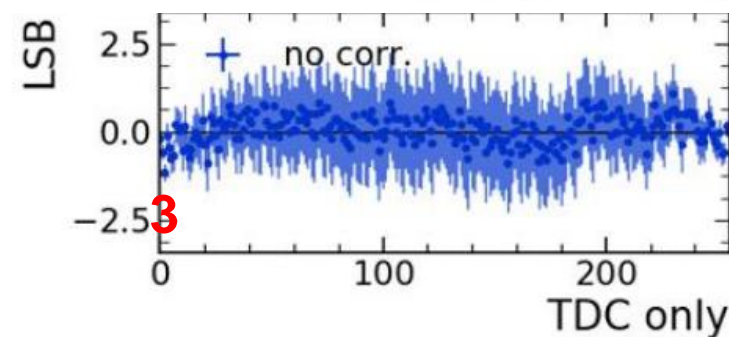
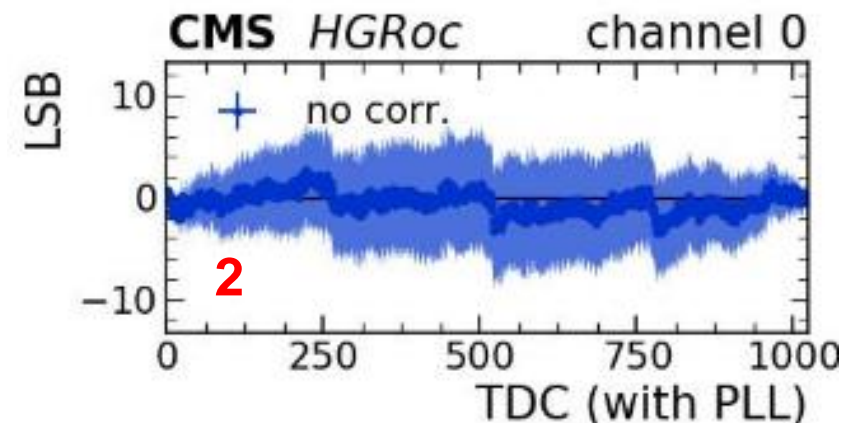
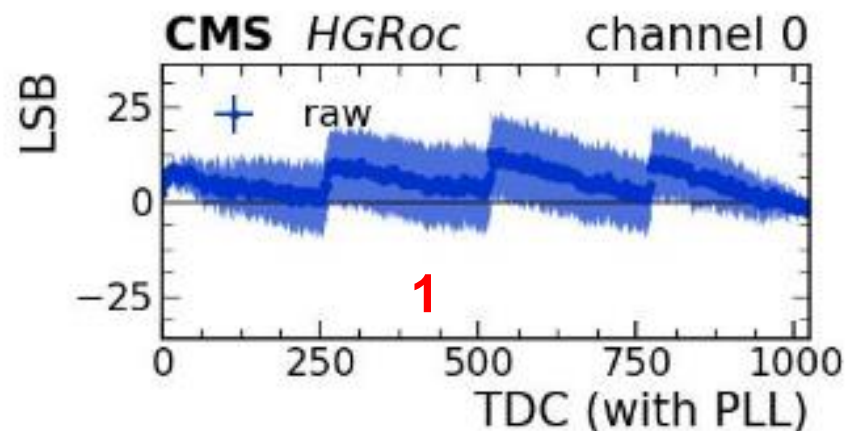
V2



V3



- Channel-wise TDC made of a fine TDC (3 bits), a coarse TDC (5 bits) and a counter (2 bits)
 - Works only when a ToA (or a ToT) occurs
- Two servo-controlled master TDCs allow to control speed of channel-wise TDCs
 - First adjustment allow to configure the Master TDC (1 → 2)
 - External pulse injection used here : will be replaced by a Random Pulse Generator (RPG) in V3



- Then channel-wise adjustment allows to achieve the best performances (~1 LSB INL)

- S-curves show the ToA efficiency as a function of the charge
- Minimum charge providing ToA events is 20 fC
 - Limited by the digital coupling

