

SGWB characterization with LISA

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GW primordial cosmology workshop

May 17, 2021

Outline

- 1 Introduction
 - Signals at LISA
 - Data generation and pre-processing
- 2 Binned reconstruction (SGWBinner)
 - Methodology
 - Some examples
- 3 PCA reconstruction
 - Methodology
 - Some examples
- 4 Conclusions

SGWBs at GW detectors

The **data** \tilde{d} (in frequency space) can be expressed as

$$\tilde{d} = \tilde{s} + \tilde{n}$$

For an **isotropic SGWB** $\rightarrow \langle h_\lambda(\vec{k}) h_{\lambda'}^*(\vec{k}') \rangle = P_h^\lambda(k) (2\pi)^3 \delta_{\lambda\lambda'} \delta(\vec{k} - \vec{k}')$

Assuming $\langle \tilde{s}\tilde{n} \rangle = 0$ and Gaussian signal

$$\langle \tilde{d}^2 \rangle = \langle \tilde{s}^2 \rangle + \langle \tilde{n}^2 \rangle = \mathcal{R} P_h^\lambda + N \equiv \mathcal{R} [P_h^\lambda + S_n]$$

where we have introduced

- The **response function** of the instrument \mathcal{R}
- The **noise power spectrum** N
- The **(square of the) Strain sensitivity** S_n (in 1/Hz)

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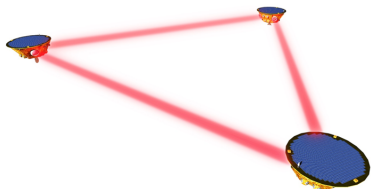
In order to compare with cosmological predictions it's customary to introduce

$$\Omega_n(f) = \frac{4\pi^2}{3H_0^2} f^3 S_n(f), \quad \text{and} \quad \Omega_{\text{GW}} \equiv \frac{1}{3H_0^2 M_p^2} \frac{\partial \rho_{\text{GW}}}{\partial \ln f} = \frac{4\pi^2}{3H_0^2} f^3 \sum_\lambda P_h^\lambda$$

where $H_0 \simeq 3.24 \times 10^{-18} h_0$ Hz is the Hubble constant today.

Laser Interferometer Space Antenna

Few details on **LISA**:



- First **direct GW** detector **in space**
- Constellation of three satellites
- **2.5 million km** arm lengths
- Peak sensitivity $10^{-2} \div 10^{-3}$ Hz
- Three correlated interferometers (XYZ basis)
- \sim two independent detectors (AET basis)
- Expected launch in **2034**
- Operating for **4yrs (nominal)**

Very interesting for cosmology since we can:

- Measure H_0 (see 1601.07112)
- Test modified gravity (see 1906.01593)
- **(Hopefully) detect and characterize SGWBs!** (This talk!)

LISA response function

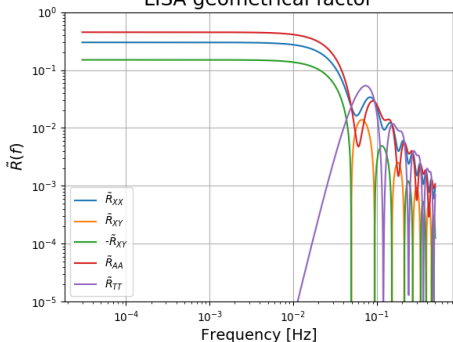
For an **isotropic** and **non-chiral spectrum** we get (see 2009.11845):

$$\langle \Delta F_{i(jk)}^{TDI} \Delta F_{l(mn)}^{TDI} \rangle = \int dk P_h(k) \mathcal{R}_{ij}(k), \quad \mathcal{R}_{ij}(k) \equiv 4 (2\pi kL)^2 |W(kL)|^2 \tilde{\mathcal{R}}_{il(jk)(mn)}(k).$$

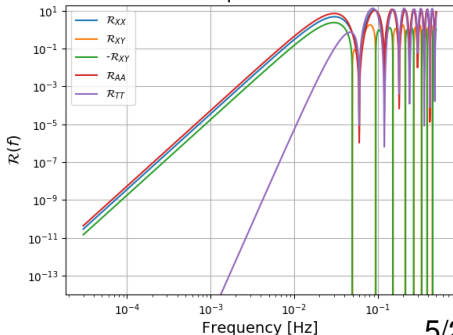
where $\mathcal{R}_{ij}(k)$ is the **LISA response function**.

For XYZ/AET (AET is noise diagonal) combinations we get:

LISA geometrical factor



LISA response function



The noise model I

Two analytical approximations for **acceleration** and **interferometric** noise:

$$P_{acc}(f, A) = A^2 \cdot 10^{-30} \cdot \left[1 + \left(\frac{4 \cdot 10^{-4}}{f} \right)^2 \right] \left[1 + \left(\frac{f}{8 \cdot 10^{-3}} \right)^4 \right] \left(\frac{1}{2\pi f} \right)^4 \left(\frac{2\pi f}{c} \right)^2,$$

$$P_{IMS}(f, P) = P^2 \cdot 10^{-24} \cdot \left[1 + \left(\frac{2 \cdot 10^{-3}}{f} \right)^4 \right] \left(\frac{2\pi f}{c} \right)^2.$$

The **power spectral densities** are ($L = 2.5 \times 10^9$ m is the arm length):

$$P_{PSD}^{XX}(f) = 16 \sin^2 \left(\frac{2\pi fL}{c} \right) \left\{ P_{IMS}(f, P) + \left[3 + \cos \left(\frac{4\pi fL}{c} \right) \right] P_{acc}(f, A) \right\},$$

$$P_{PSD}^{XY}(f) = -8 \sin^2 \left(\frac{2\pi fL}{c} \right) \cos \left(\frac{2\pi fL}{c} \right) \{ P_{IMS}(f, P) + 4P_{acc}(f, A) \},$$

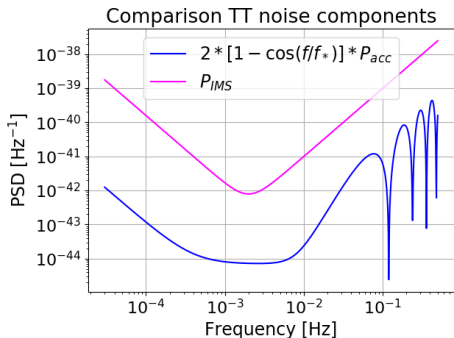
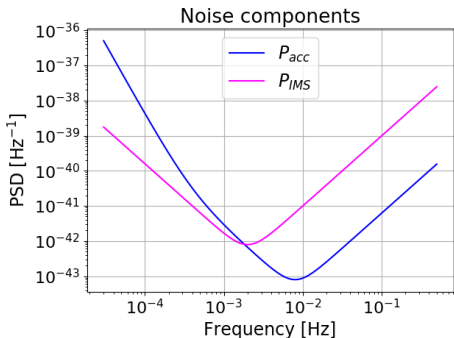
which for the TT combination gives:

$$P_{PSD}^{TT}(f, A, P) = 16 \sin^2 \left(\frac{2\pi fL}{c} \right) \left\{ 2 \left[1 - \cos \left(\frac{2\pi fL}{c} \right) \right]^2 P_{acc}(f, A) + \left[1 - \cos \left(\frac{2\pi fL}{c} \right) \right] P_{IMS}(f, P) \right\}.$$

The noise model II

At low frequencies this becomes ($f_* \equiv (2\pi L/c)^{-1} \simeq 0.019$ Hz):

$$P_{PSD}^{TT}(f, A, P) \simeq 8 \left(\frac{f}{f_*} \right)^2 \sin^2 \left(\frac{f}{f_*} \right) \left[\left(\frac{f}{f_*} \right)^2 P_{acc}(f, A) + P_{IMS}(f, P) \right],$$

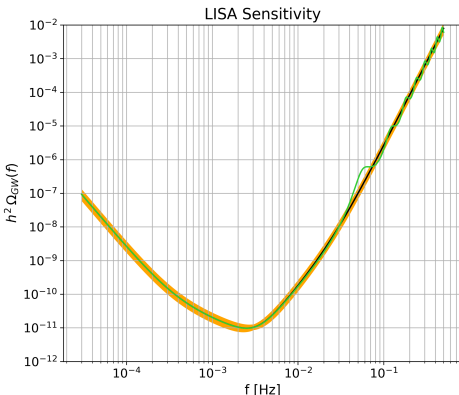
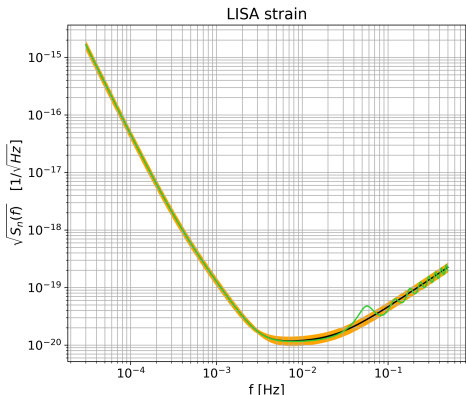


The **interferometric noise dominates TT at all frequencies!**

Analytical vs numerical

An approximation of the **response function** and the **strain sensitivity** are:

$$\tilde{R}(f) = \frac{0.3}{1 + 0.6 \left(\frac{2\pi fL}{c}\right)^2}, \quad S_n(f, P, A) = \frac{P_{PSD}(f, P, A)}{\tilde{R}(f) \times 16 \sin^2\left(\frac{2\pi fL}{c}\right) \times \left(\frac{2\pi fL}{c}\right)^2}.$$



Central values in black (analytical) / green (numerical) $\pm 20\%$ in orange

Data generation

Assume **signal and noise** (Ω units) to be **Gaussian distributed**
 The spectra (Ω_{GW} and Ω_n) quantify the **variance of fluctuations**

$$\tilde{s}_c(f_i) = \left| \frac{G(0, \sqrt{\Omega_{\text{GW}}(f_i)}) + i G(0, \sqrt{\Omega_{\text{GW}}(f_i)})}{\sqrt{2}} \right|$$

$$\tilde{n}_c(f_i) = \left| \frac{G(0, \sqrt{\Omega_n(f_i)}) + i G(0, \sqrt{\Omega_n(f_i)})}{\sqrt{2}} \right|$$

For **each data segment and frequency** we generate a **gaussian realization**.

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For **each data segment and frequency** we generate a **gaussian realization**.

Given that:

- LISA will be operating for **4yrs (75% efficiency)**
- We choose **data segments** of **roughly 12 days**

we conclude that:

- Roughly **95 independent measurements at each frequency**.
- The **resolution** of the detector is **roughly 10^{-6}Hz**

Data pre-processing and likelihood

Starting from $D_c(f_i)$ (our data), defined as:

$$D_c(f_i) \equiv \langle \tilde{d}_c^2(f_i) \rangle = \langle (\tilde{s}_c(f_i) + \tilde{n}_c(f_i))^2 \rangle = \langle \tilde{s}_c^2(f_i) \rangle + \langle \tilde{n}_c^2(f_i) \rangle .$$

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- We **average over the (95) data segments**:

This leaves us with some $D(f_i)$ (the averaged data) and an estimate of the error $\sigma(f_i)$ (the standard deviation or the data).

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- We **coarse grain** the data:
i.e. from the initial linear 10^{-6} Hz spacing ($\sim 5 \times 10^5$ points)
→ we go to some final (and less dense) set of frequencies f_i
This leaves us with the **final data set D_i and errors σ_i** .

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Finally we assume the data to be described by the likelihood:

$$\mathcal{L}(\vec{\theta}, \vec{n}) \propto \exp \left[-\frac{N_{chunks}}{2} \sum_i \left(\frac{D_i - h^2 \Omega_{GW}(f_i, \vec{\theta}) - h^2 \Omega_n(f_i, \vec{n})}{\sigma_i} \right)^2 \right]$$

with i labeling the data points and Ω_{GW} , Ω_n models for signal and noise.

SGWBinner algorithm

Based on two LISA COSWG projects:

- 1906.09244: C. Caprini, D. Figueroa, R. Flauger, M.P., G. Nardini, M. Peloso, A. Ricciardone, G. Tasinato
- 2009.11845: R. Flauger, N. Karnesis, G. Nardini, M. P., A. Ricciardone, J. Torrado

We look for **best approximation** of the signal **with a multi-PL**

$$h^2 \Omega_{\text{GW}}(f, \vec{\theta}) = \sum_i 10^{\alpha_i} \left(\frac{f}{\sqrt{f_{\min,i} f_{\max,i}}} \right)^{p_i} \Theta(f - f_{\min,i}) \Theta(f_{\max,i} - f) .$$

where Θ is the Heaviside step function.

N bins $\rightarrow 4N (f_{\min,i}, f_{\max,i}, \alpha_i, p_i) + 2N$ (noise) parameters.

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The basic procedure is composed of four steps

- ① Build a robust prior for the noise model
(to force bin-by-bin measurements)
- ② Divide the frequency range in a set of bins and reconstruct the signal
- ③ Merge as many bins as possible (to avoid overfitting)
- ④ Define a procedure to compute the error on the reconstruction
- ⑤ Final MCMC run with common noise parameters

Few more detail on steps 1 and 3 ...

Characterizing the noise

Bad noise
reconstruction



- No detection/false detections
- Bad parameter reconstruction

Some useful observations:

- Noise parameters are correlated over the full frequency range!
- Noise is expected to dominate at small and at large frequencies
- Noise dominates over signal in TT

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As a consequence (single channel version):

- 1 We divide the range into three bands
(small and large frequencies + a central band)
- 2 Estimate signal and noise parameters in the external bands
- 3 Use this as a prior for the measurements in the bins in the central part

As a consequence (three channels version):

- 1 Estimate signal and noise parameters in TT
- 2 Use this as a prior for the AA/EE.

Criterion for merging

Two motivations for merging:

- Larger N means smaller bins which implies larger errors
- For large values of N , the reconstruction with N bins (i.e. $2N$ parameters) **may overfit** the signal

Typically **reducing** the number N of bins may **improve the analysis!**

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AIC for model comparison we use:

$$\text{AIC} = -2 \ln \mathcal{L} + 2k = \chi^2 + 2k$$

For a couple of consecutive bins $(i, i + 1)$ we can compute

$$\Delta \text{AIC} = \text{AIC}_{\text{after merging}} - \text{AIC}_{\text{before merging}} = \chi_{\text{after merging}}^2 - \chi_{\text{before merging}}^2 - 2k_{1\text{-bin}}$$

According to the AIC definition:

$$\Delta \text{AIC} < 0$$

→

It is convenient to merge the two bins

A more accurate likelihood

A **Gaussian likelihood** would give a **systematic low bias!**

(astro-ph/9808264, astro-ph/0205387, astro-ph/0302218, 0801.0554)

Consider the Gaussian likelihood:

$$\ln \mathcal{L}_G(\vec{\theta}, \vec{n}) \propto -\frac{N_{\text{chunks}}}{2} \sum_{i,j} \sum_k w_{ij}^{(k)} \left(\frac{D_{ij}^{(k)} - h^2 \Omega_{\text{GW}}(f_{ij}^{(k)}, \vec{\theta}) - h^2 \Omega_n(f_{ij}^{(k)}, \vec{n})}{h^2 \Omega_{\text{GW}}(f_{ij}^{(k)}, \vec{\theta}) + h^2 \Omega_n(f_{ij}^{(k)}, \vec{n})} \right)^2$$

and the Lognormal likelihood:

$$\ln \mathcal{L}_{LN}(\vec{\theta}, \vec{n}) \propto -\frac{N_{\text{chunks}}}{2} \sum_{i,j} \sum_k w_{ij}^{(k)} \ln^2 \left(\frac{h^2 \Omega_{\text{GW}}(f_{ij}^{(k)}, \vec{\theta}) + h^2 \Omega_n(f_{ij}^{(k)}, \vec{n})}{D_{ij}^{(k)}} \right)$$

Then we define our likelihood as (astro-ph/0302218, 2009.11845)

$$\ln \mathcal{L} = \frac{1}{3} \ln \mathcal{L}_G + \frac{2}{3} \ln \mathcal{L}_{LN}$$

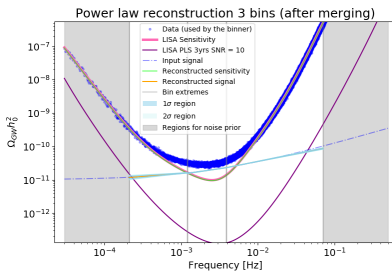
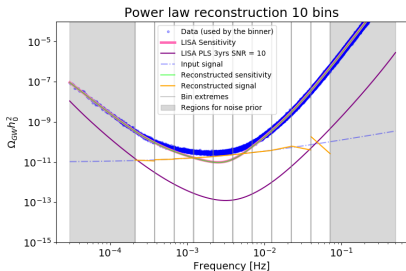
which removes the skewness contributions and thus is more accurate.

Some examples

Linear signal + “LIGO binaries”

As a first example let us consider

$$h^2 \Omega_{\text{GW}}(f) = h^2 \Omega_{\text{GW, const}}(f) + h^2 \Omega_{\text{GW, BHB+NSB}}(f) = 10^{-11} + 5.4 \times 10^{-12} \left(\frac{f}{0.001} \right)^{2/3}$$



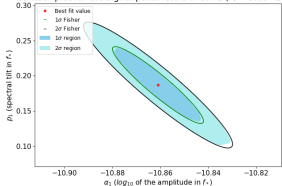
After the merging procedure only 3 bins with small error bands are left.

Some examples

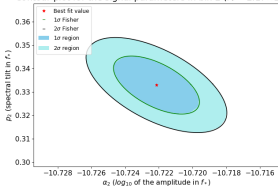
Linear signal + “LIGO binaries” contour plots

Let us have a closer look at the contour plots in each bin

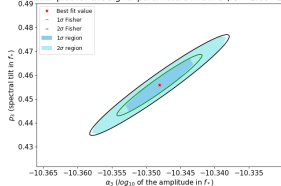
Contour plot for the signal parameters in bin 1 ($f_s = 5.03 \times 10^{-4}$)



Contour plot for the signal parameters in bin 2 ($f_s = 2.17 \times 10^{-3}$)



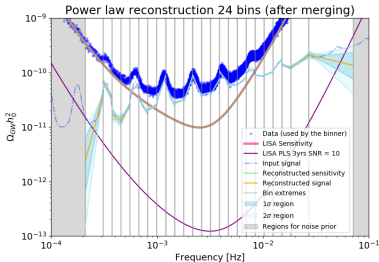
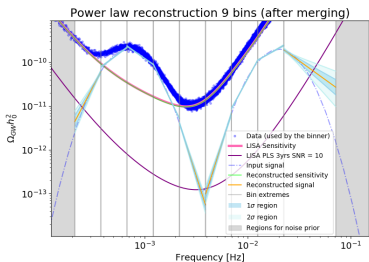
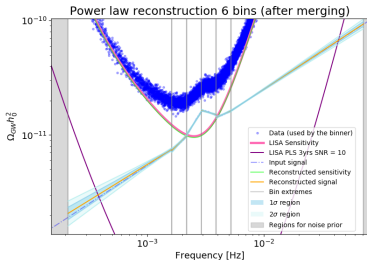
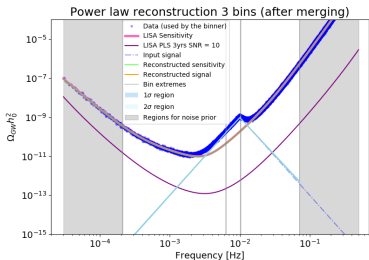
Contour plot for the signal parameters in bin 3 ($f_s = 1.66 \times 10^{-2}$)



The three contour plots clearly show a **progressive increase in the slope**
Consistent with the values of the ratio $\Omega_{\text{GW,const}}/\Omega_{\text{GW,binaries}}$ in the three bins!

Some examples

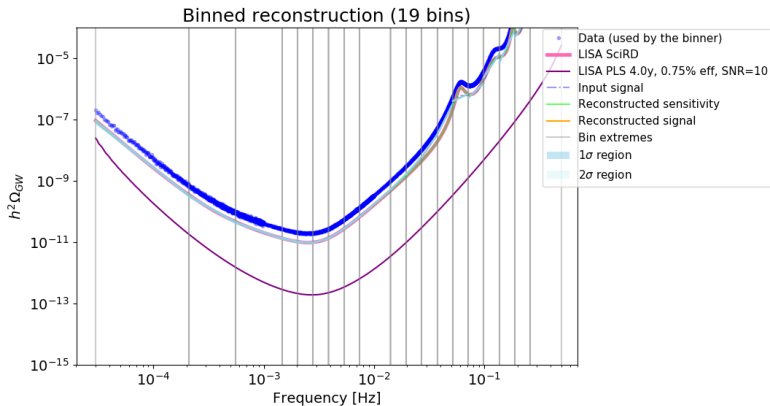
More cases



Some examples

Degenerate cases

Since the **noise in TT** has a different shape
this may help in **breaking degeneracies!**



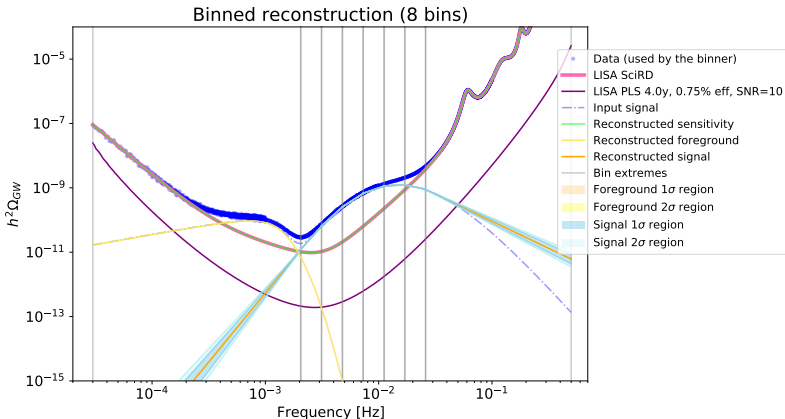
Some examples

Foreground removal

The code can now account for perform component separation too!

Consider for example the signal due to Galactic binaries:

$$h^2 \Omega_{\text{GW}} = 10^{\alpha_{\text{FG}}} f^{2/3} e^{-a_1 f + a_2 f \sin(a_3 f)} \{1 + \tanh[a_4(f_k - f)]\} .$$



An exact solution for the parameters

Based on 2004.01135, in collaboration with Enrico Barausse

If the **model** (for both signal and noise) is **linear in the $\vec{\theta}$** :

- The log **likelihood is quadratic** in the parameters
- The **Fisher matrix does not depend on the parameters**
- Finding the **best fit** reduces to **solving a linear equation**

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Starting from:

$$-\ln \mathcal{L}(\vec{\theta}, \vec{n}) \propto \frac{1}{2} \sum_i \left(\frac{D_i - M(f_i, \vec{\theta})}{\sigma_i} \right)^2,$$

for a linear model we get:

$$F_{lk} \equiv \sum_i \frac{1}{\sigma_i^2} \frac{\partial M(f_i, \vec{\theta})}{\partial \theta_l} \frac{\partial M(f_i, \vec{\theta})}{\partial \theta_k}, \quad \bar{\theta}_l = F_{lk}^{-1} \sum_i \frac{1}{\sigma_i^2} D_i \frac{\partial M(f_i, \vec{\theta})}{\partial \theta_k}$$

where $\bar{\theta}_l$ is the MLE for the parameters.

A simple model for the signal

Assume the signal can be expressed as:

$$S(f) = \sum_{j=1}^n a_j \delta_w(f - f_j),$$

where:

- a_j are the parameters
- w is some correlation length
- $\delta_w(f - f_j)$ are some functions

The choice of $\delta_w(f - f_j)$ defines a **basis to express the signal**.

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The choice of $\delta_w(f - f_j)$ defines a **basis to express the signal**.

Depending on the choice of w we have two regimes:

- Small w : the measurements in f_j are not correlated
- Large w : the measurements in f_j are correlated

Properly **choosing w** we can **smooth the signal!**

Principal component analysis

It is interesting to notice that:

- In general the parameters a_j are correlated
- Eigenvectors $e_j^{(i)}$ of F_{lk} are uncorrelated combinations of a_j .
- Eigenvalues $\lambda^{(i)}$ of F_{lk} give the information on the $e_j^{(i)}$.

Principal component analysis

It is interesting to notice that:

- In general the parameters a_j are **correlated**
- **Eigenvectors** $e_j^{(i)}$ of F_{lk} are **uncorrelated** combinations of a_j .
- **Eigenvalues** $\lambda^{(i)}$ of F_{lk} give the **information on the** $e_j^{(i)}$.

Principal Component Analysis (PCA):

- 1 Compute the eigensystem of F_{lk}
- 2 Cut $e_j^{(i)}$ corresponding to $\lambda^{(i)}$ smaller than some threshold
- 3 Project $\delta_w(f - f_j)$ and a_j on this subset of $e_j^{(i)}$ (say $\eta_k(f)$, b_k)
- 4 Reconstruct the signal as: $S(f) = \sum_k b_k \eta_k(f)$

Corresponds to **reconstructing** the signal in terms of the **components which can be well determined!**

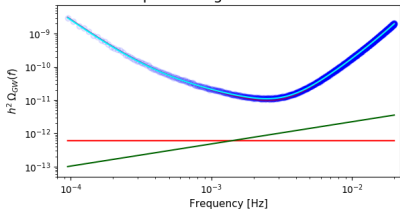
In the following plots all parameters are normalized to 1!

Some examples

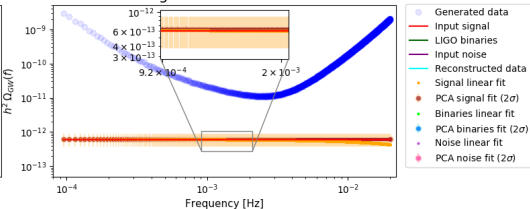
Subtracting the foreground 1

Flat signal (SNR ~ 30) + LIGO binaries (gaussian prior $\sigma = 0.5$)

Inputs and generated data

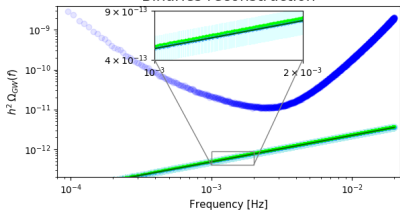


Signal reconstruction

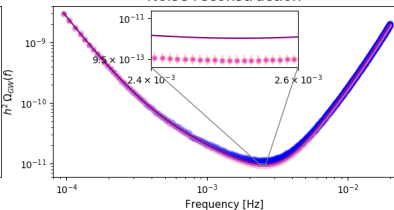


- Generated data
- Input signal
- LIGO binaries
- Input noise
- Reconstructed data
- Signal linear fit
- PCA signal fit (2σ)
- Binaries linear fit
- PCA binaries fit (2σ)
- Noise linear fit
- PCA noise fit (2σ)

Binaries reconstruction



Noise reconstruction

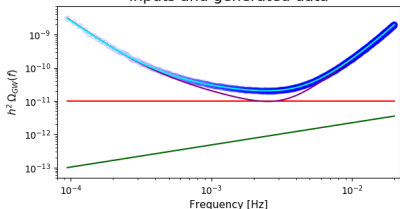

 $L = 1.011 \pm 0.111$, $A \simeq 0.972 \pm 0.005$ and $O \simeq 0.973 \pm 0.001$

Some examples

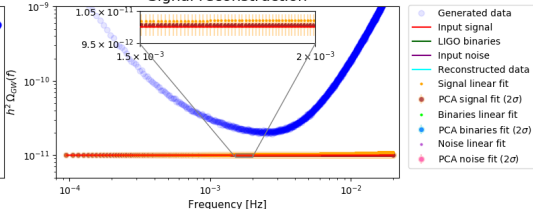
Subtracting the foreground 2

Flat signal (SNR ~ 520) + LIGO binaries (gaussian prior $\sigma = 0.5$)

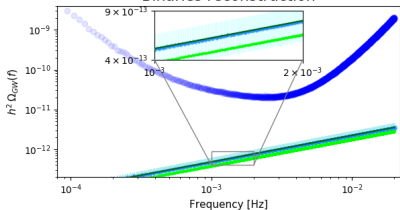
Inputs and generated data



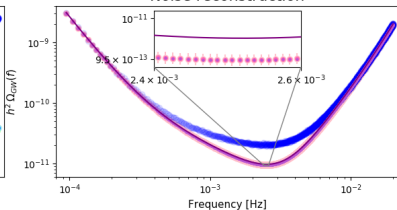
Signal reconstruction



Binaries reconstruction



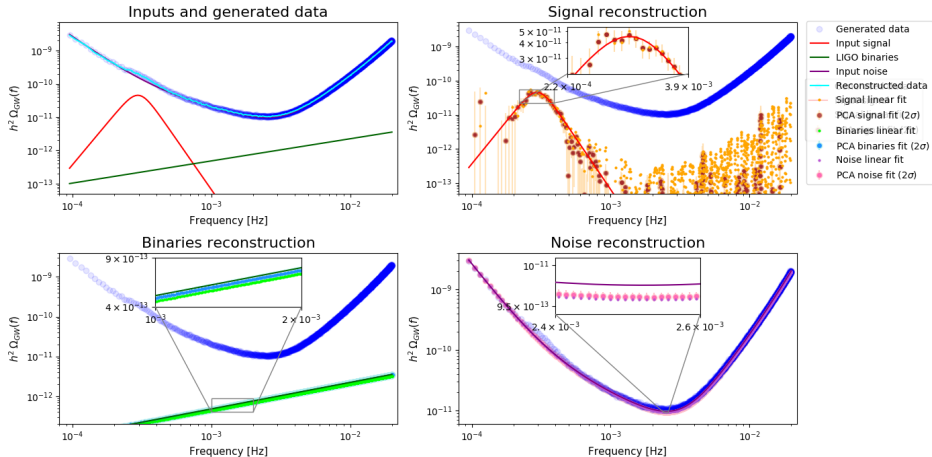
Noise reconstruction



$$L = 0.989 \pm 0.141, A = 0.971 \pm 0.005, O = 0.977 \pm 0.001$$

Some examples

No degeneracy 1

Broken PL (SNR ~ 30) + LIGO binaries (gaussian prior $\sigma = 0.5$)

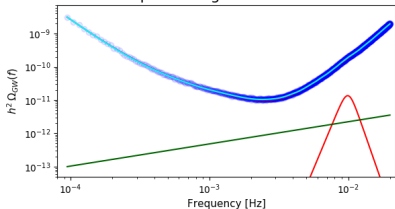
$$L \simeq 0.960 \pm 0.033, A \simeq 0.990 \pm 0.004, O \simeq 0.976 \pm 0.001$$

Some examples

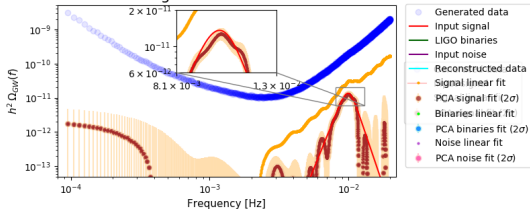
No degeneracy 2

Another broken PL (SNR ~ 30) + LIGO binaries (gaussian prior $\sigma = 0.5$)

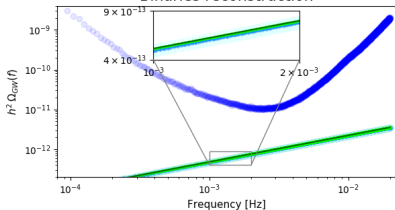
Inputs and generated data



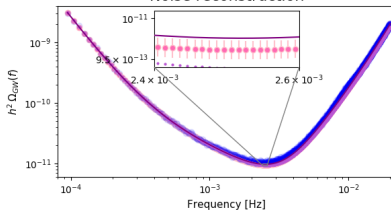
Signal reconstruction



Binaries reconstruction



Noise reconstruction



$$L \simeq 0.972 \pm 0.054, A \simeq 0.987 \pm 0.008, O \simeq 0.982 \pm 0.002$$

Conclusions and future perspective

Conclusions

- PLS is (qualitatively) useful but not the end of the story
- SGWBinner: a flexible algorithm to reconstruct general signal (multiple power laws, bumps,...)
- PCA reconstruction: an alternative approach for SGWB reconstruction
- In both cases it's possible to perform component separation

Future perspectives

- Keep improving on detector modeling
- Application to concrete case (inflation, phase transitions, ...)
- New techniques?

Last Slide

The End

Thank you

Bayes theorem and data analysis

Given some **events** D and $\vec{\theta}$ we define:

$$P(D|\theta_j) \equiv \frac{P(D \cap \theta_j)}{P(\theta_j)}$$

the conditional probability of D occurring given $\vec{\theta}$.

Since $P(D \cap \vec{\theta}) = P(\vec{\theta} \cap D)$ we get the Bayes theorem:

$$P(\vec{\theta}|D) = \frac{P(D|\vec{\theta}) \cdot P(\vec{\theta})}{P(D)} \propto P(D|\vec{\theta}) \cdot P(\vec{\theta})$$

D is the set of data and $\vec{\theta}$ is the vector of parameters of the theory.

- $P(\vec{\theta}|D)$ are the *Posterior probabilities*
- $P(\vec{\theta})$ are the *Prior probabilities*
- $P(D|\vec{\theta})$ is the *Likelihood* which from now on is denoted $\mathcal{L}(\vec{\theta})$
- $P(D)$ is the *Model Evidence*

Some statistical tools

- The maximum likelihood estimate of **best fit parameters** $\vec{\theta}_0$ is

$$\partial_{\vec{\theta}} \ln \mathcal{L}(\vec{\theta}_0) = 0$$

- The **Fisher matrix** (*i.e.* the inverse of the **covariance matrix** C_{ij}) is

$$F \equiv C_{ij}^{-1} = -\langle \partial_i \partial_j \ln \mathcal{L} |_{\vec{\theta}_0} \rangle$$

- **Gaussian approximation of \mathcal{L}** around $\vec{\theta}_0$

$$\mathcal{L}(\vec{\theta}) \simeq \frac{1}{\sqrt{\det(2\pi C)}} \exp \left\{ -\frac{1}{2} (\vec{\theta} - \vec{\theta}_0)^T C^{-1} (\vec{\theta} - \vec{\theta}_0) \right\}$$

- **Confidence intervals** are obtained by solving

$$-2 \left[\ln \mathcal{L}(\vec{\theta}) - \ln \mathcal{L}(\vec{\theta}_0) \right] = \nu_{n\sigma}(k)$$

- **Akaike Information Criterion (AIC)** for a fit with k parameters:

$$\text{AIC} = -2 \ln \mathcal{L} + 2k = \chi^2 + 2k$$